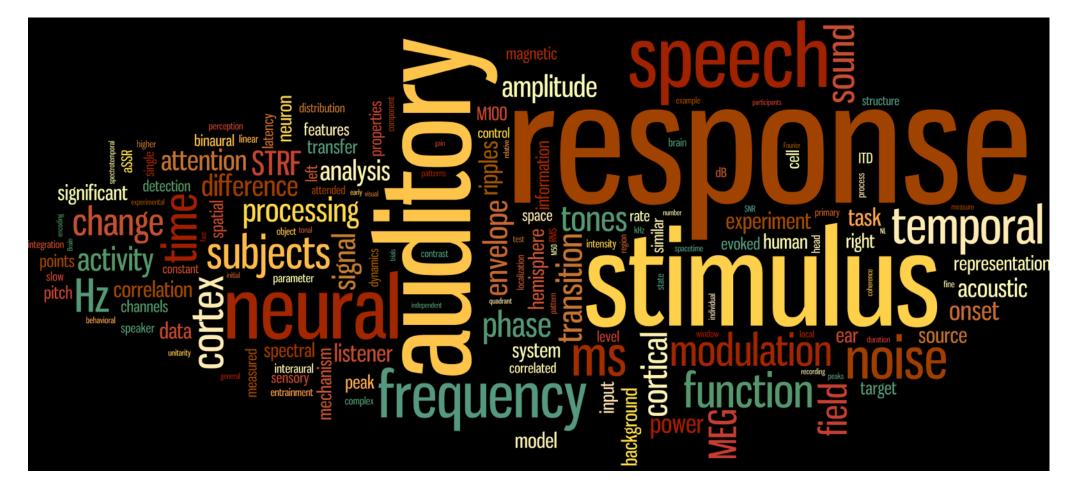
Signal Analysis Primer and Applications

Jonathan Z. Simon University of Maryland

> Simons Foundation 12 December 2014

- MEG-based Auditory Neuroscience
 - Cocktail-Party Auditory Processing
 - Role of Attention
 - Neural Representations of Speech
 - Fundamentally Temporally Neural Representations
- More at <http://www.isr.umd.edu/Labs/CSSL/simonlab/>

Research Background



- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

- Every Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & *frequency span* (f_s)
- Reciprocal relationship: frequency resolution (Δf) & time span (T)

$$x[t] = \frac{1}{N} \sum_{k=0}^{N-1} X[f_k] e^{i2\pi f_k t} \text{ where}$$

$$t = \underbrace{0, \Delta t, 2\Delta t, \dots, T - \Delta t}_{N}$$

$$f_{k} = \underbrace{0, \Delta f, 2\Delta f, \dots, f_{s} - \Delta f}_{N}$$

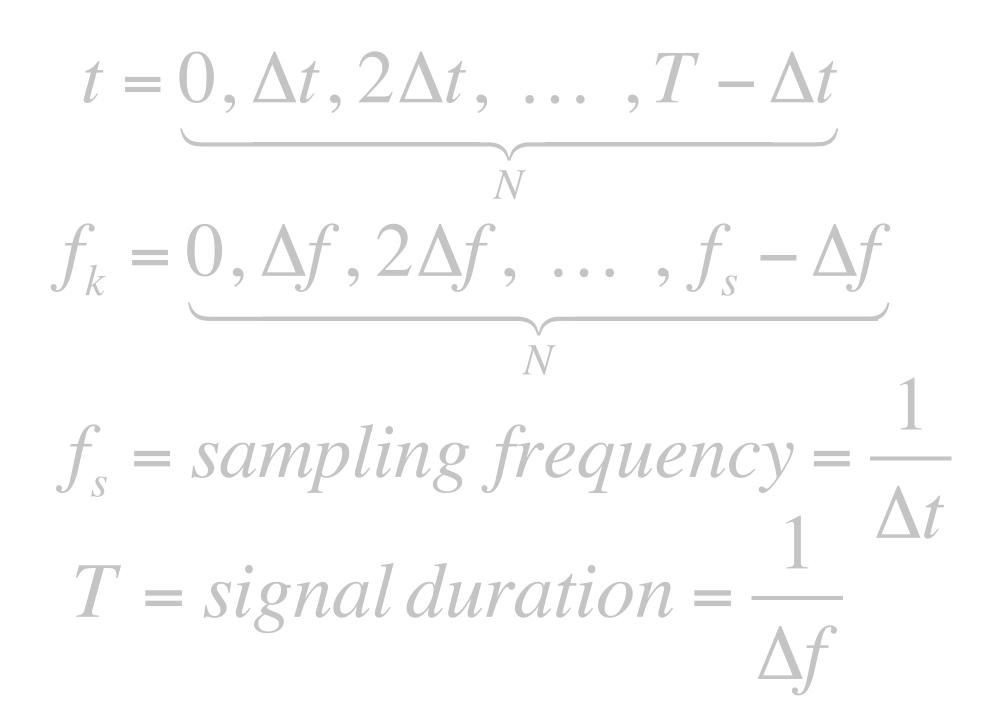
$$f_{s} = sampling frequency = \frac{1}{\Delta t}$$

$$T = signal duration = \frac{1}{\Delta t}$$

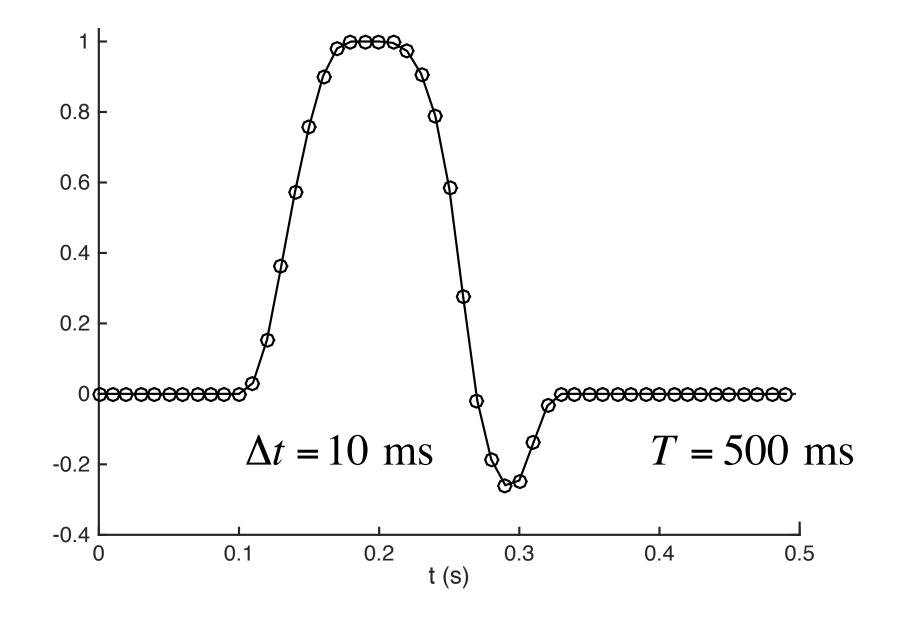
 Δf

- Every Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & *frequency span* (f_s)
- Reciprocal relationship: frequency resolution (Δf) & time span (T)

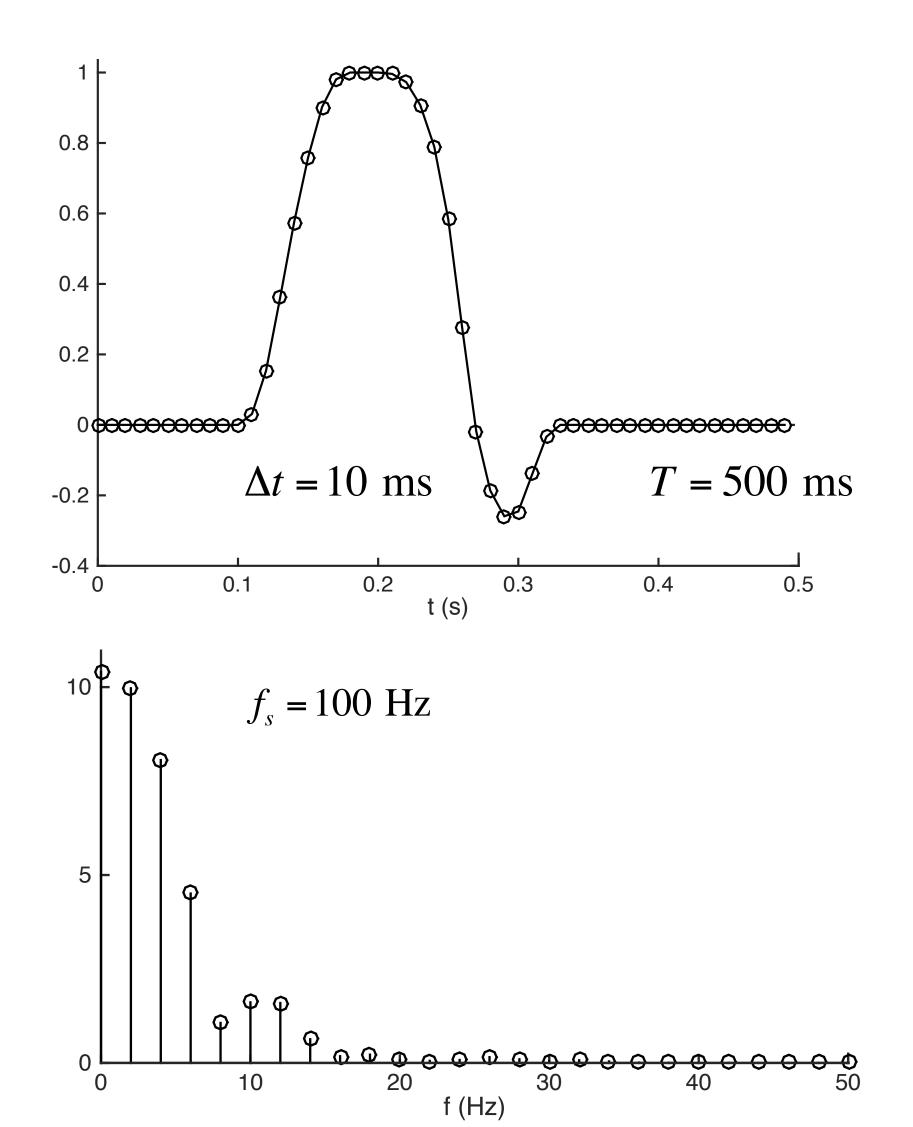




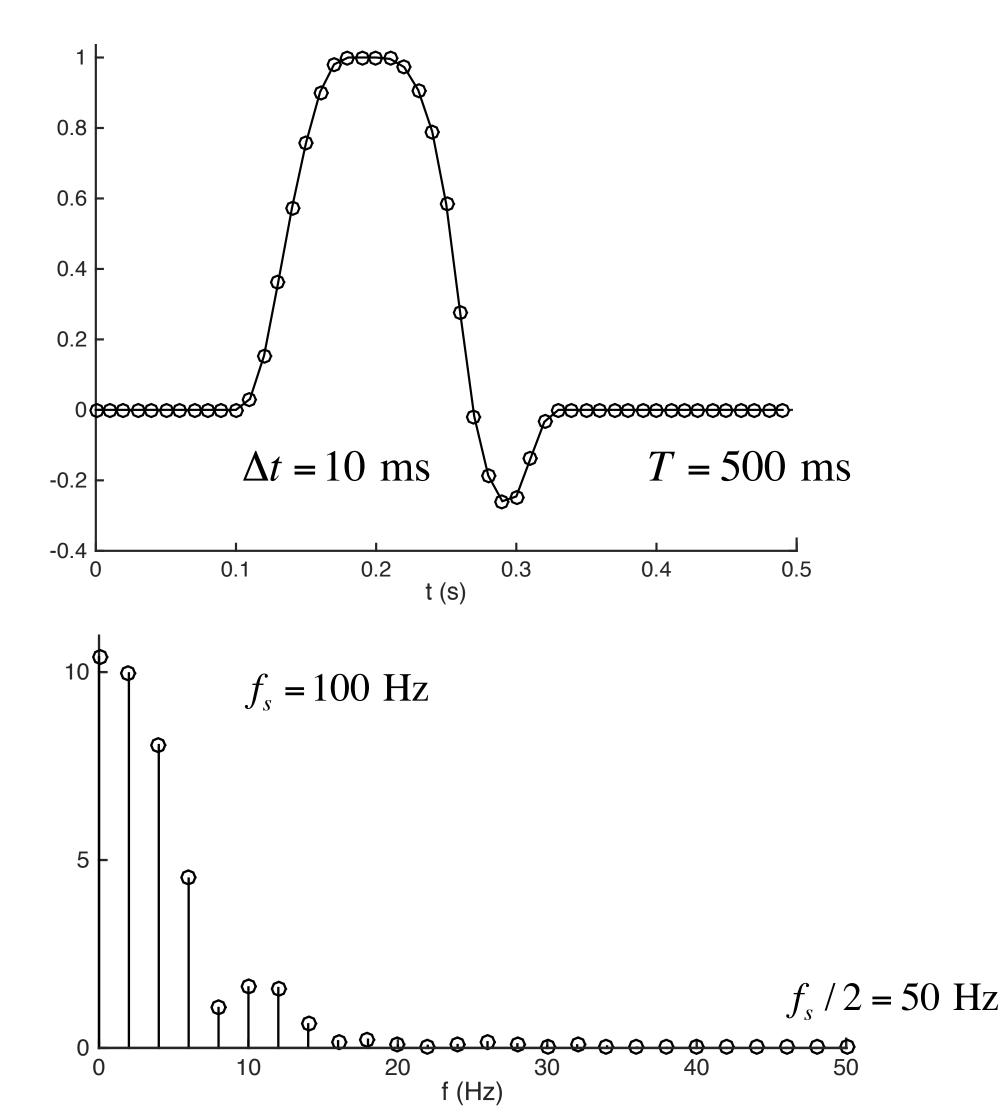
- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δ*t*) & *sample frequency* (*f_s*)
- Reciprocal relationship: frequency resolution (Δf) & duration (T)



- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δ*t*) & *sample frequency* (*f_s*)
- Reciprocal relationship: frequency resolution (Δf) & duration (T)

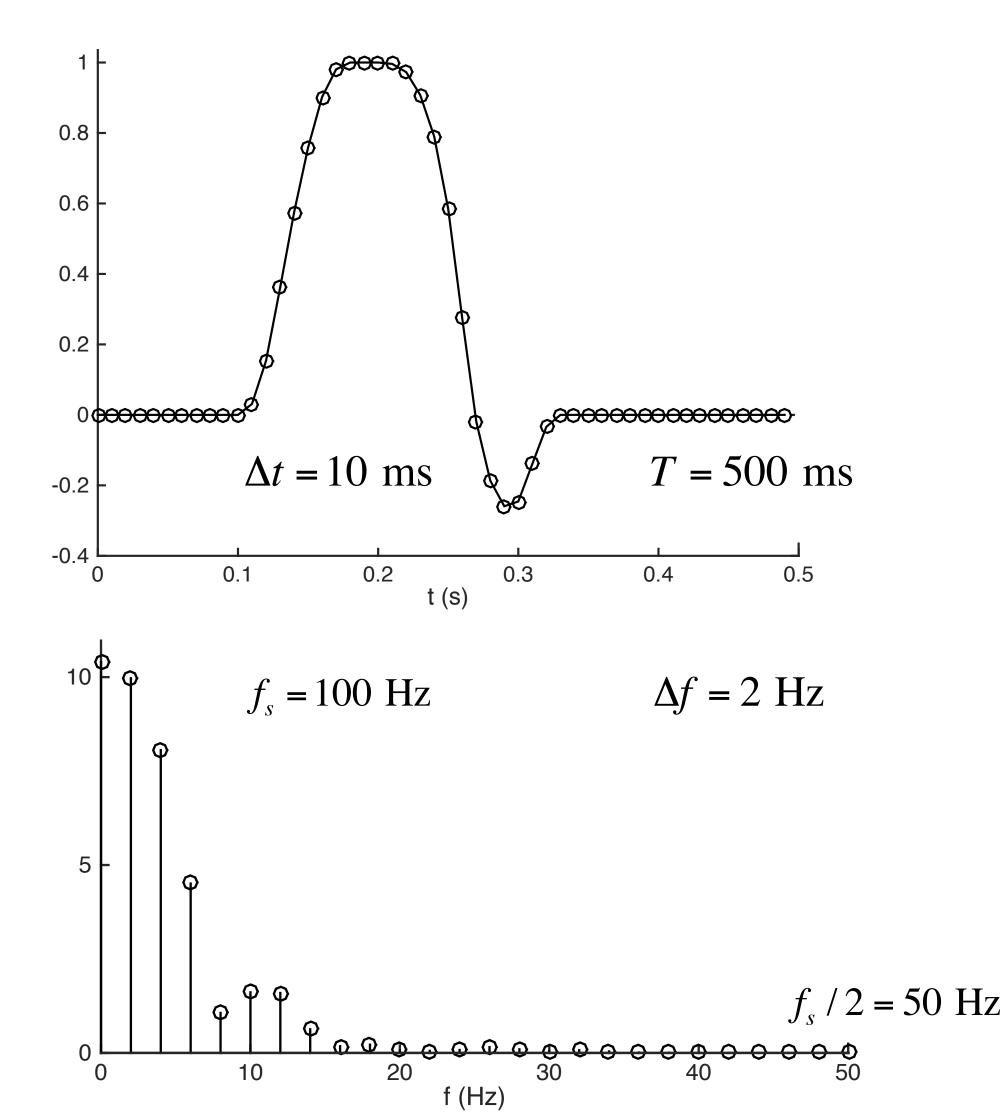


- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & sample frequency (f_S)
- Reciprocal relationship: *frequency* resolution (Δf) & duration (T)



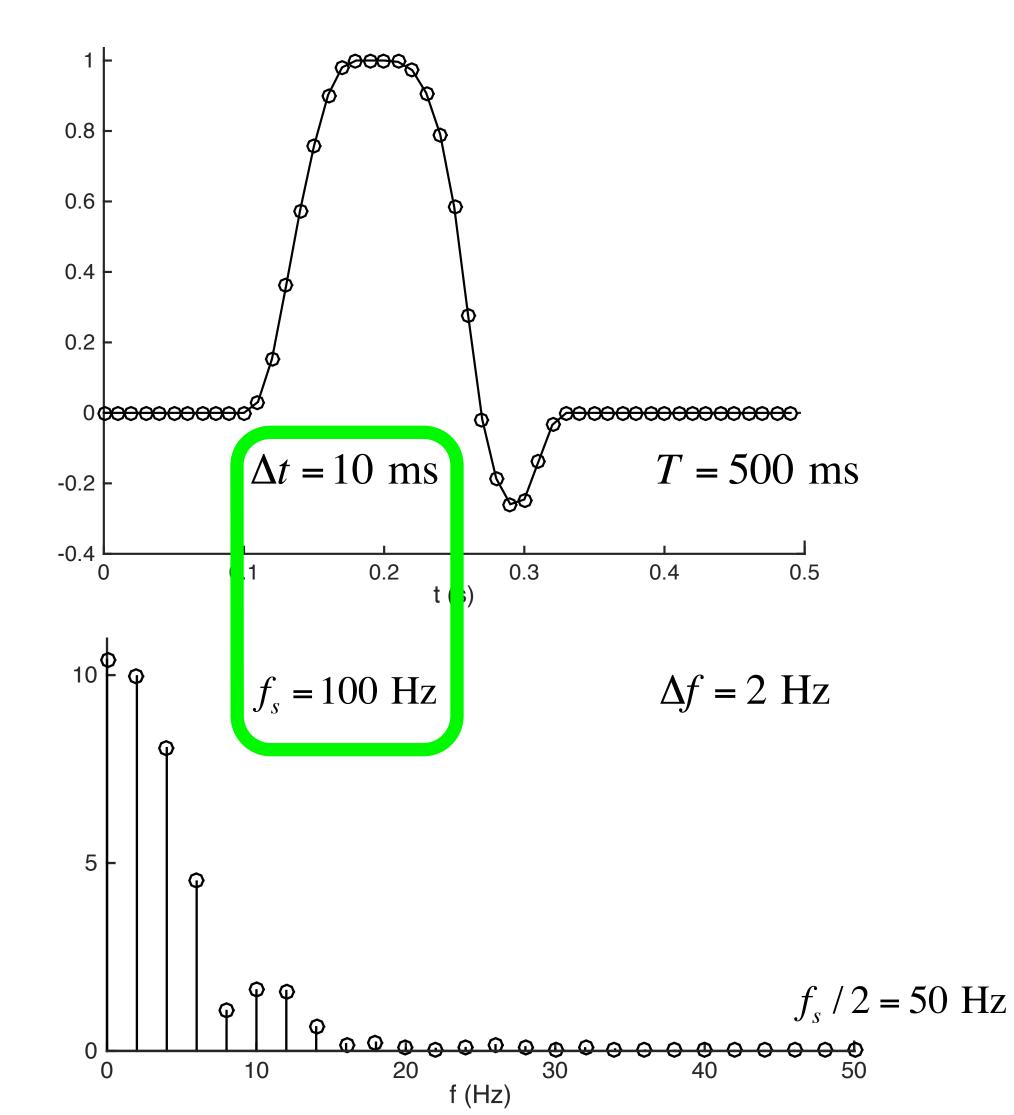


- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & sample frequency (f_S)
- Reciprocal relationship: *frequency* \bullet resolution (Δf) & duration (T)



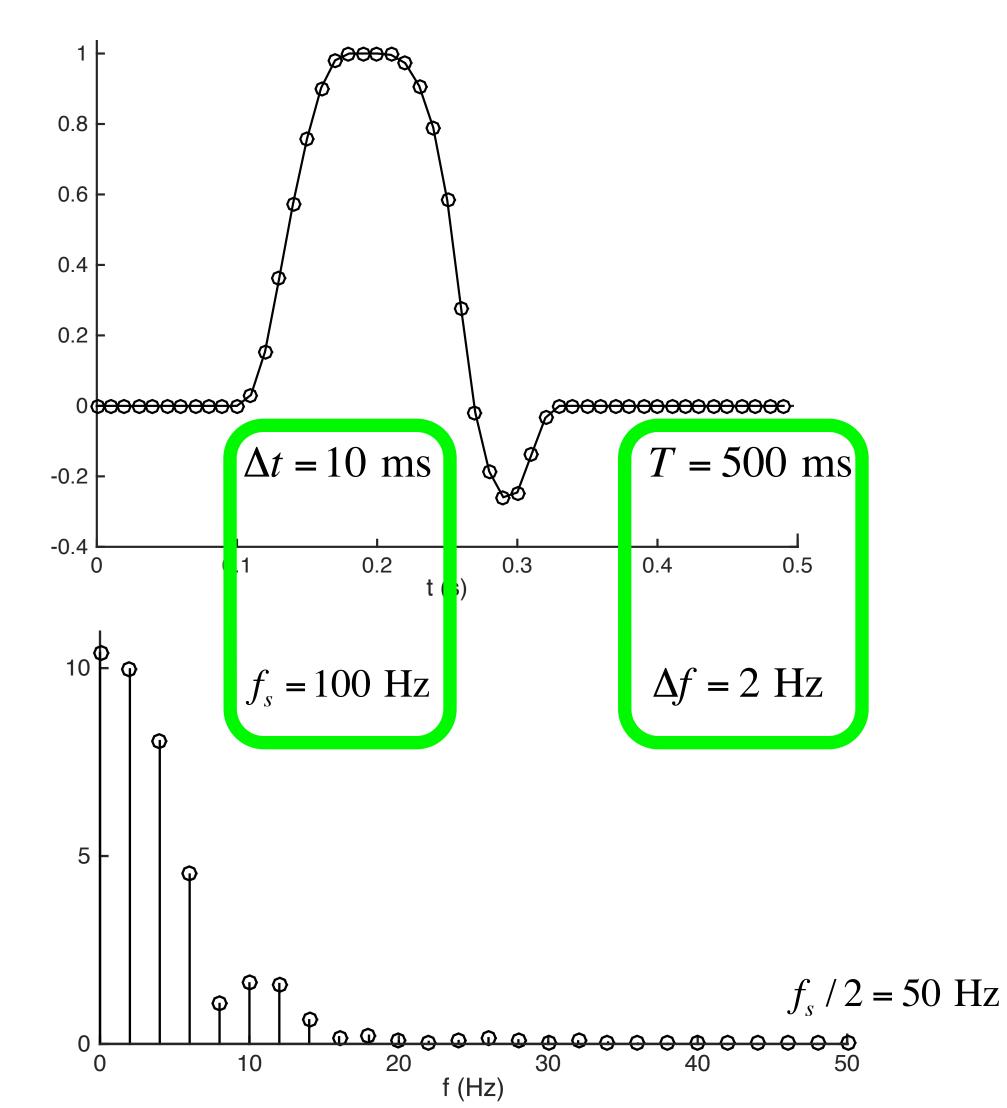


- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & sample frequency (f_S)
- Reciprocal relationship: *frequency* resolution (Δf) & duration (T)



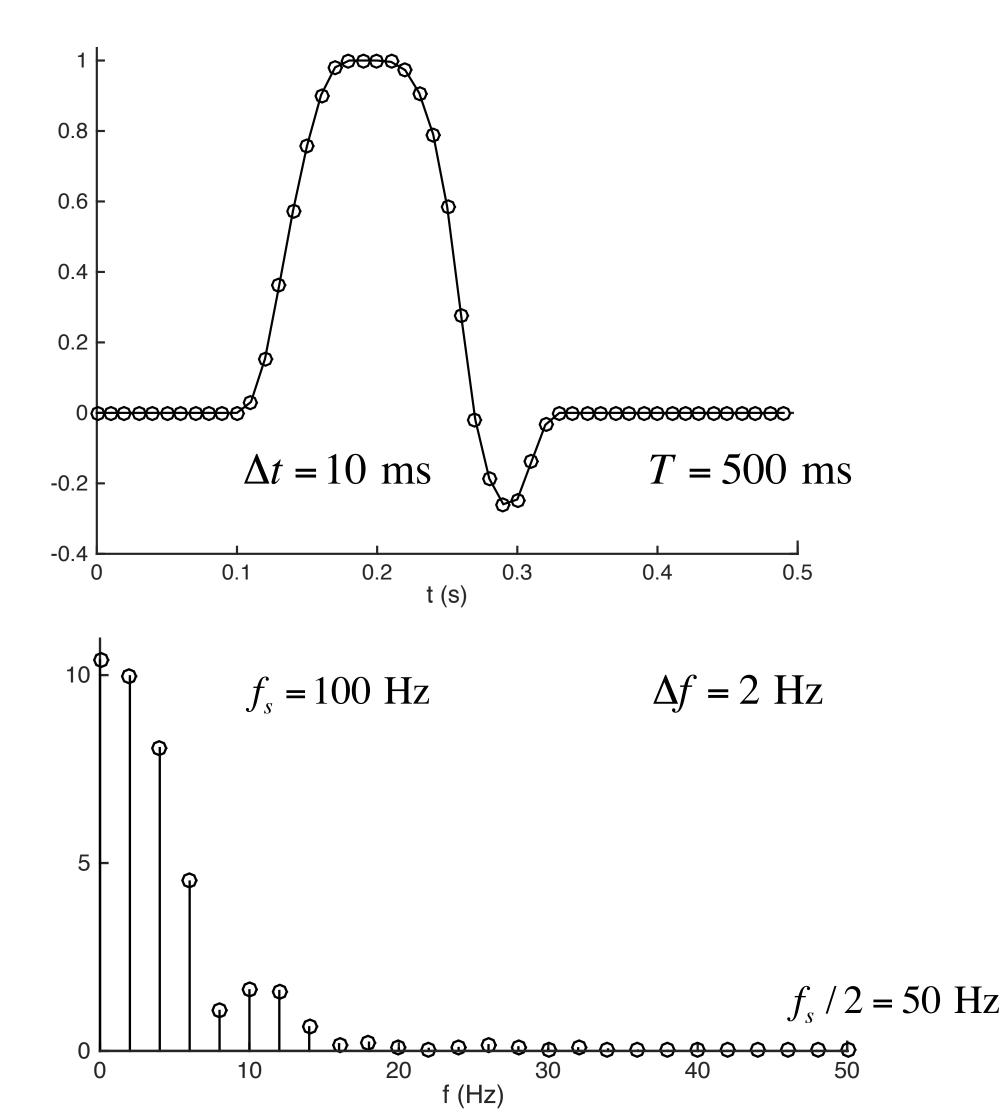


- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & sample frequency (f_S)
- Reciprocal relationship: *frequency* resolution (Δf) & duration (T)

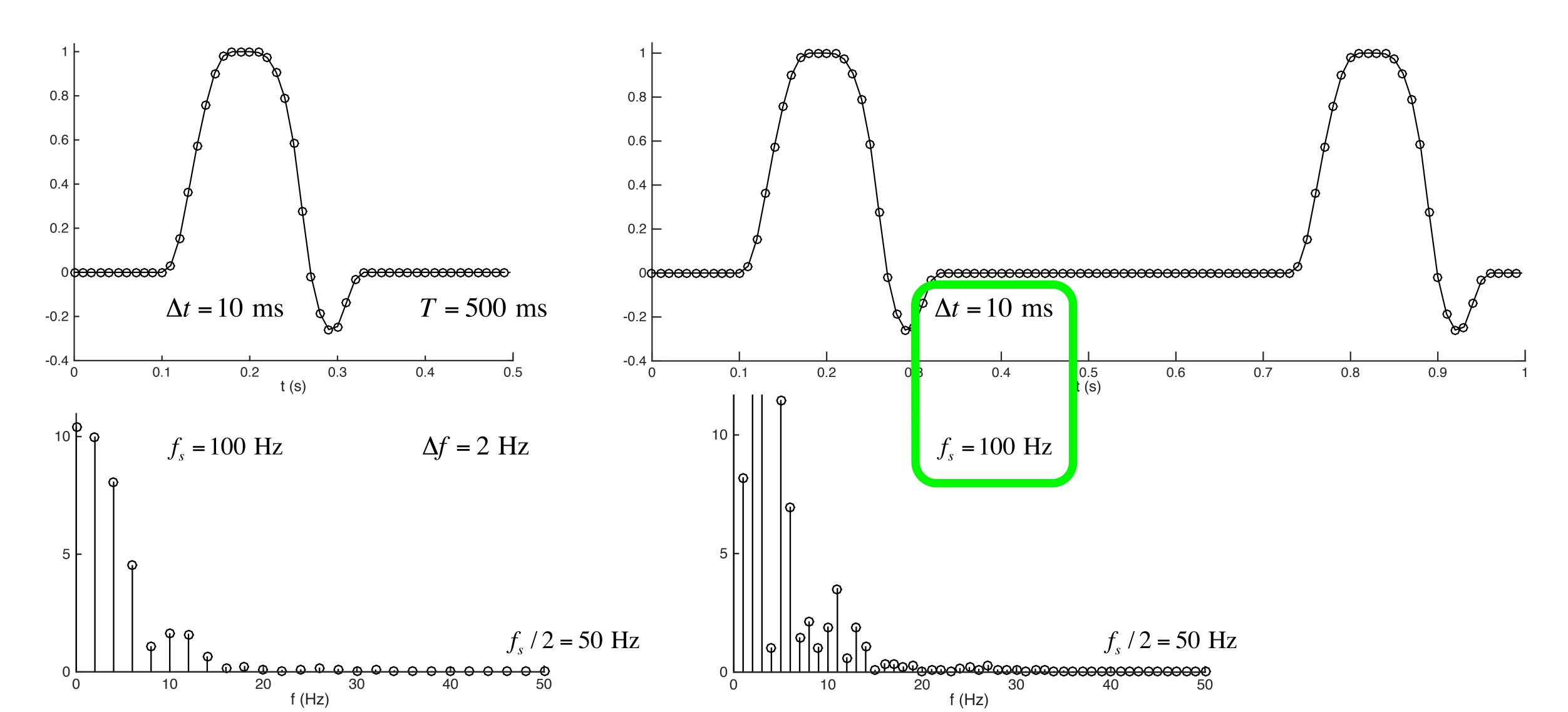


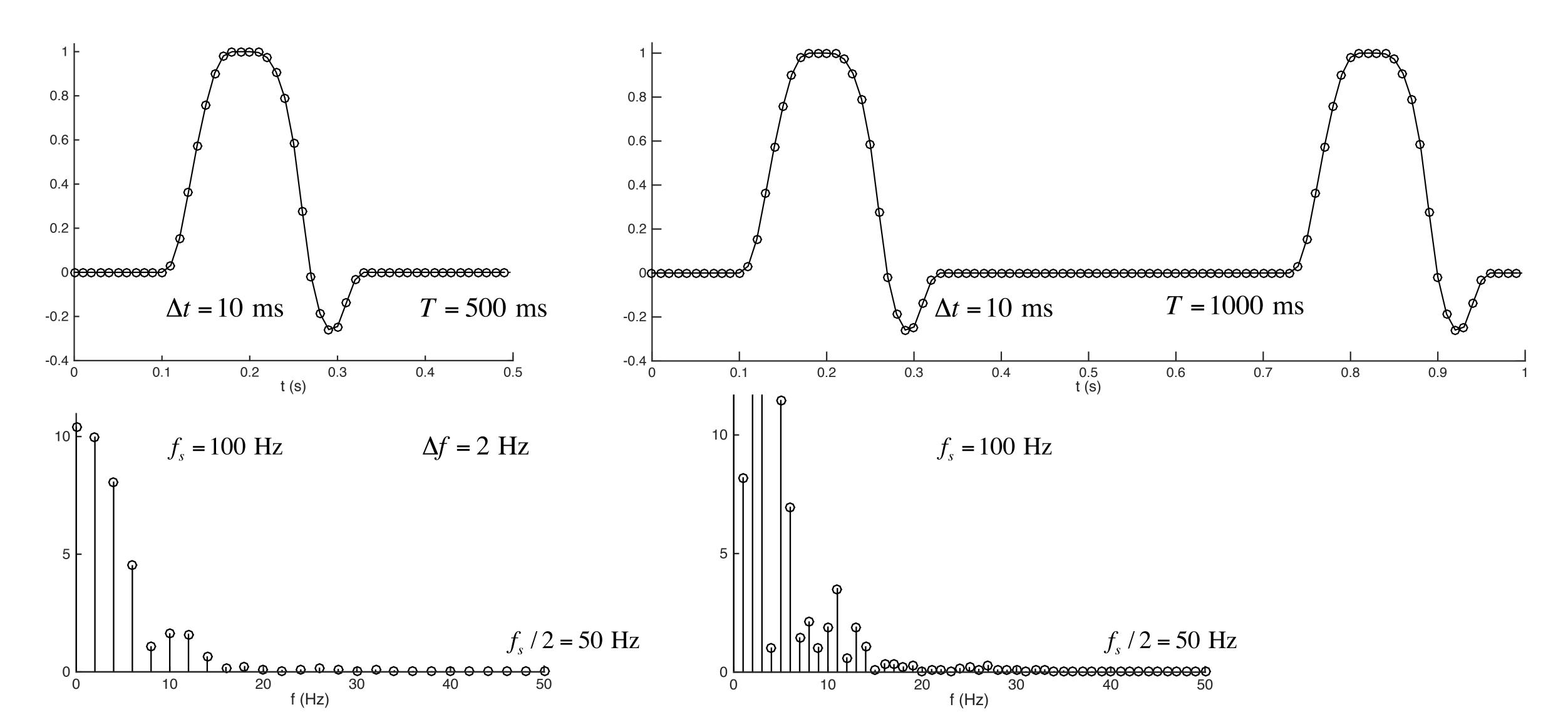


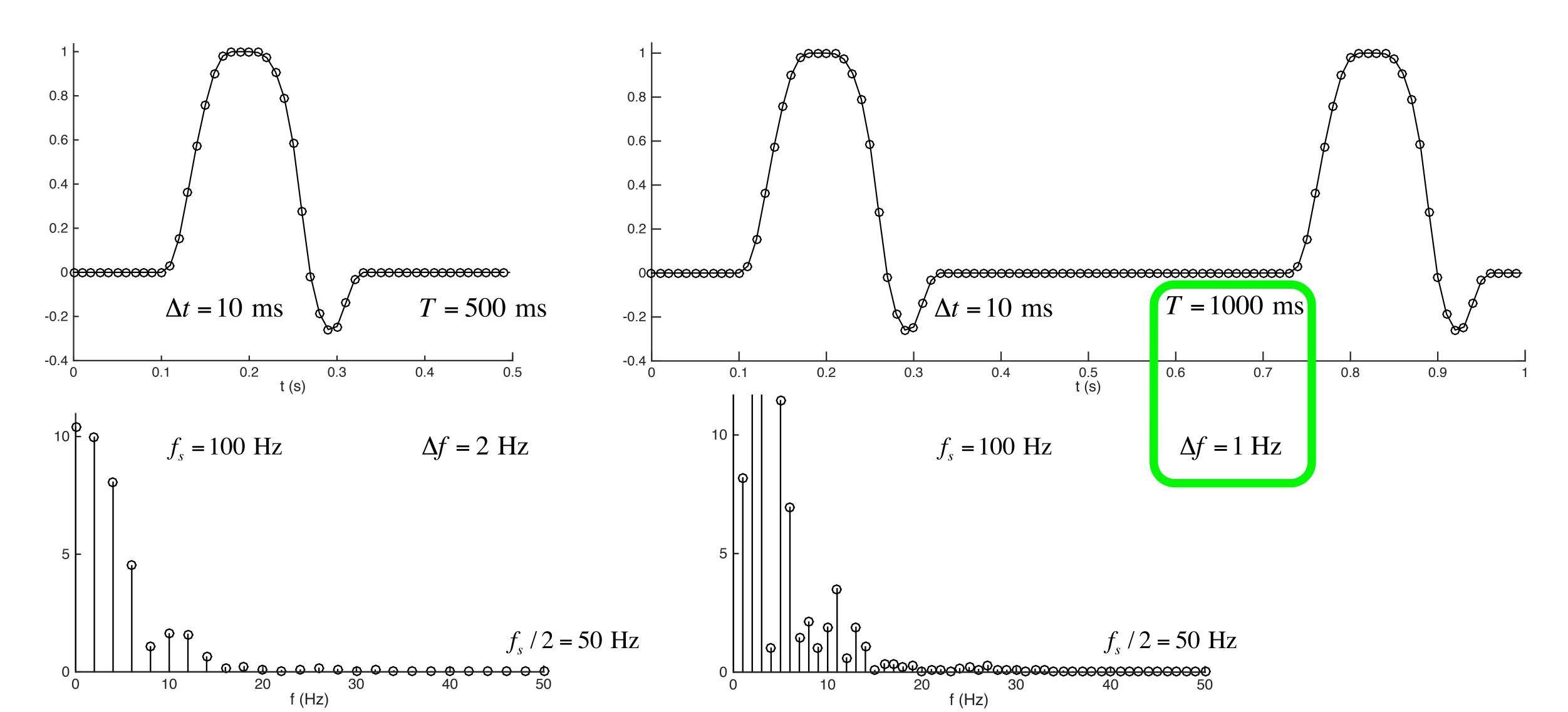
- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & sample frequency (f_S)
- Reciprocal relationship: *frequency* resolution (Δf) & duration (T)



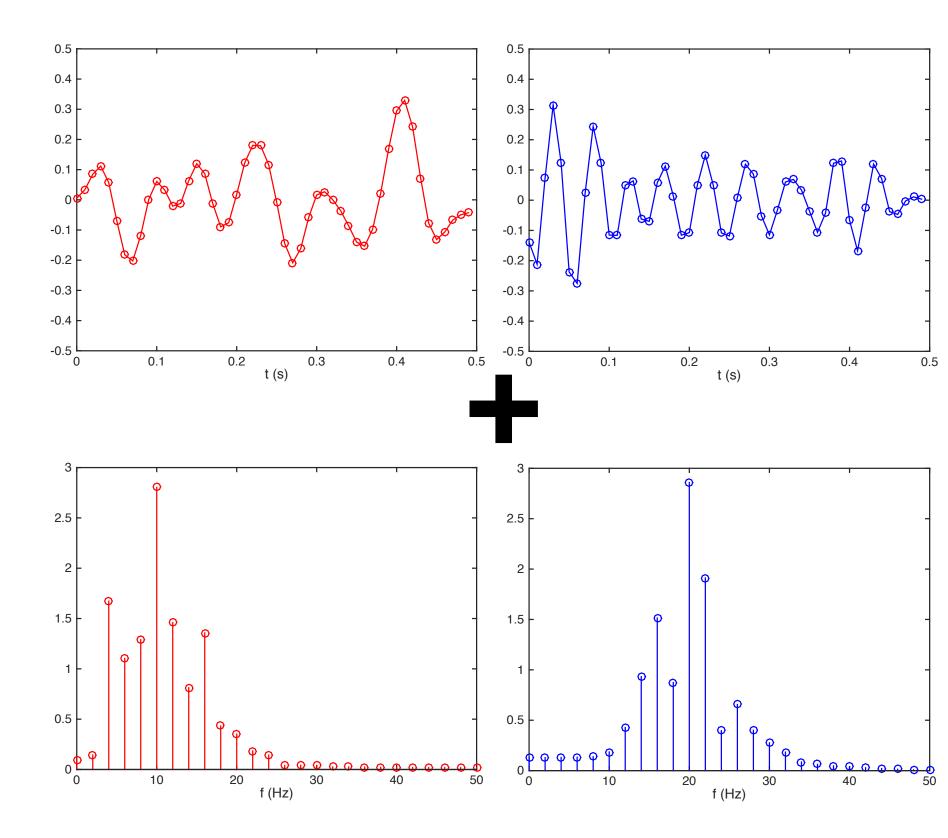




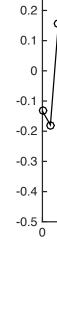




- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less

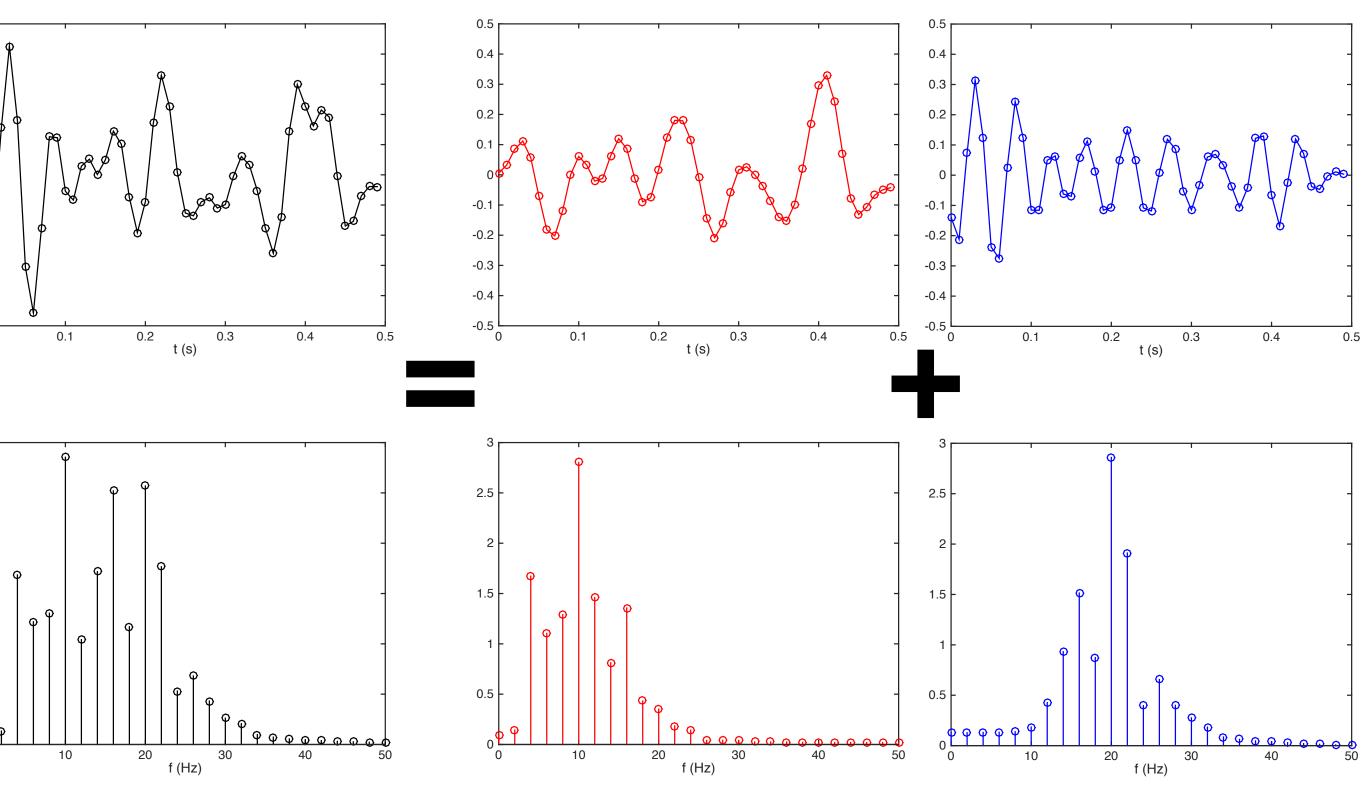


- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less

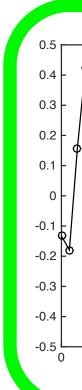


0.5

0.3

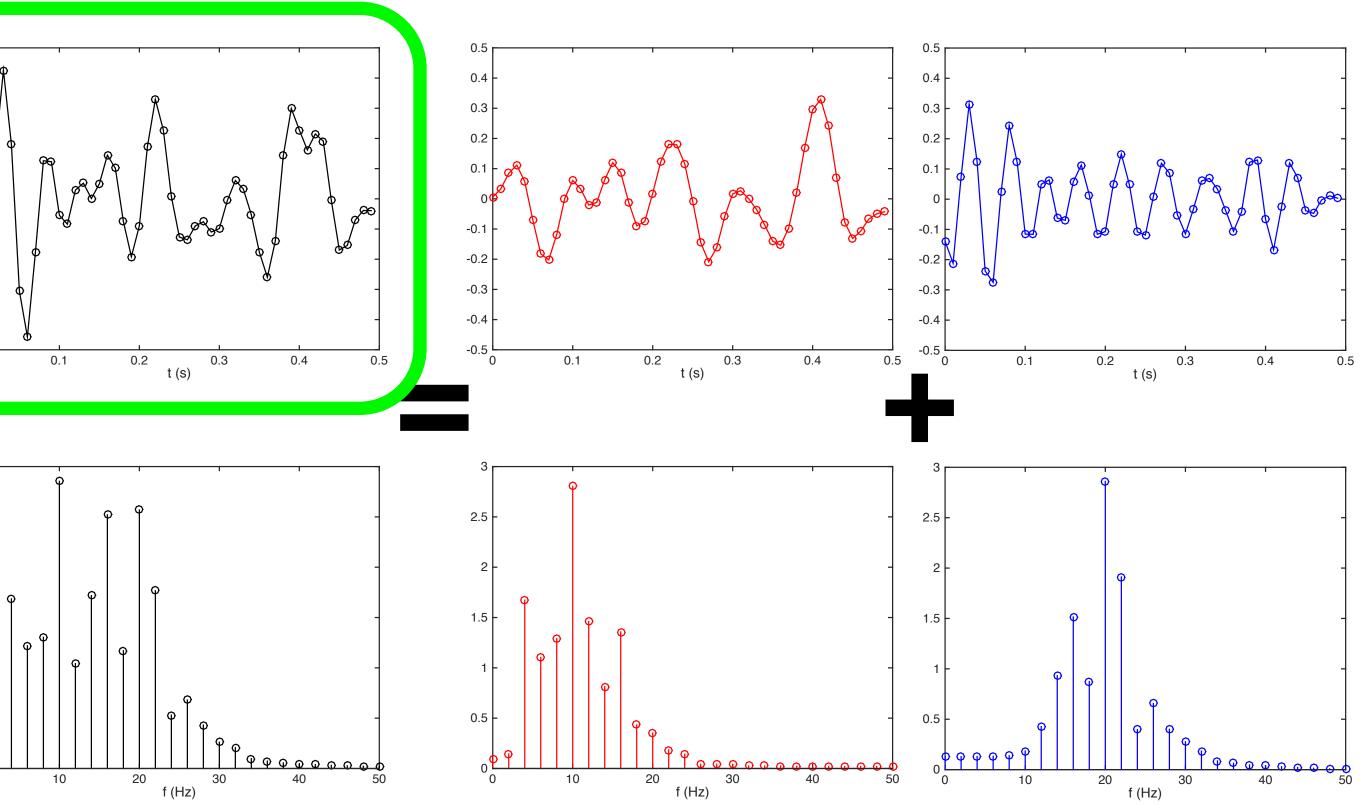


- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less

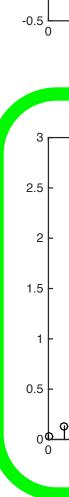


2.5

0.5



- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less



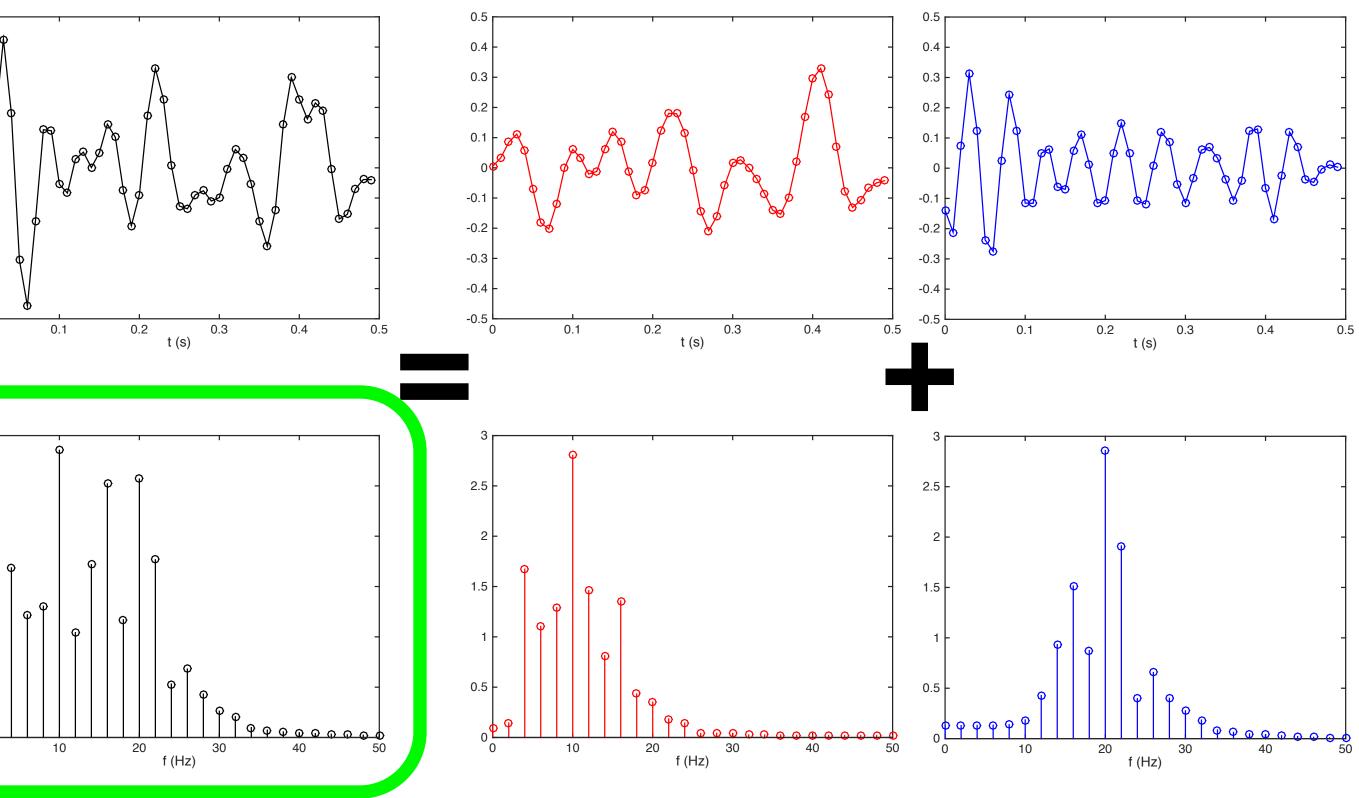
0.3

0.2

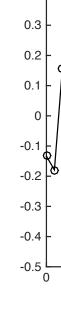
-0.2

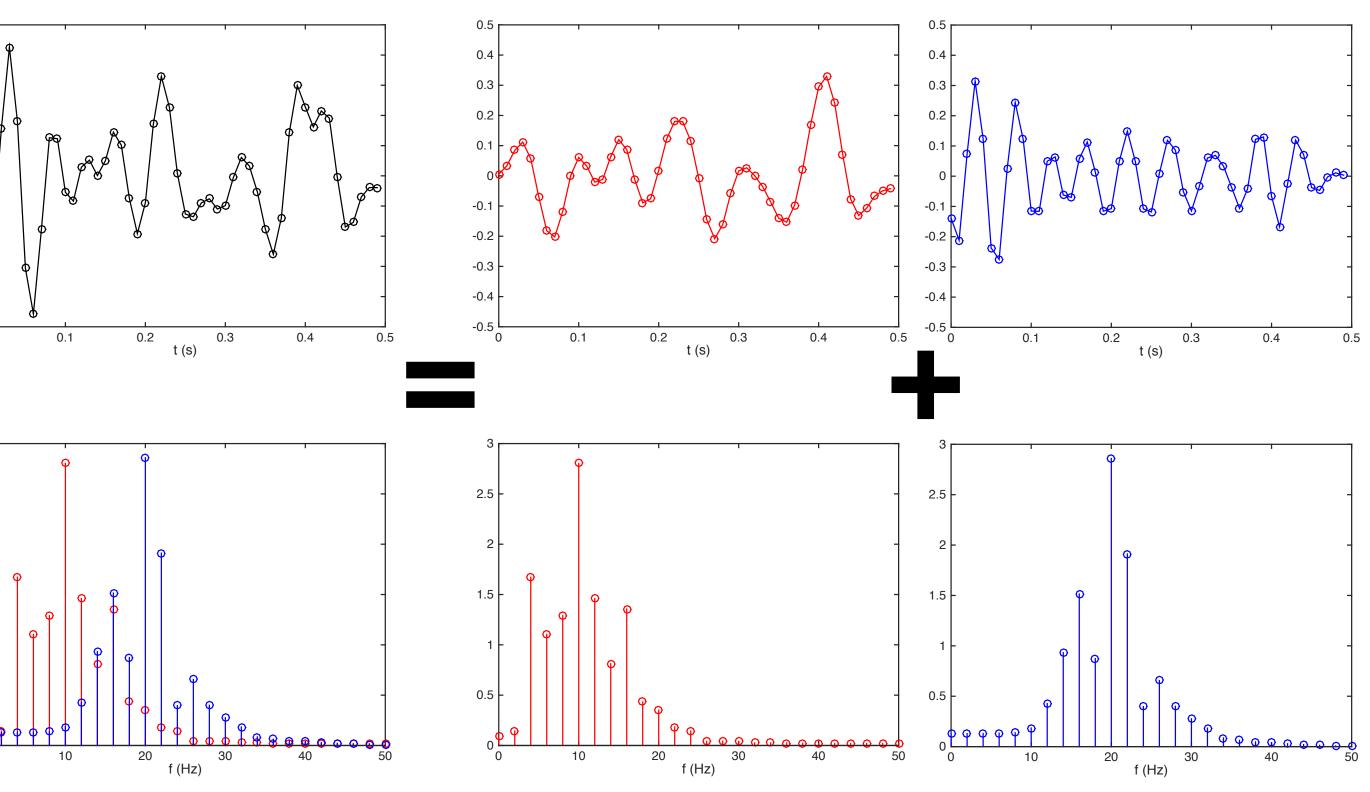
-0.3

-0.4

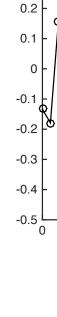


- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less



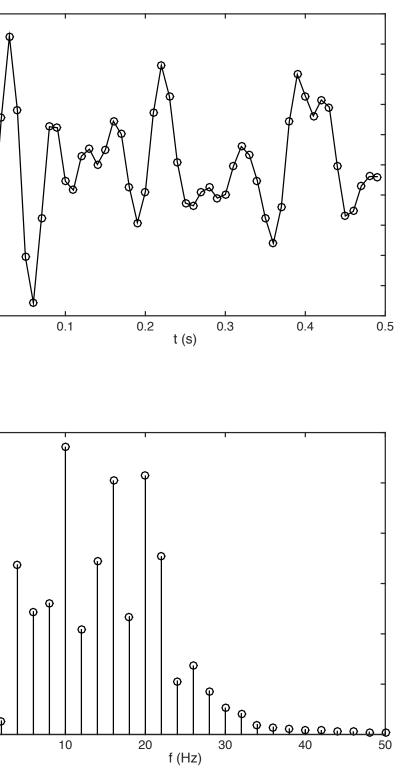


- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less

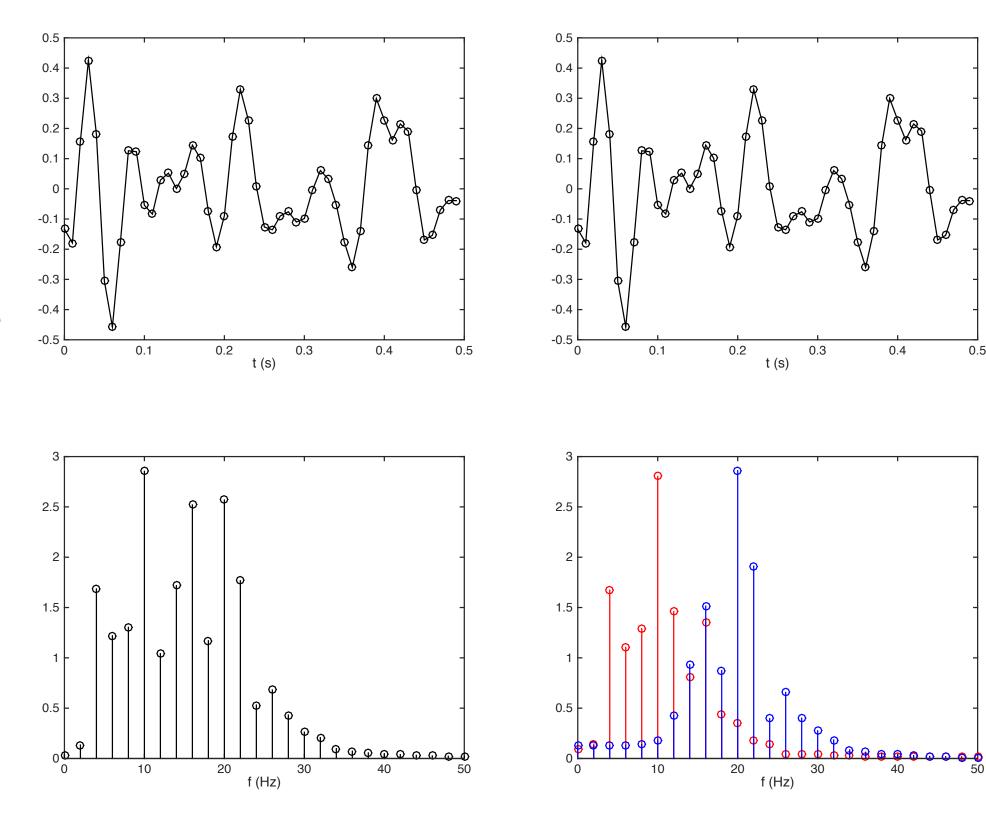


0.5

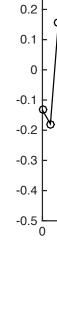
0.3



- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less

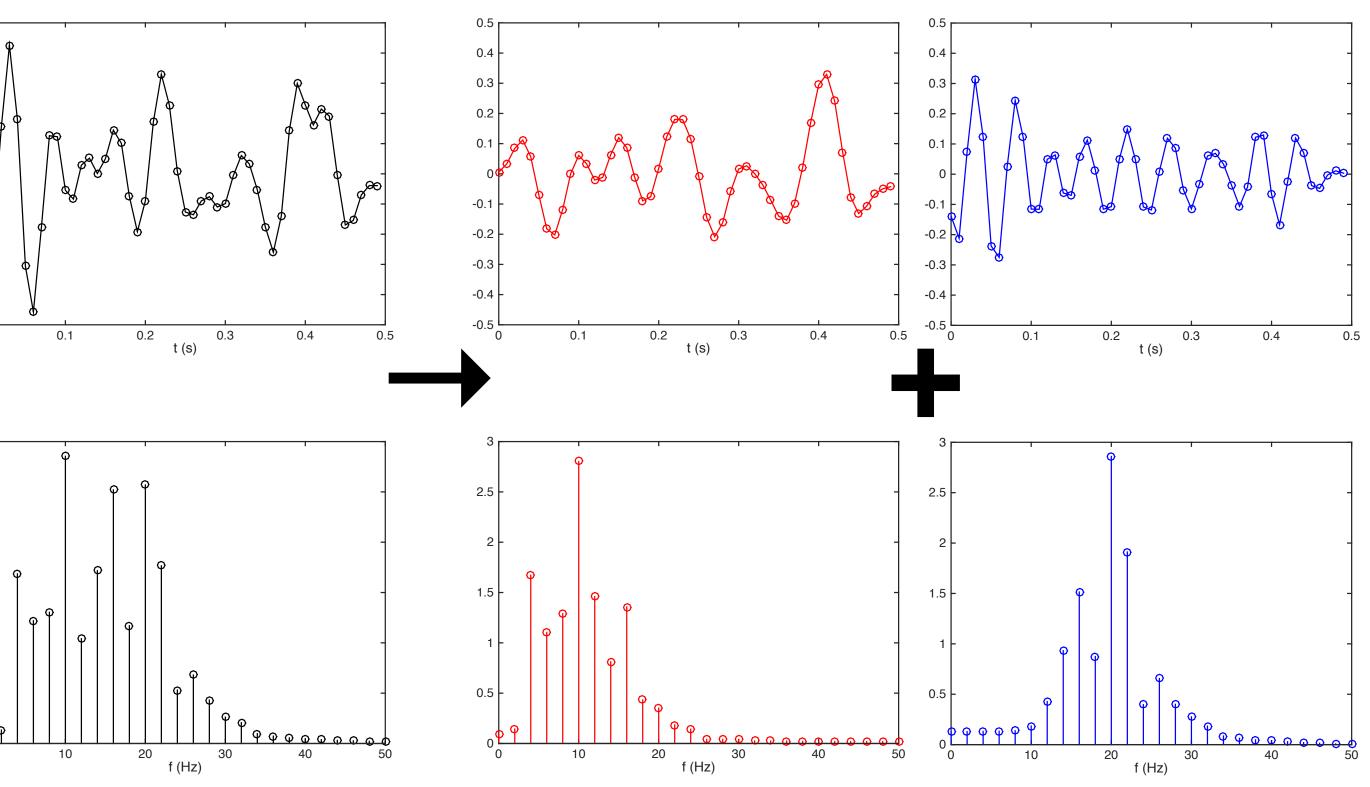


- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less

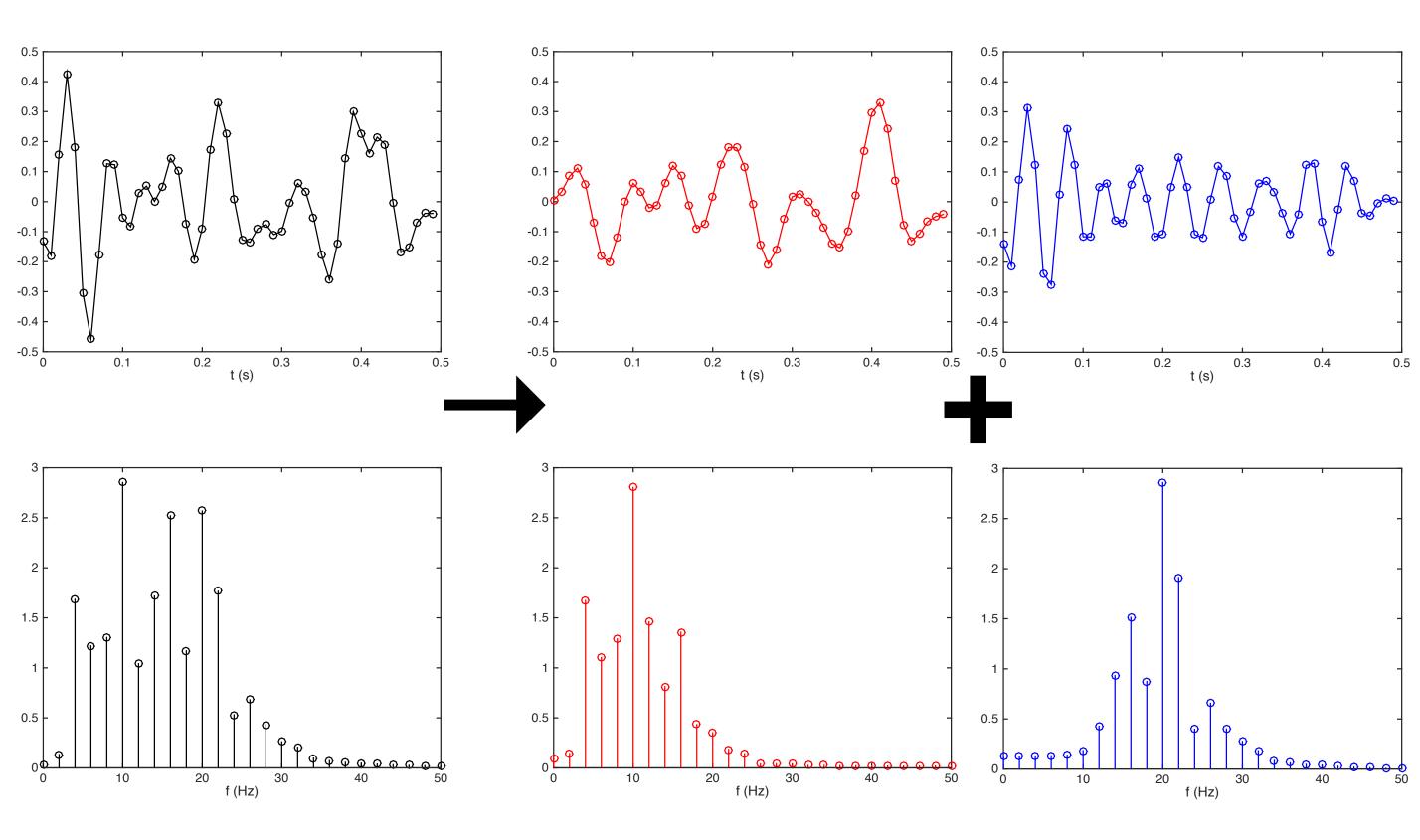


0.5

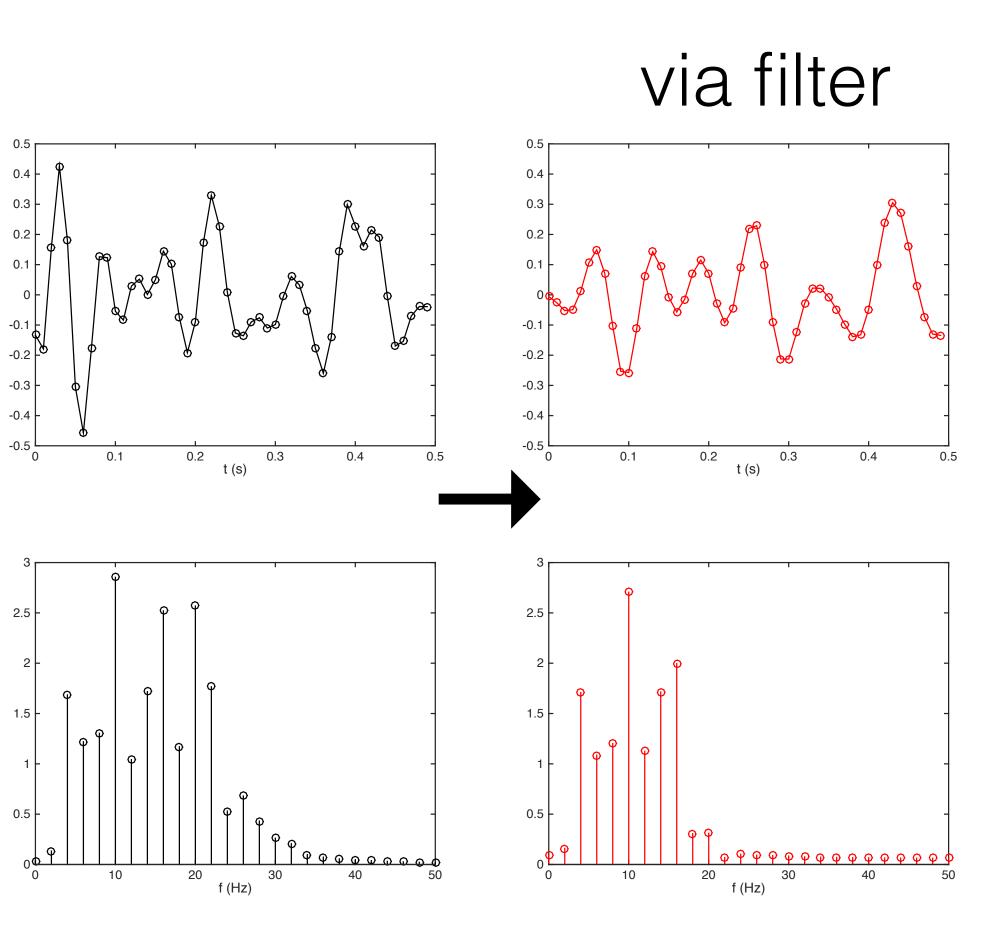
0.3



- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less
- Can *filter* measured (mixed) signal to "recover" underlying source signal



- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less
- Can *filter* measured (mixed) signal to "recover" underlying source signal

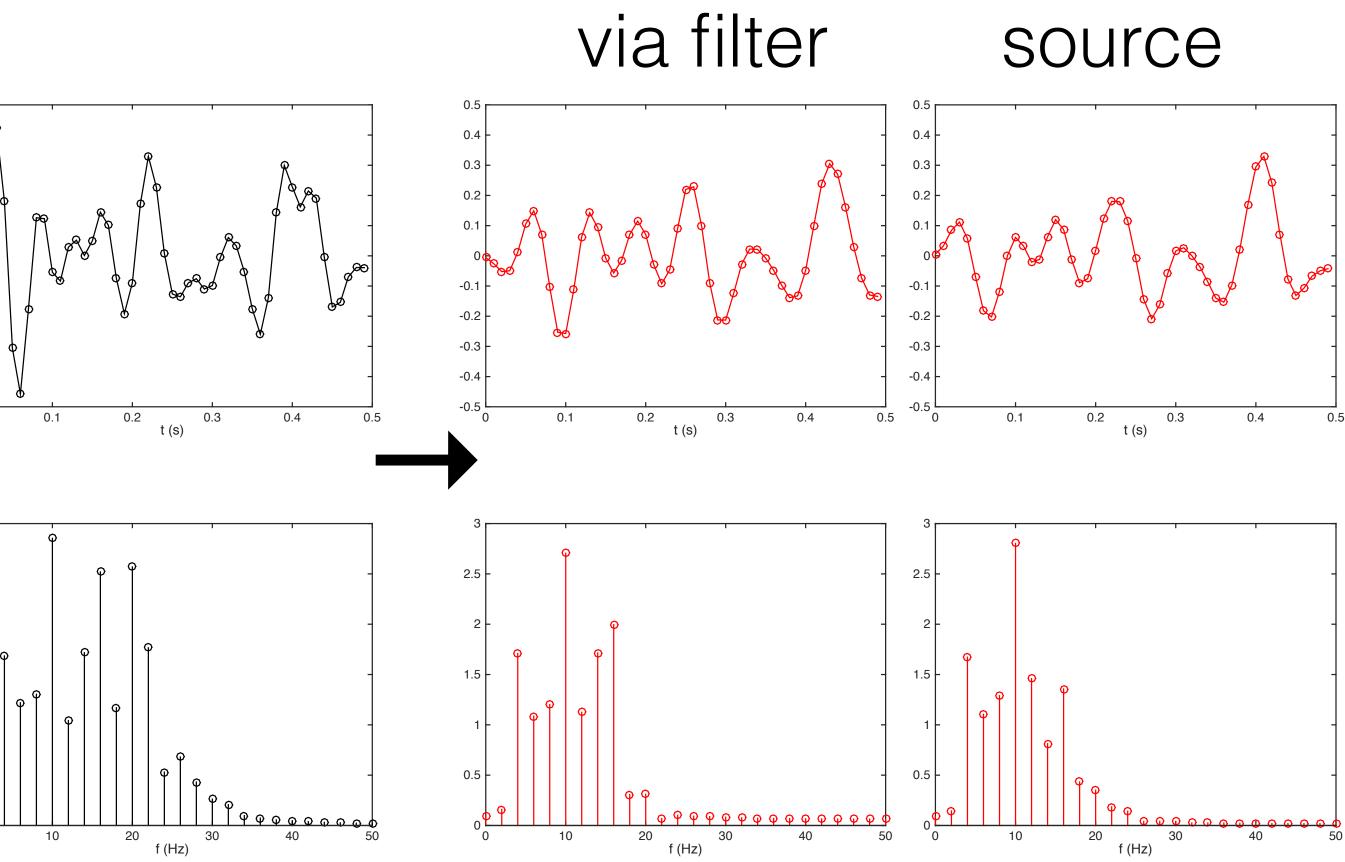


-0.2

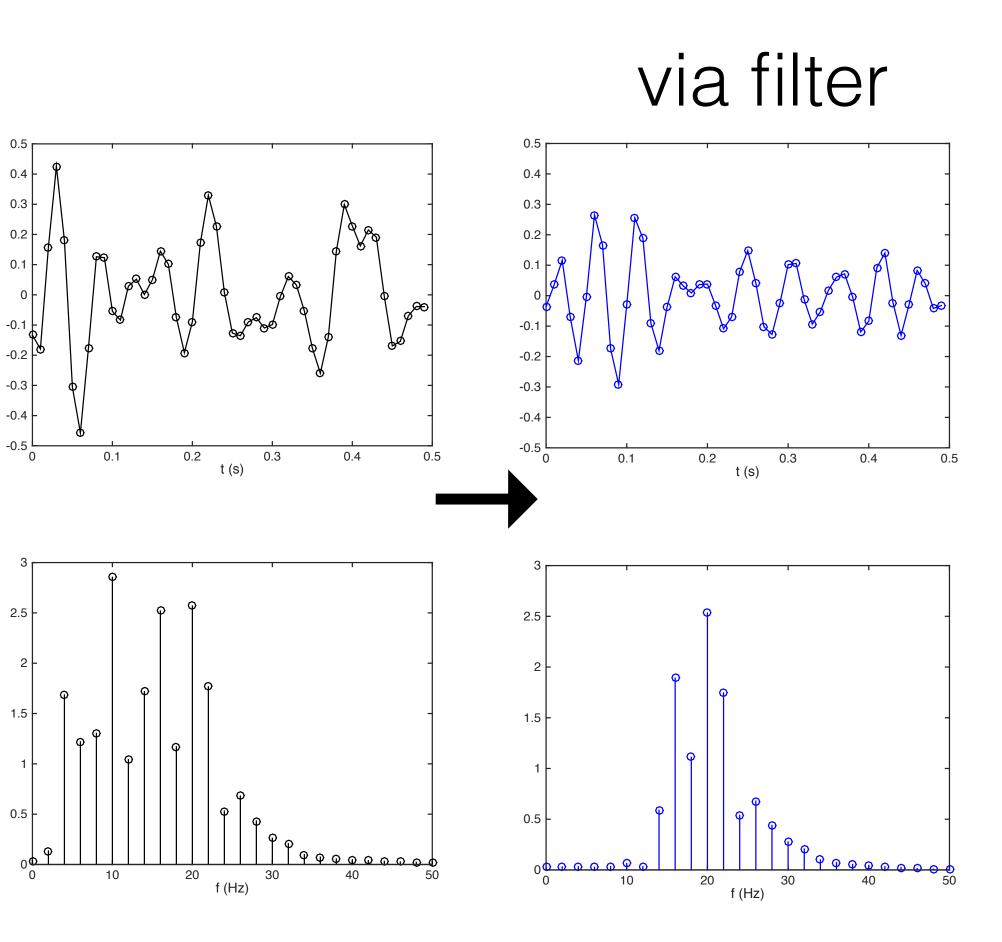
-0.3

0.5

- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less
- Can *filter* measured (mixed) signal to "recover" underlying source signal



- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less
- Can *filter* measured (mixed) signal to "recover" underlying source signal

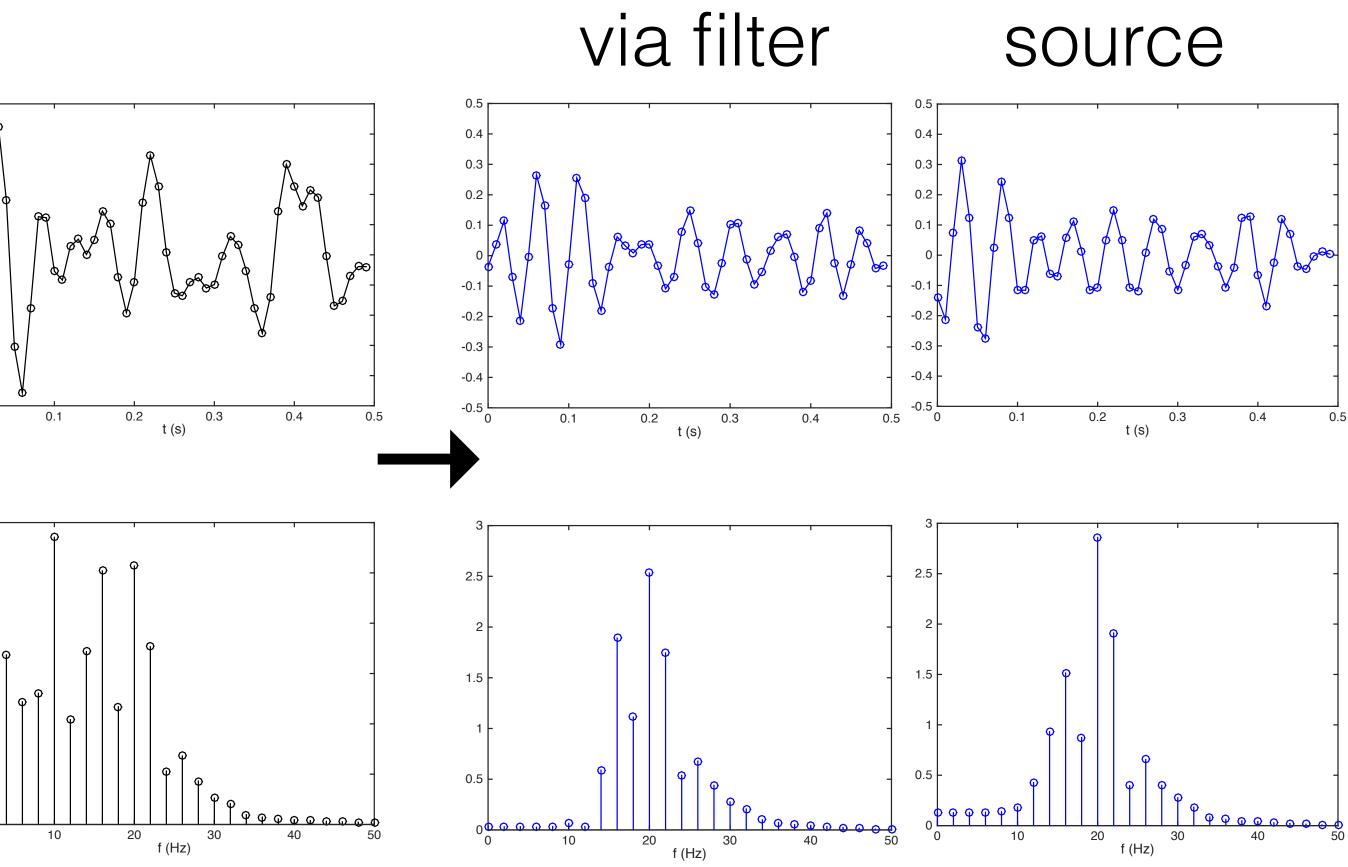


0.3

-0.2

-0.3

- Measured Signals made up of several (many?) sources
- All overlap in time
- But overlap in frequency may be much less
- Can *filter* measured (mixed) signal to "recover" underlying source signal



- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

- Filters: What They Do, and How They Do It
- Grab Bag:

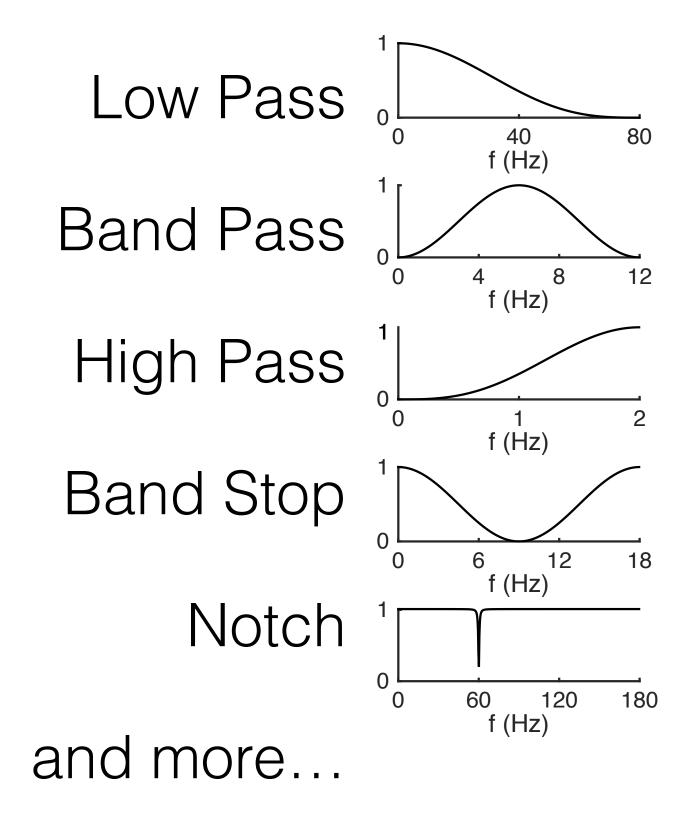
Outline

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

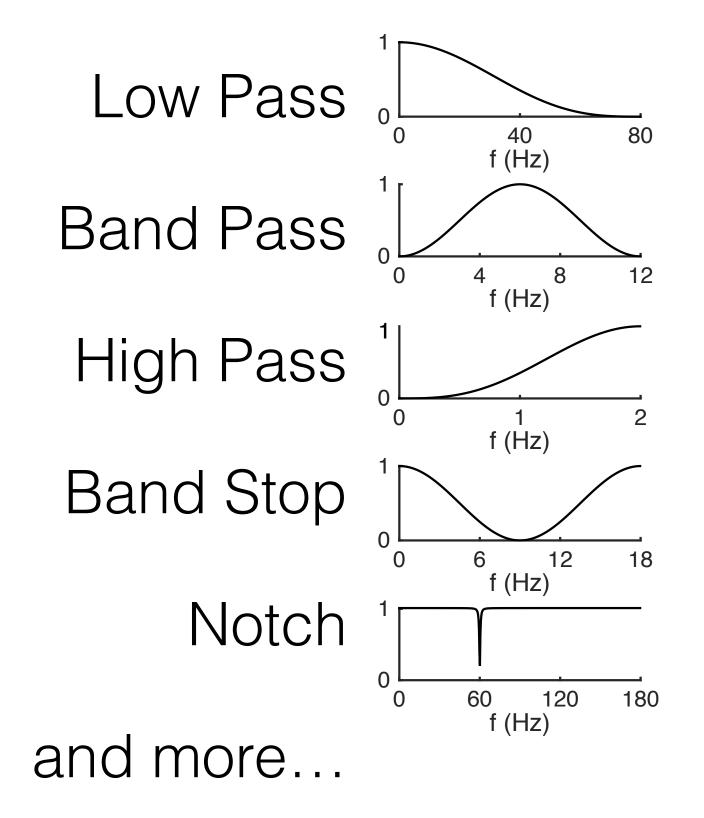
Filters: Frequency Selectivity

• Frequency Selective Filters



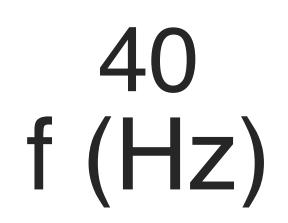
Filters: Frequency Selectivity

• Frequency Selective Filters

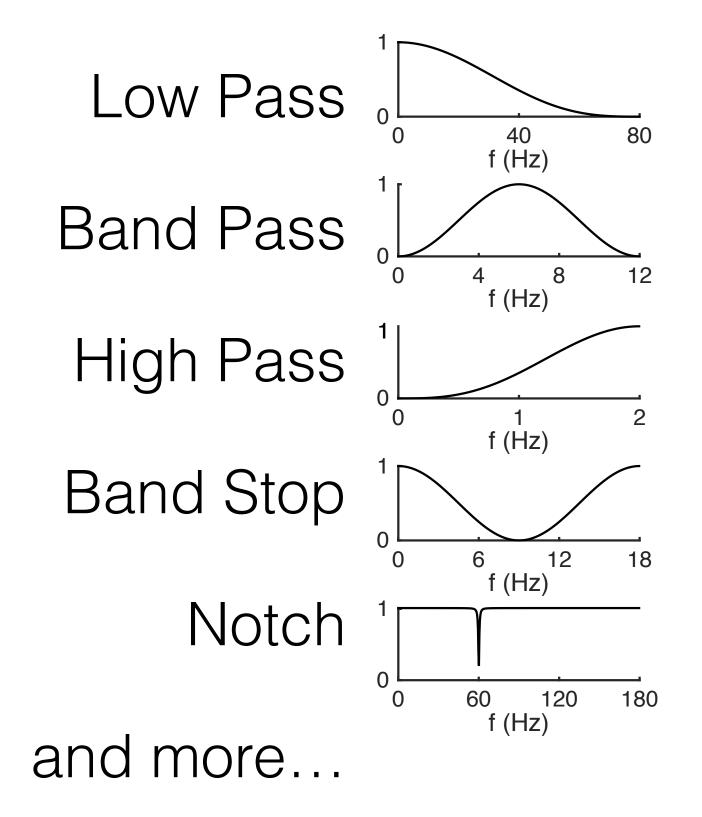


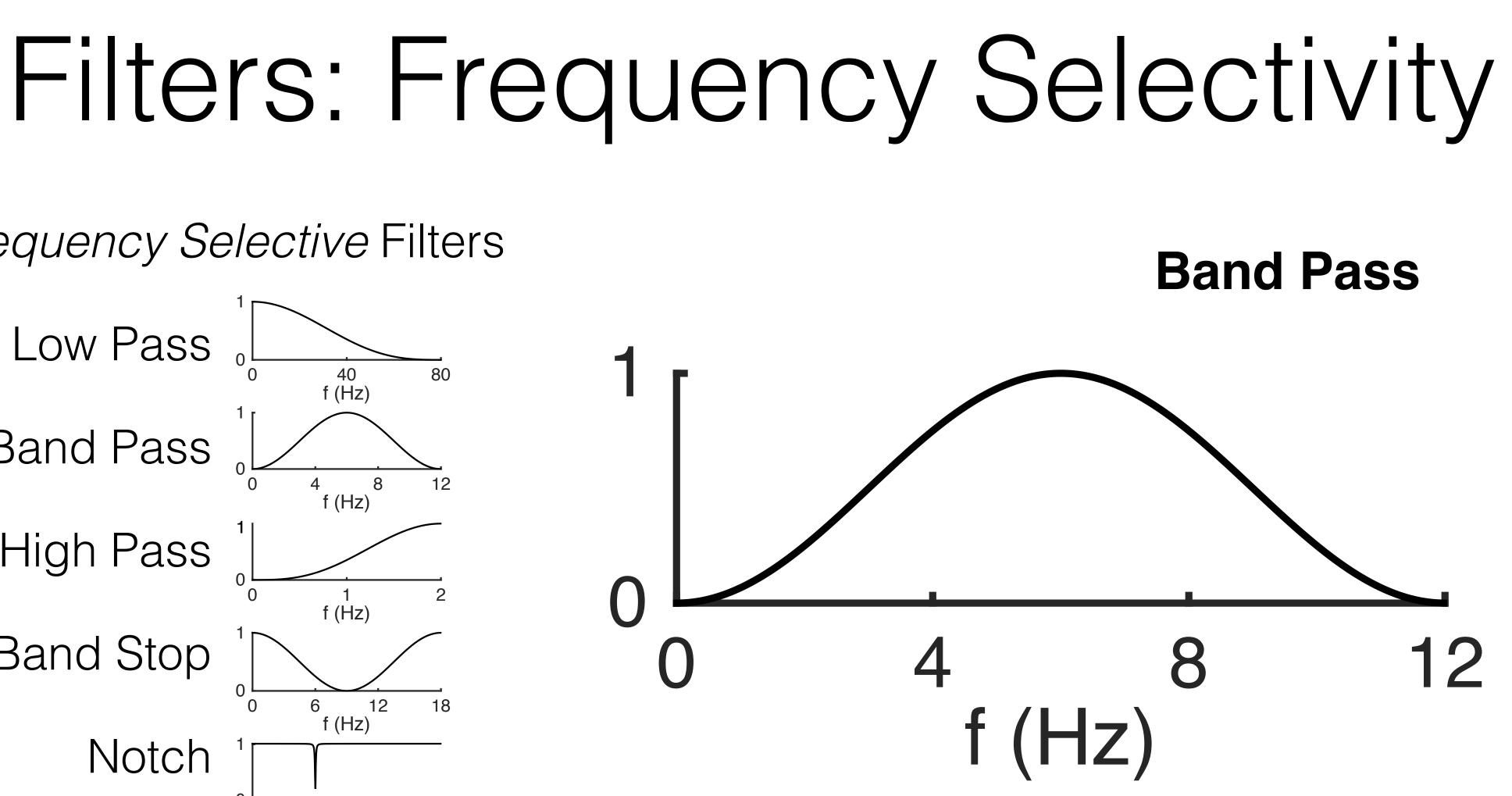
Low Pass

80



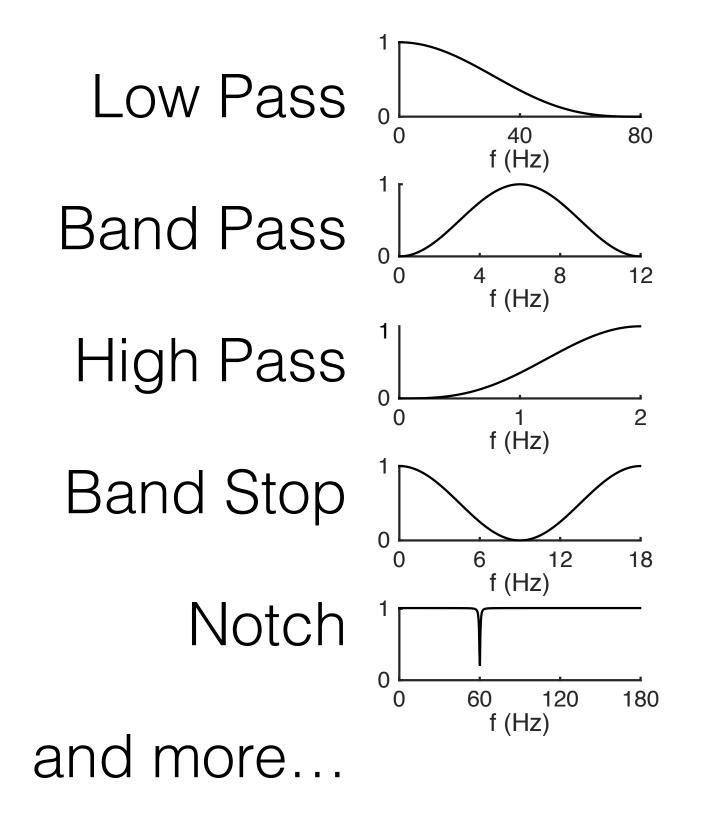
• Frequency Selective Filters



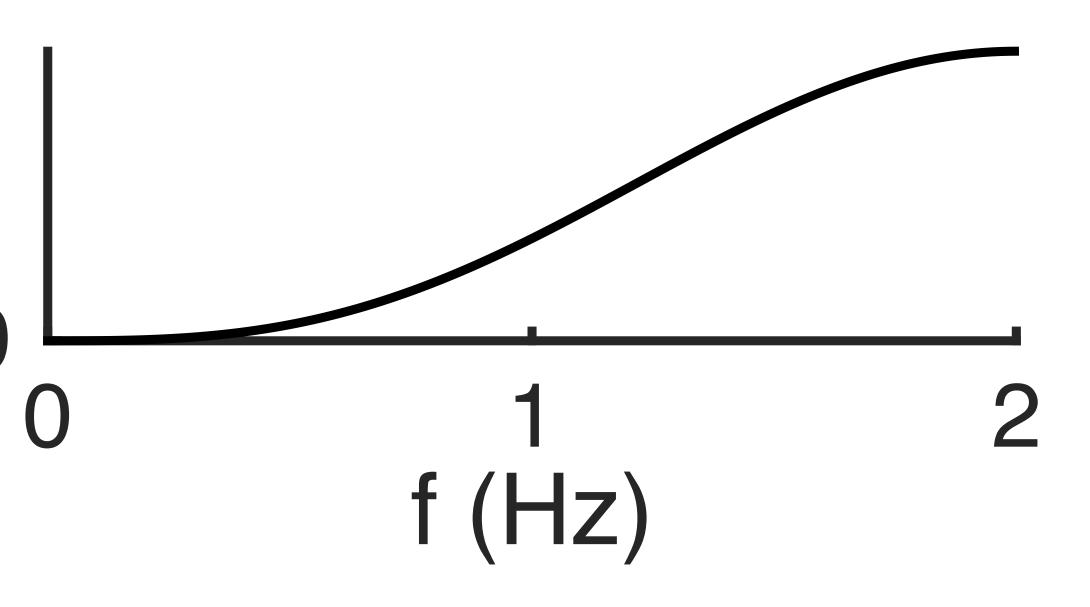


Filters: Frequency Selectivity

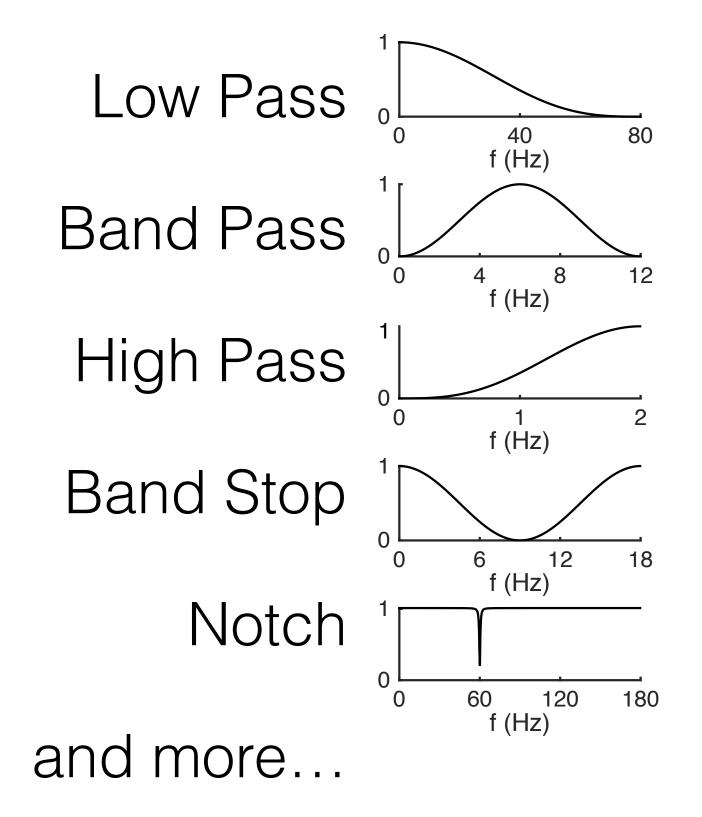
• Frequency Selective Filters

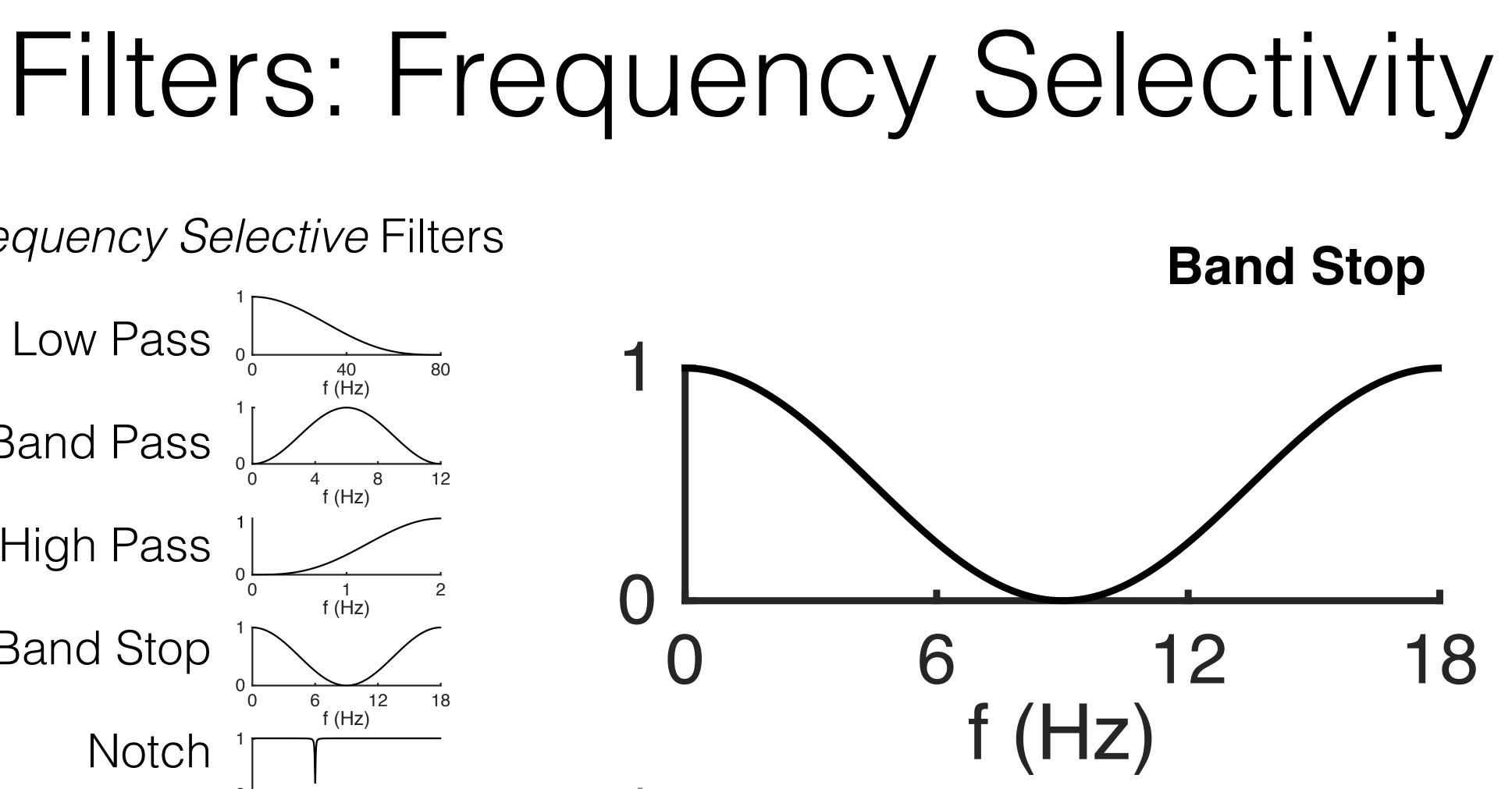


High Pass



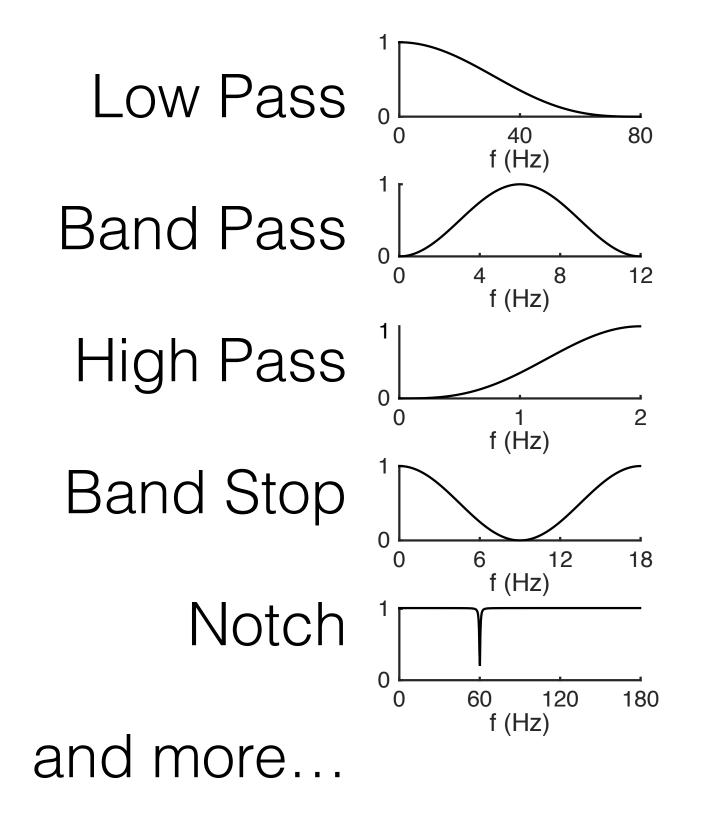
• Frequency Selective Filters





Filters: Frequency Selectivity

• Frequency Selective Filters

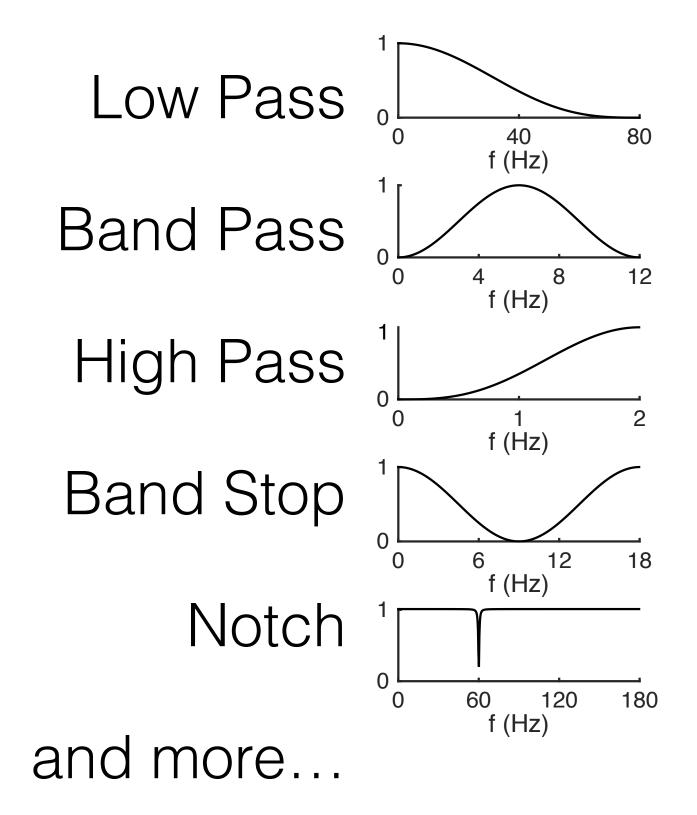


Notch

60 120 180 f (Hz)

Filters: Frequency Selectivity

• Frequency Selective Filters

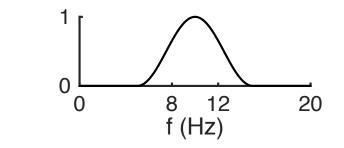


Filters: How Selective?

• How sharp a transition?

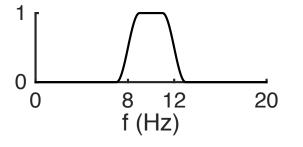








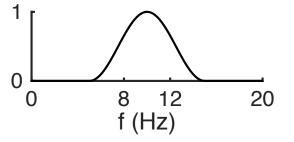
Sharp Transition



8 12 f (Hz)

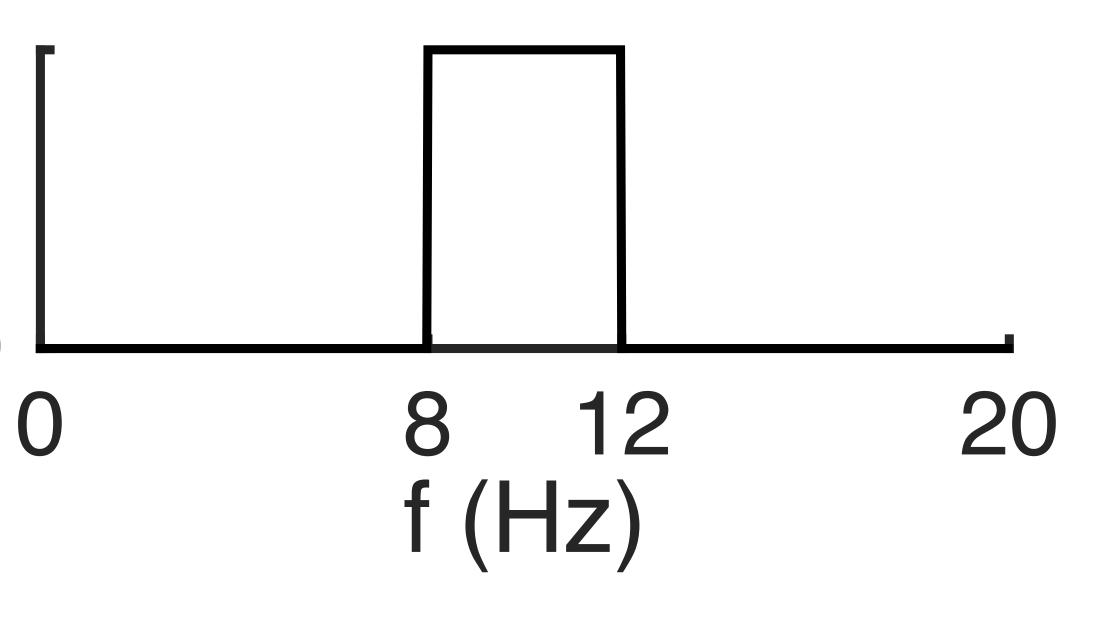
20

Soft Transition



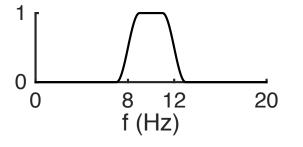


"Ideal" Filter



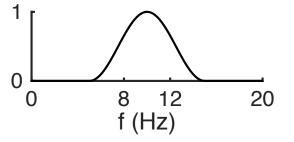


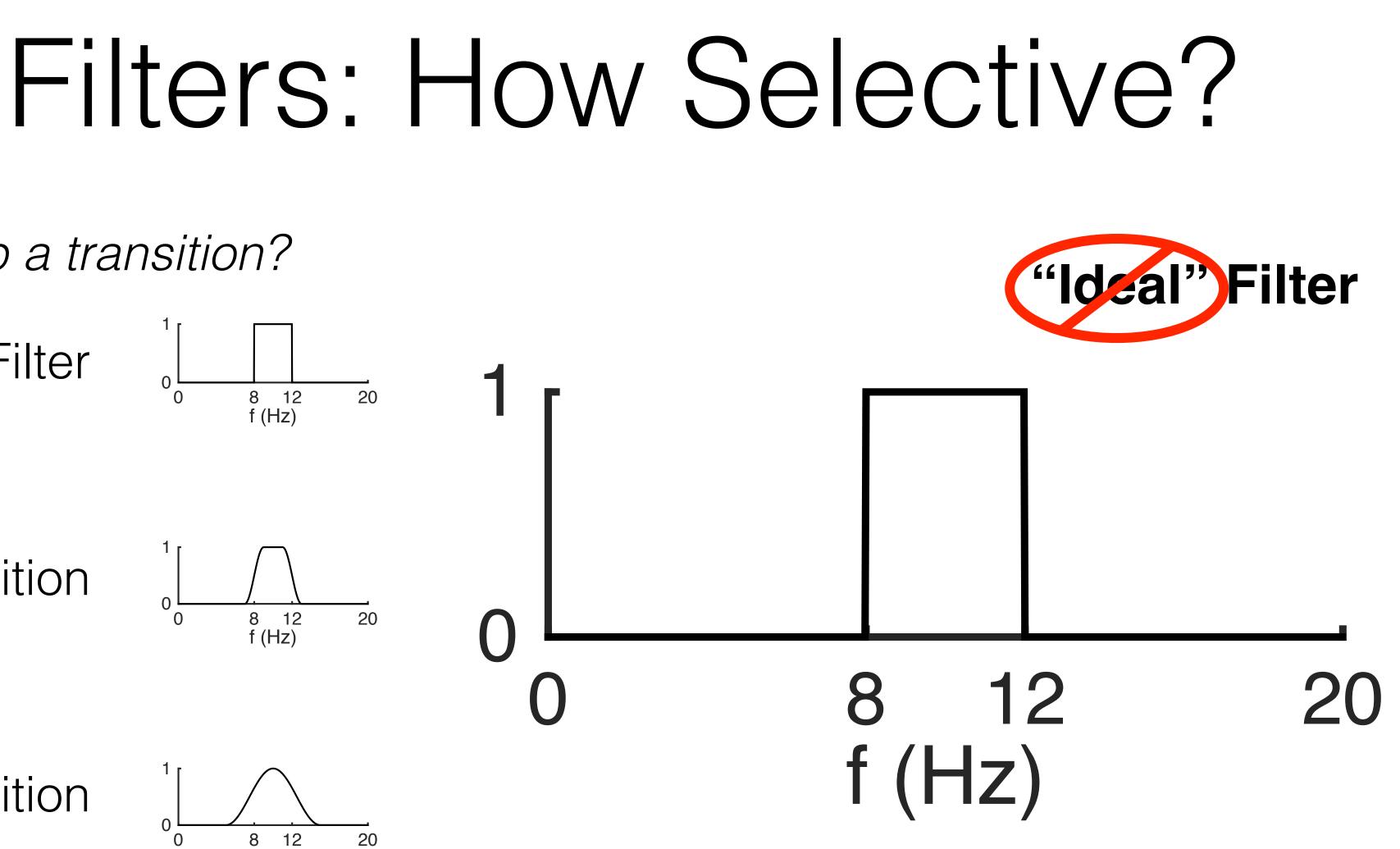
Sharp Transition



8 12 f (Hz)

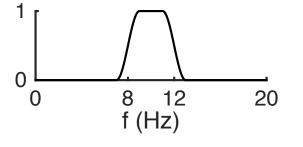
20



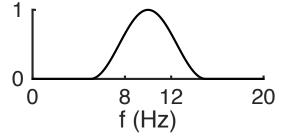


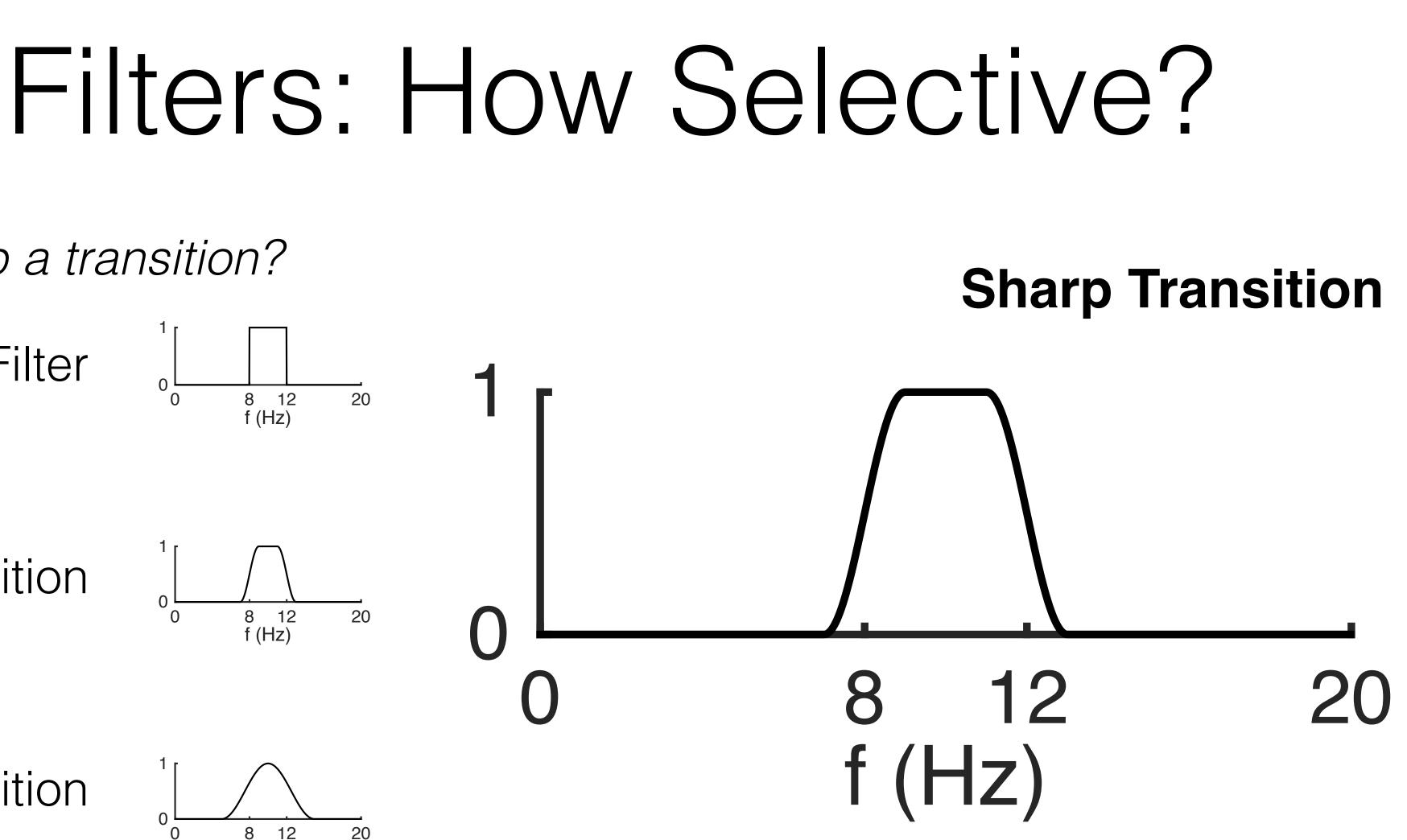


Sharp Transition



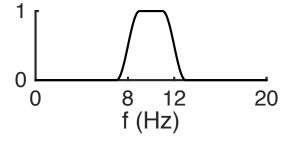
20



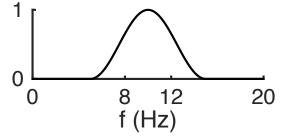


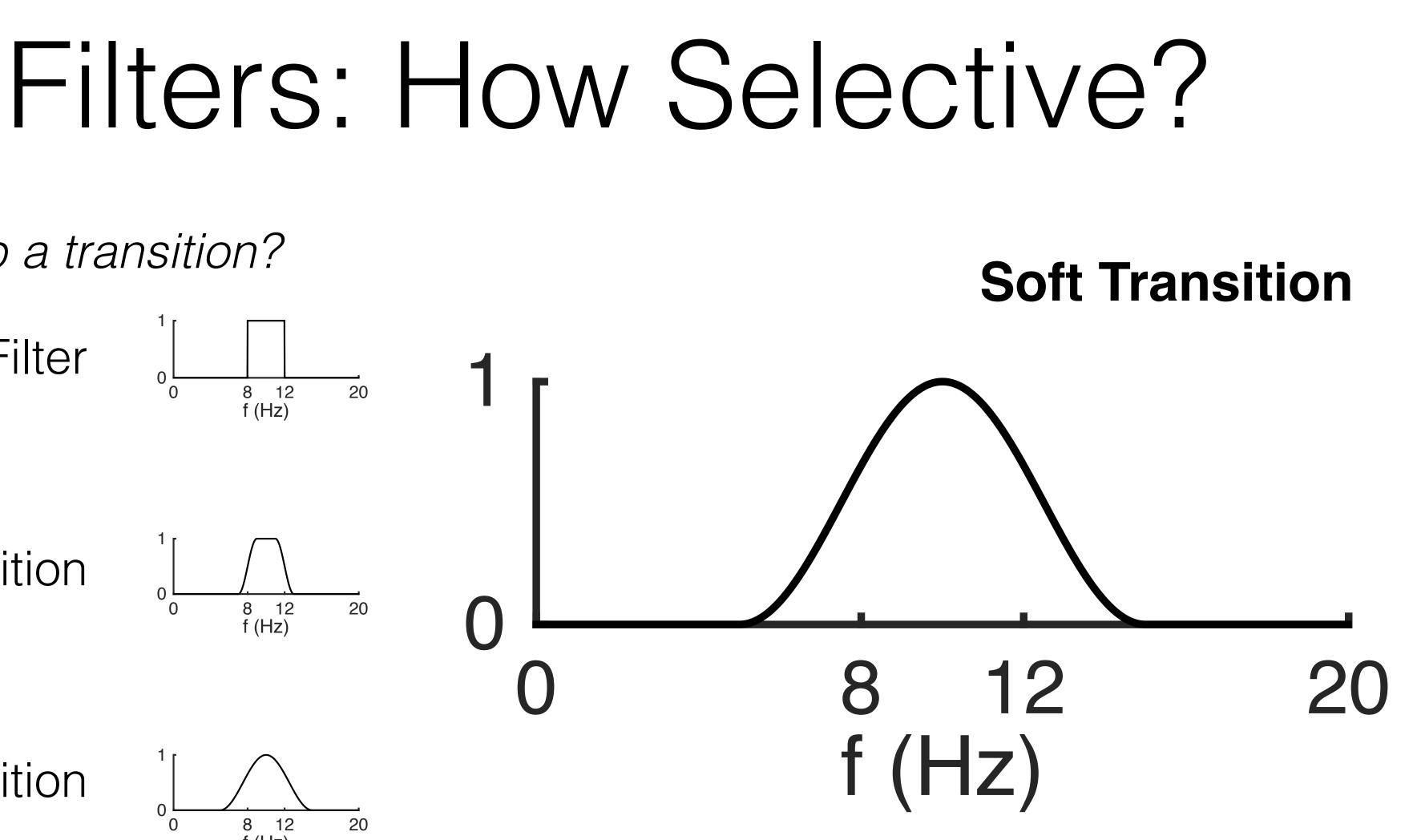


Sharp Transition



20



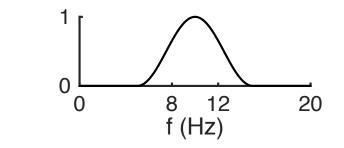


Filters: How Selective?

• How sharp a transition?







Output of Filter:

- Linear Combination of Input Signal and **Earlier Versions** of the Input Signal
- Linear Combination of Input Signal and *Earlier Versions* of the **Output** Signal
- Linear Combination of Input Signal and *Earlier Versions* of both the Input and Output Signals

Filters: How Do They Work?

$$y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$$

$$y[t] = \frac{1}{10} x[t] - \frac{9}{10} y[t - \Delta t]$$

$$y[t] = x[t] - x[t - \Delta t] + x[t - 2\Delta t] + \frac{99}{100}y[t - \Delta t] - \left(\frac{99}{100}\right)^2 y[t - \Delta t] -$$

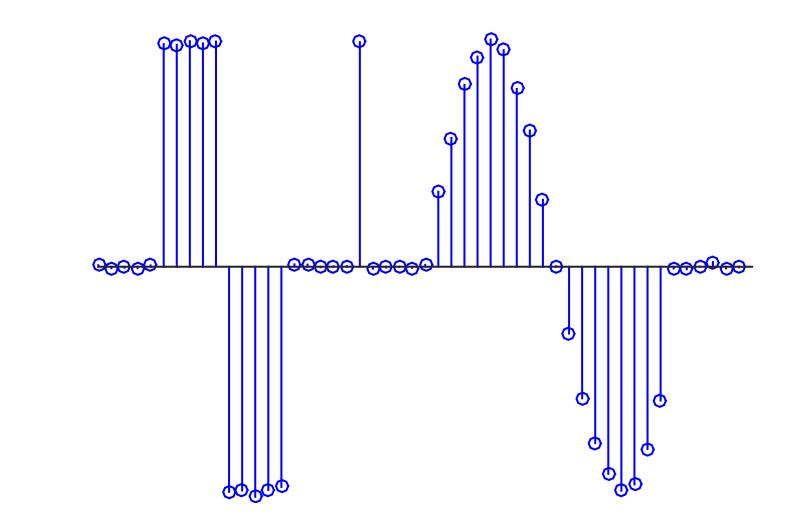


$y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

What to Expect:

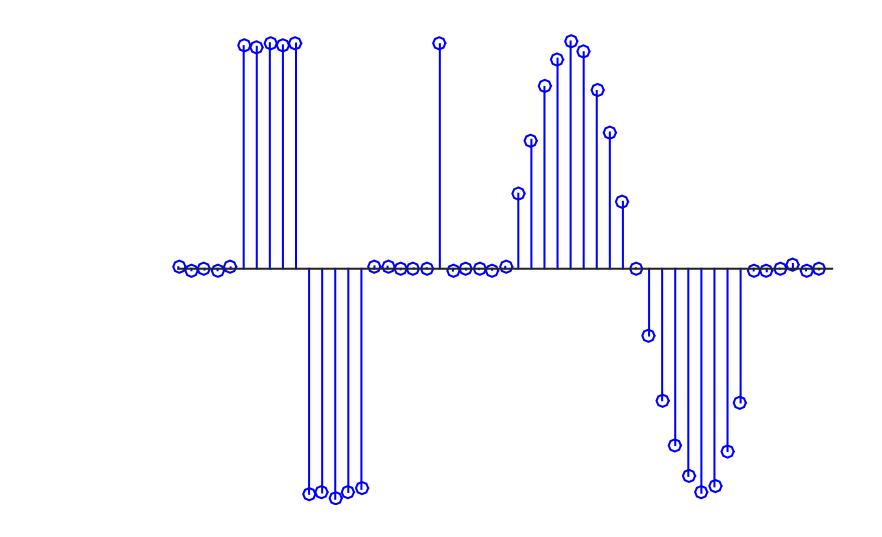
- Smooth over rough patches
- Soften sudden changes
- Leave slowly varying signals largely unchanged
- Low Pass Filter?

$y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

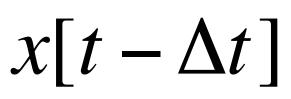


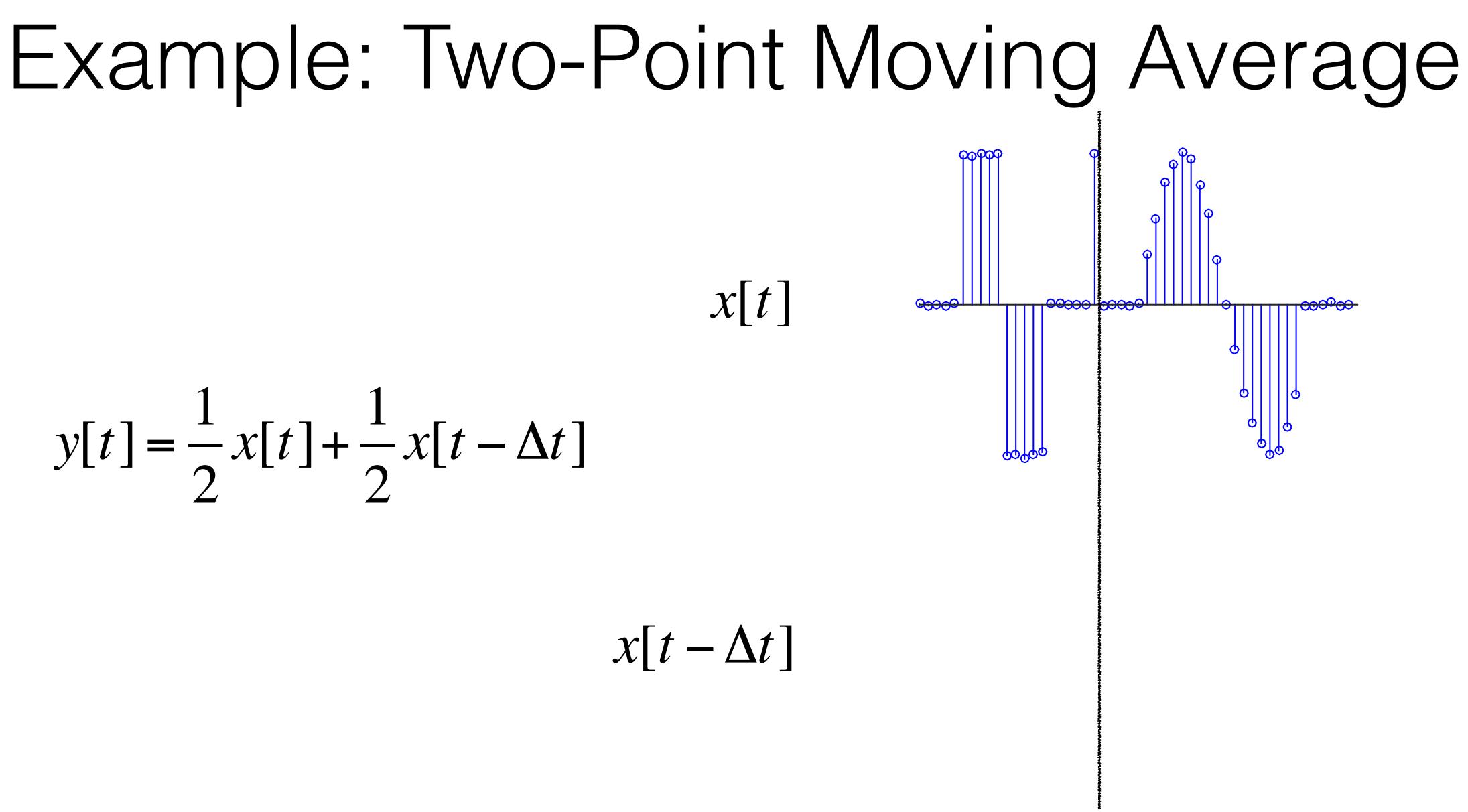
x[t]

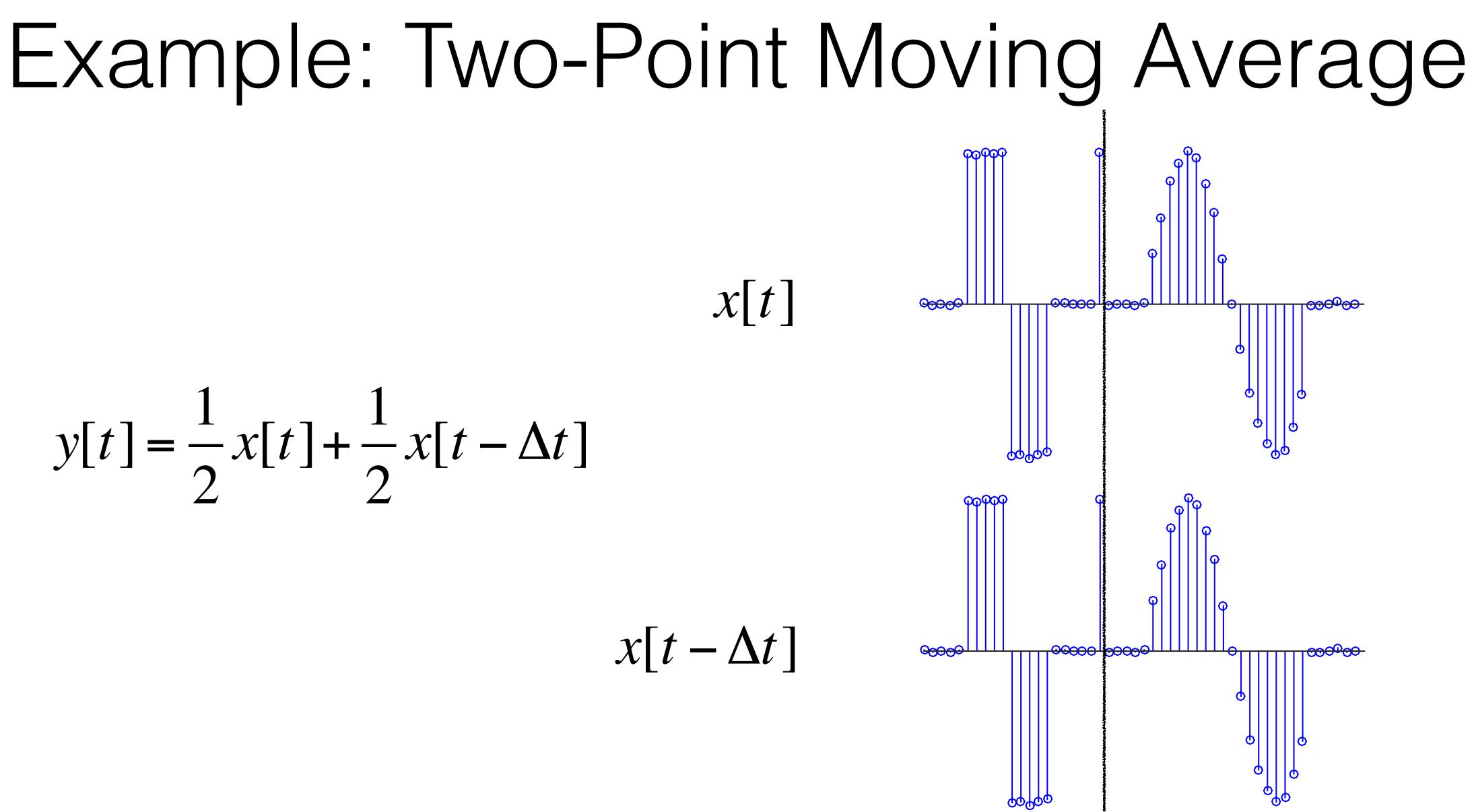
$y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

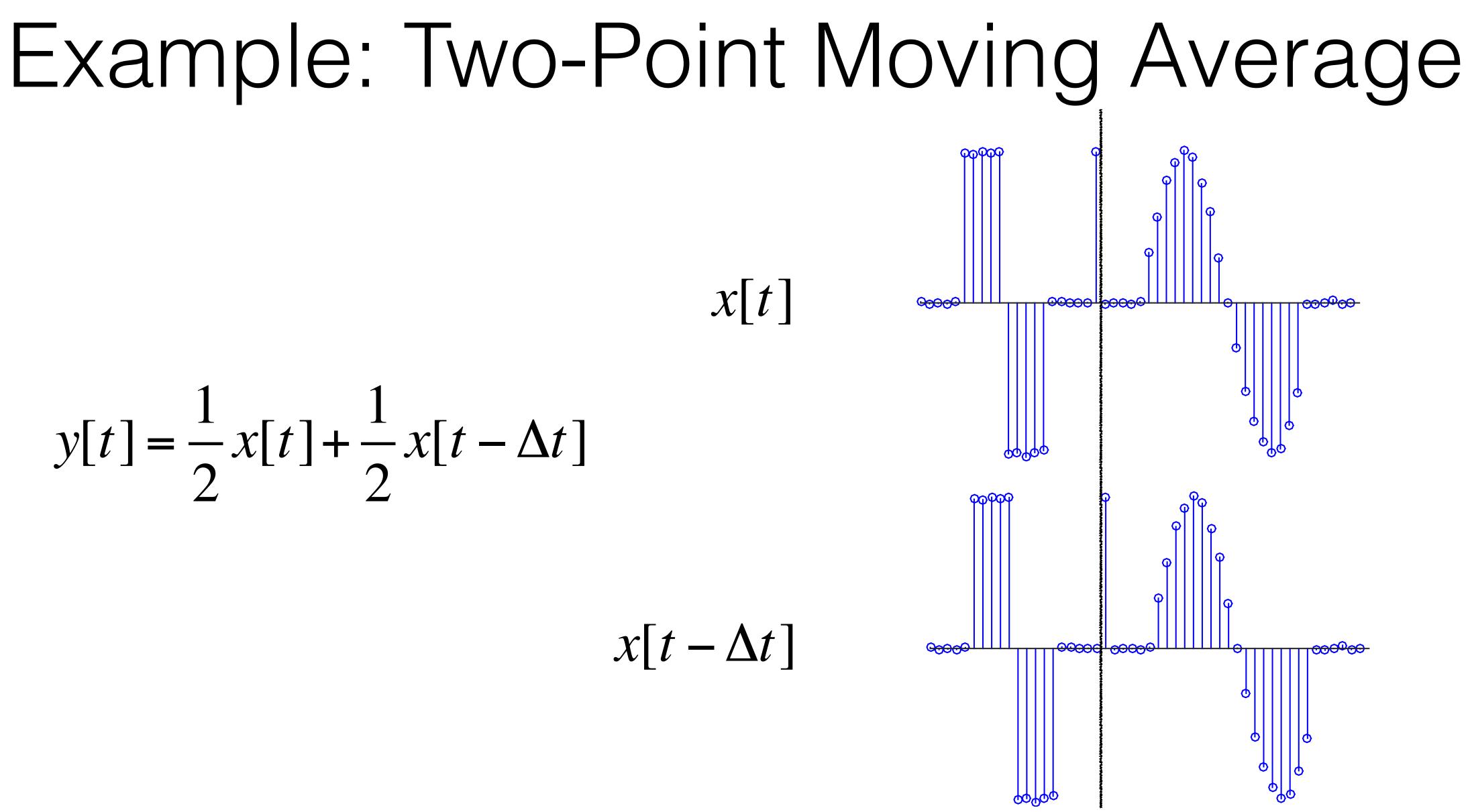


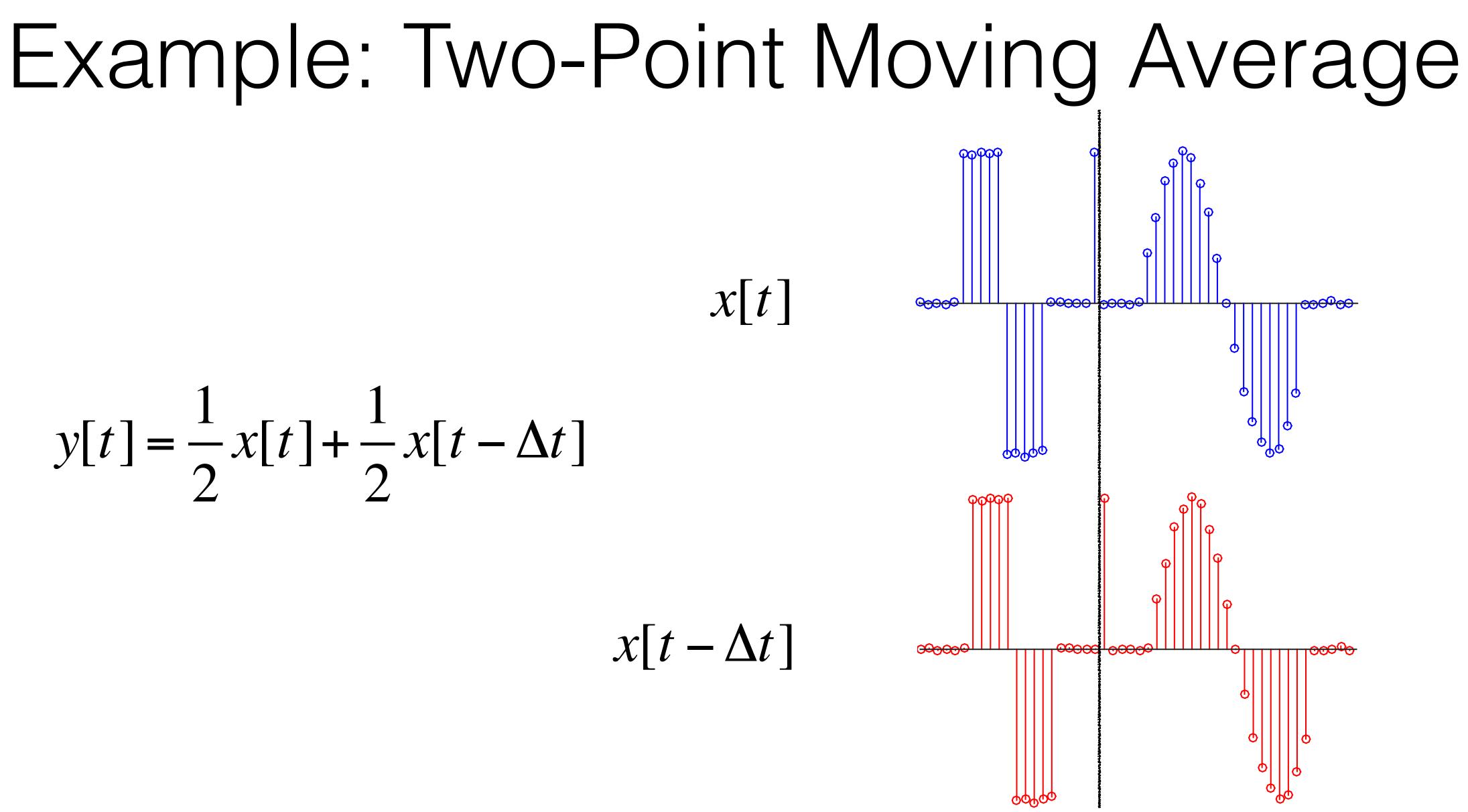
x[t]

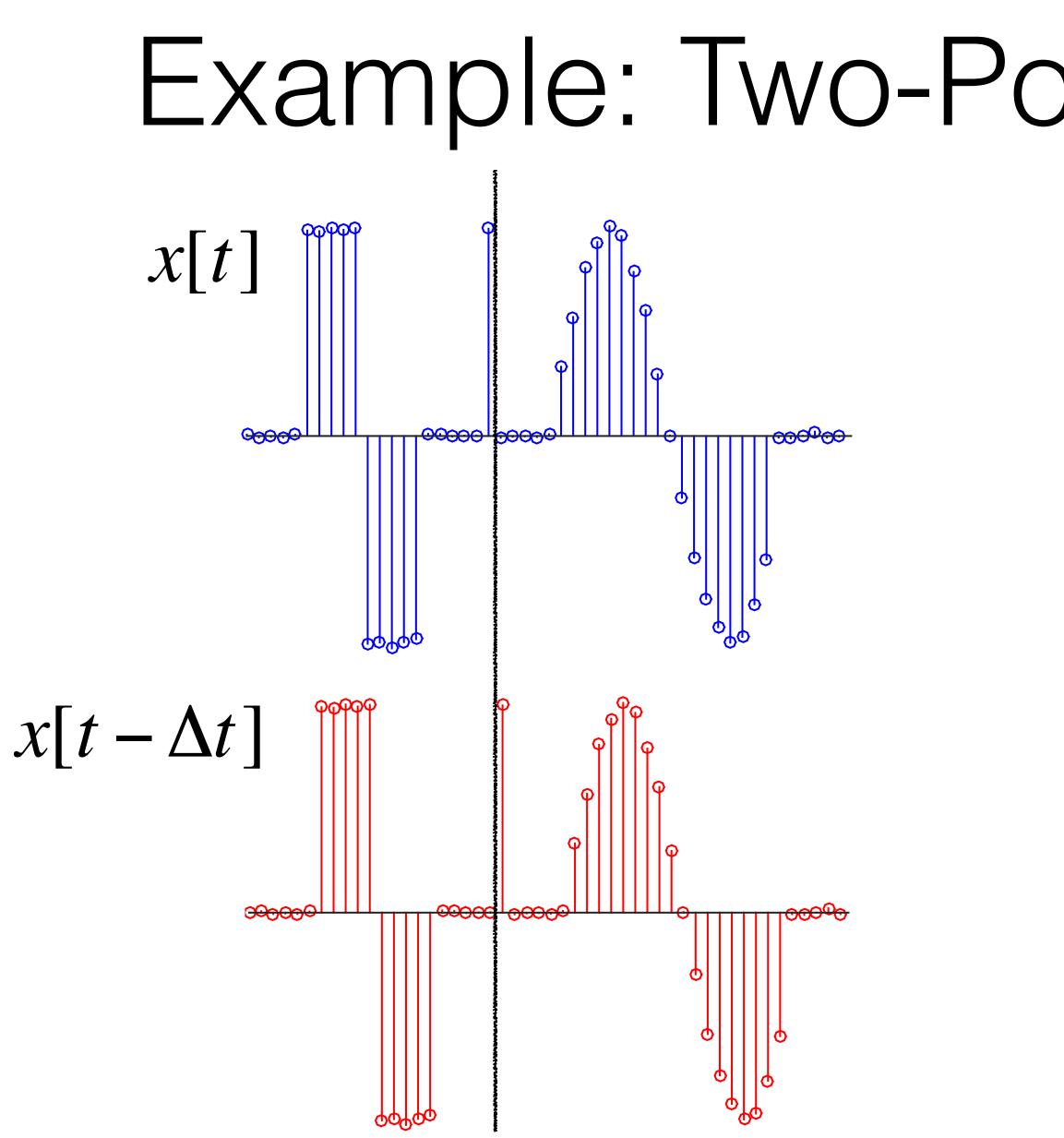


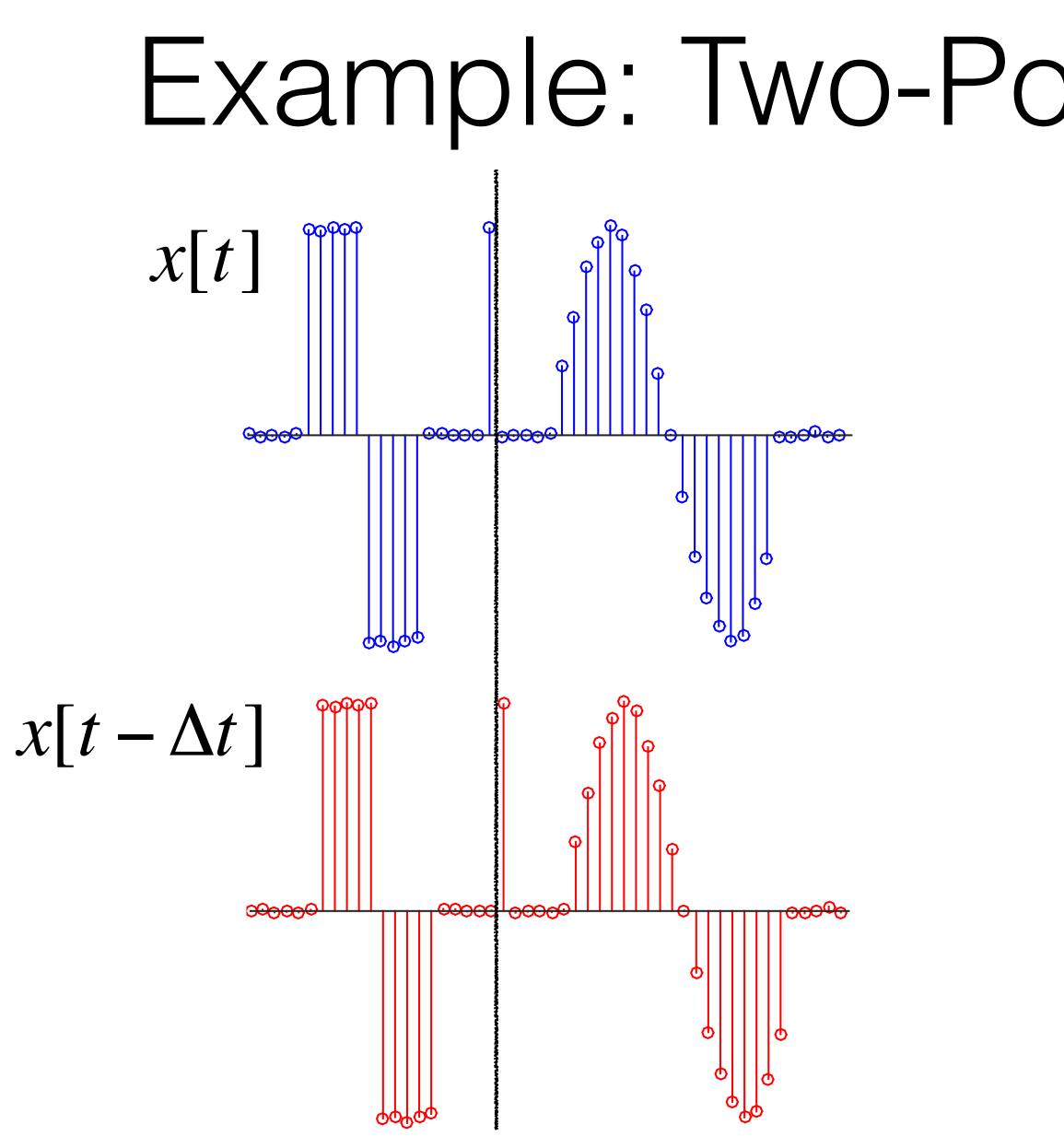


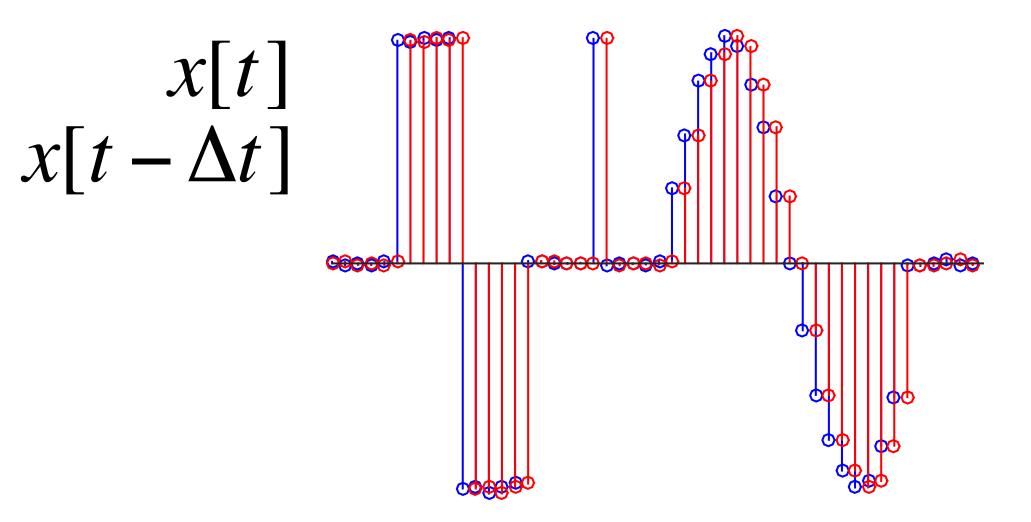


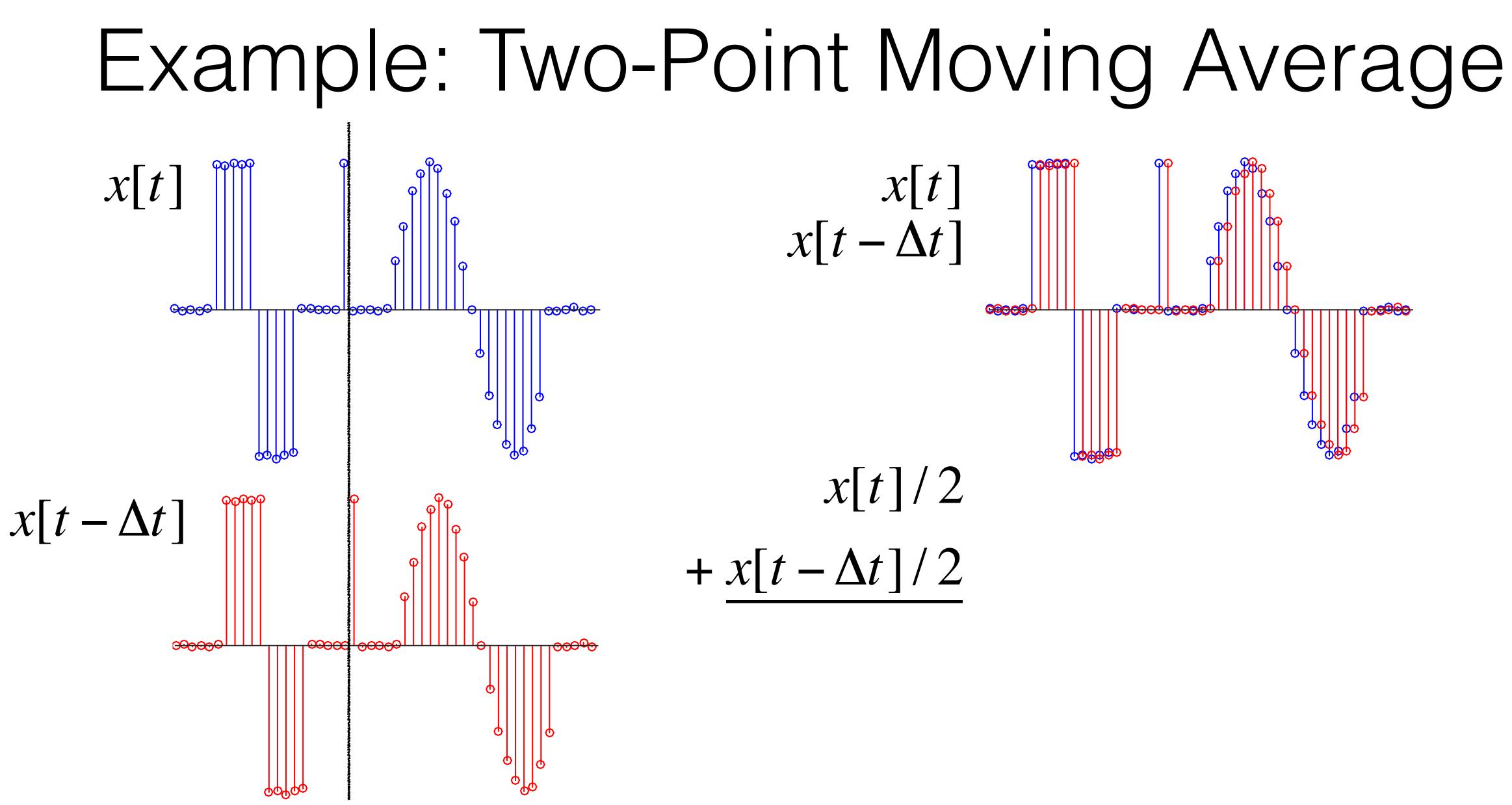


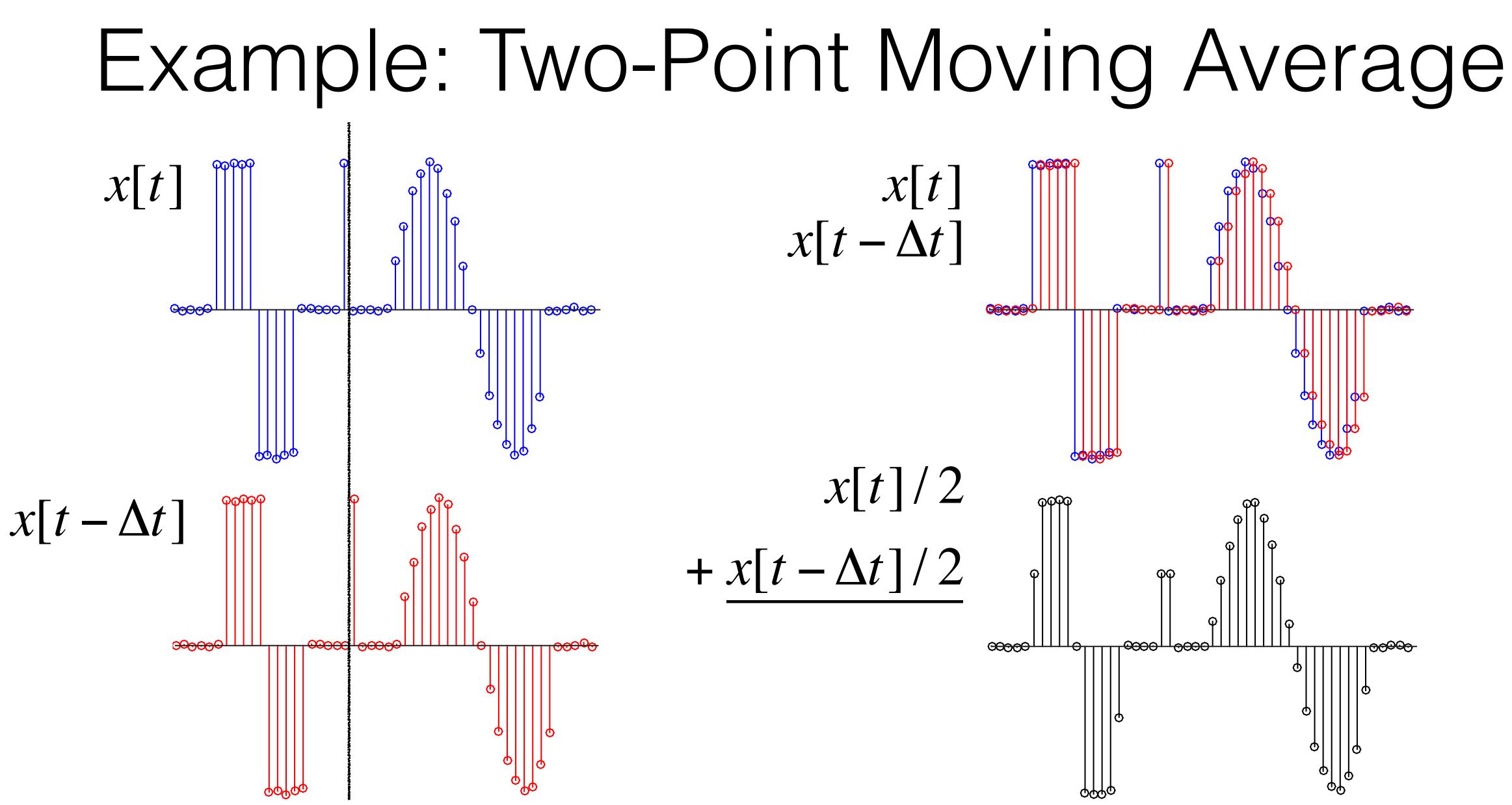


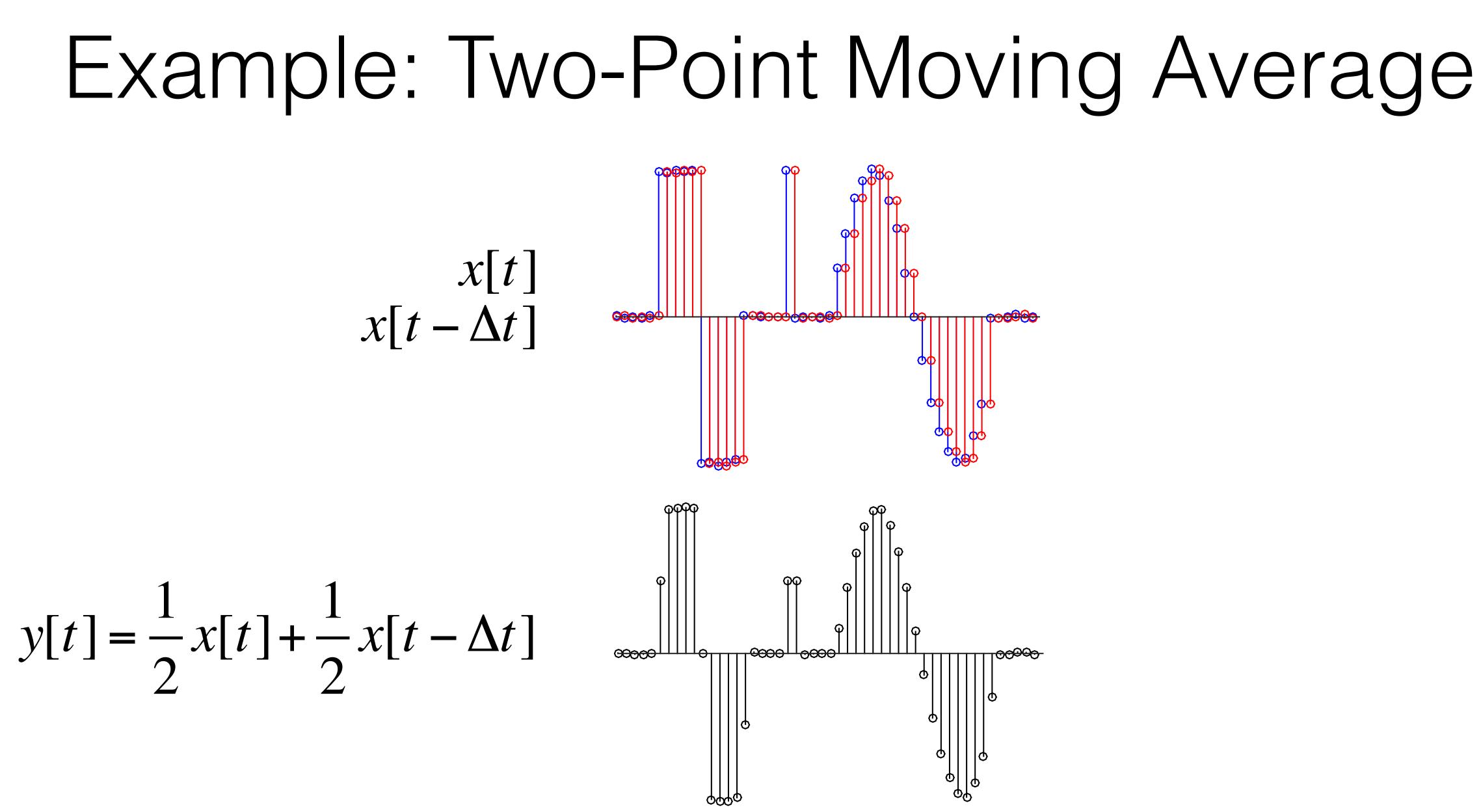


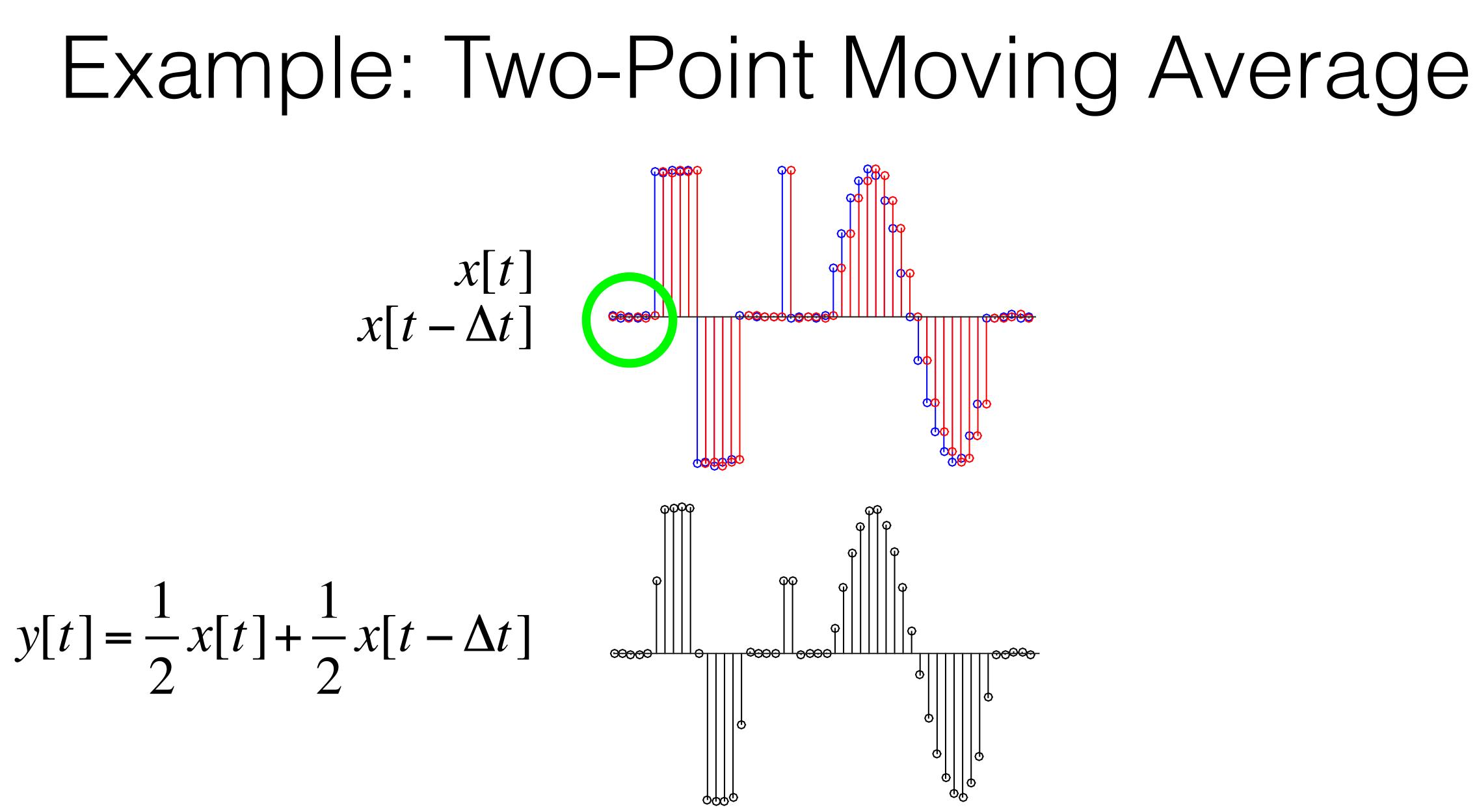


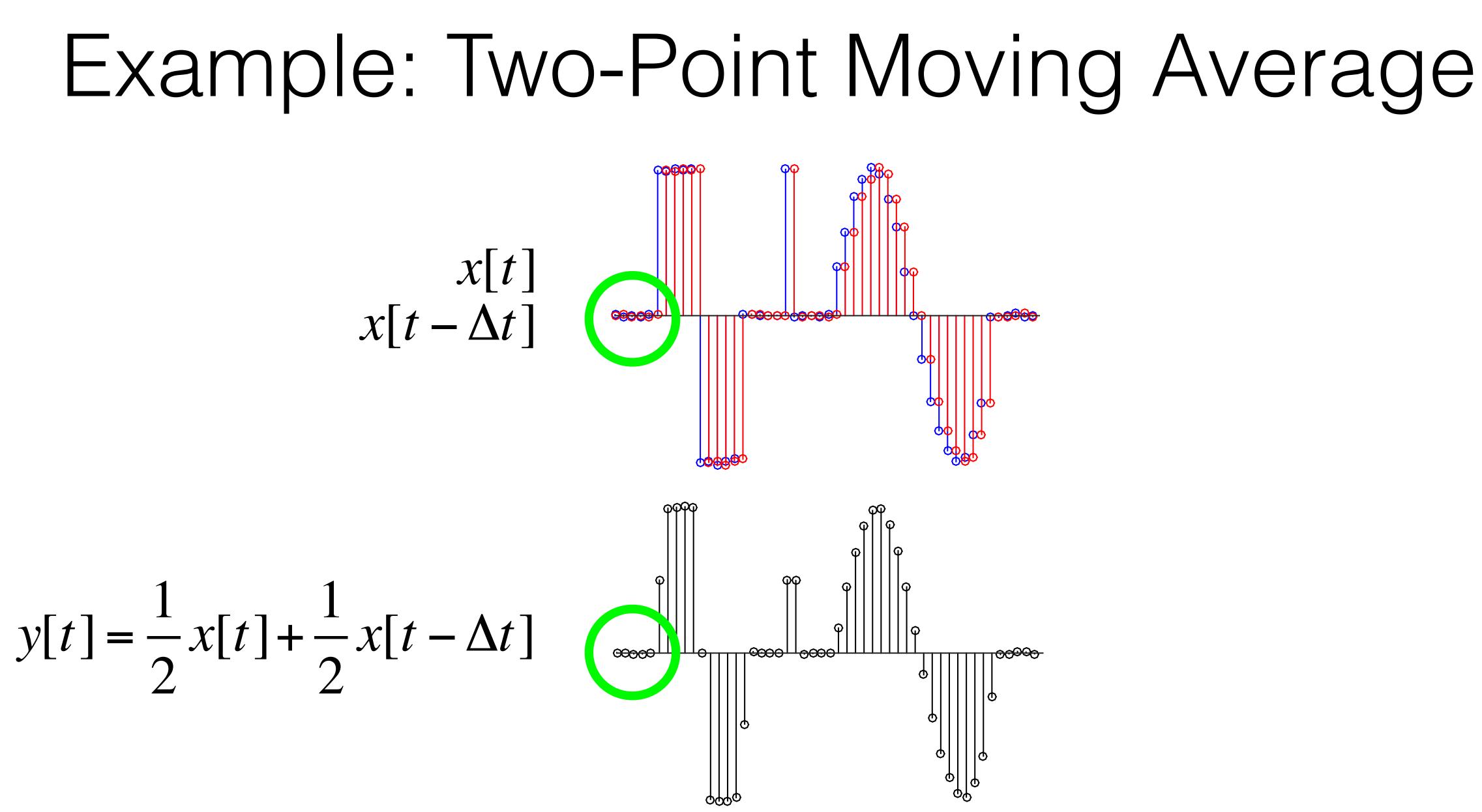


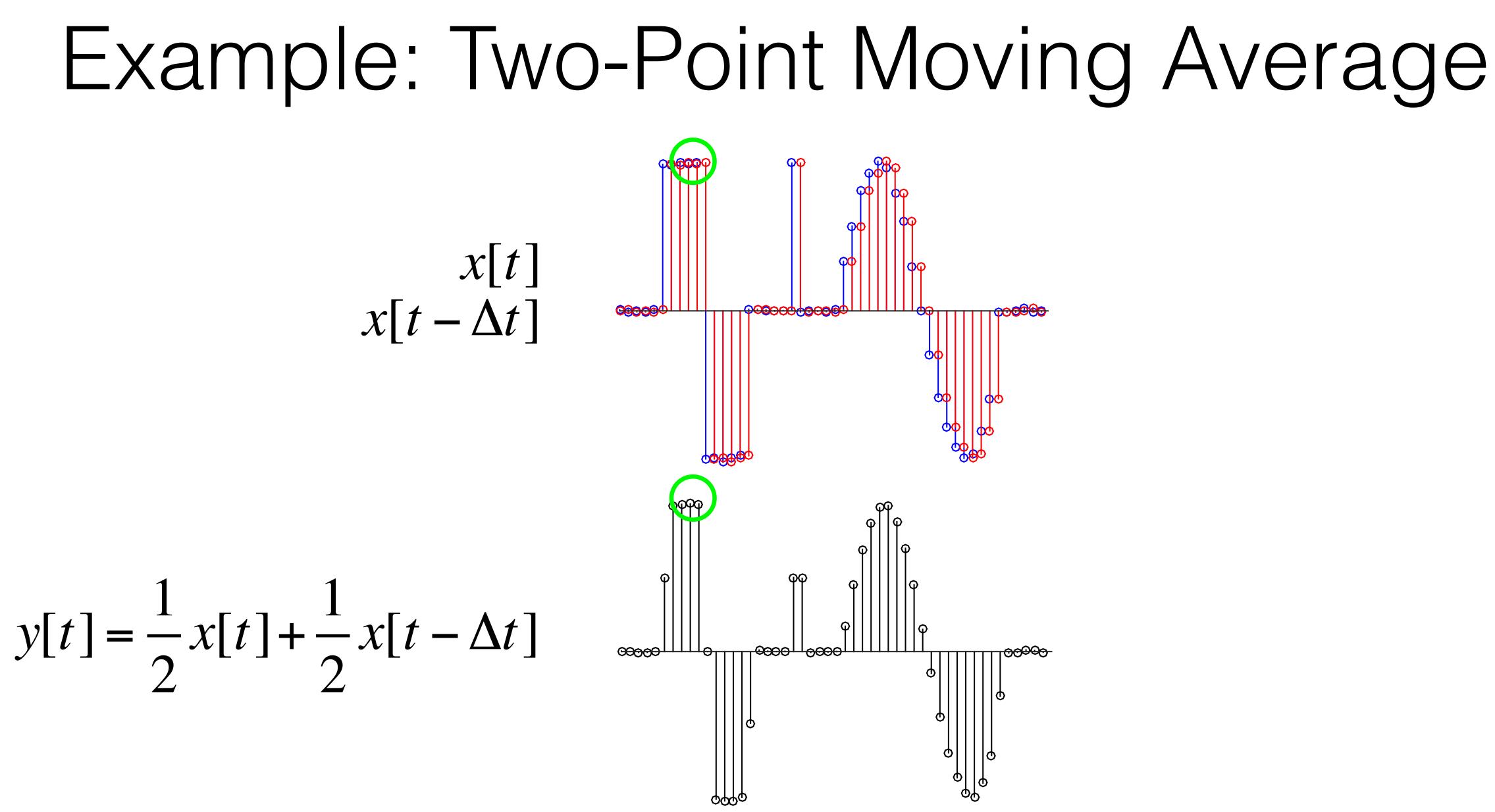


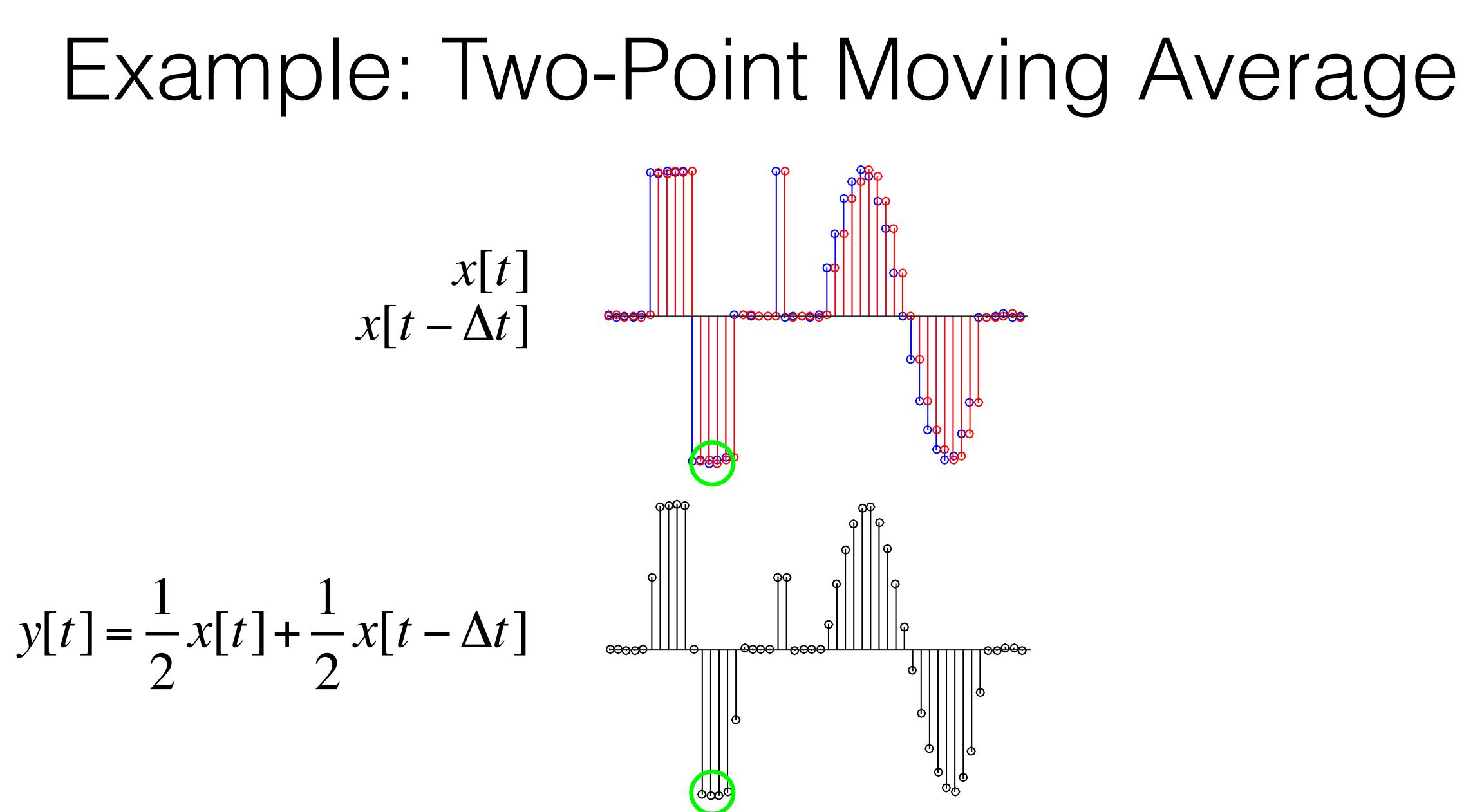


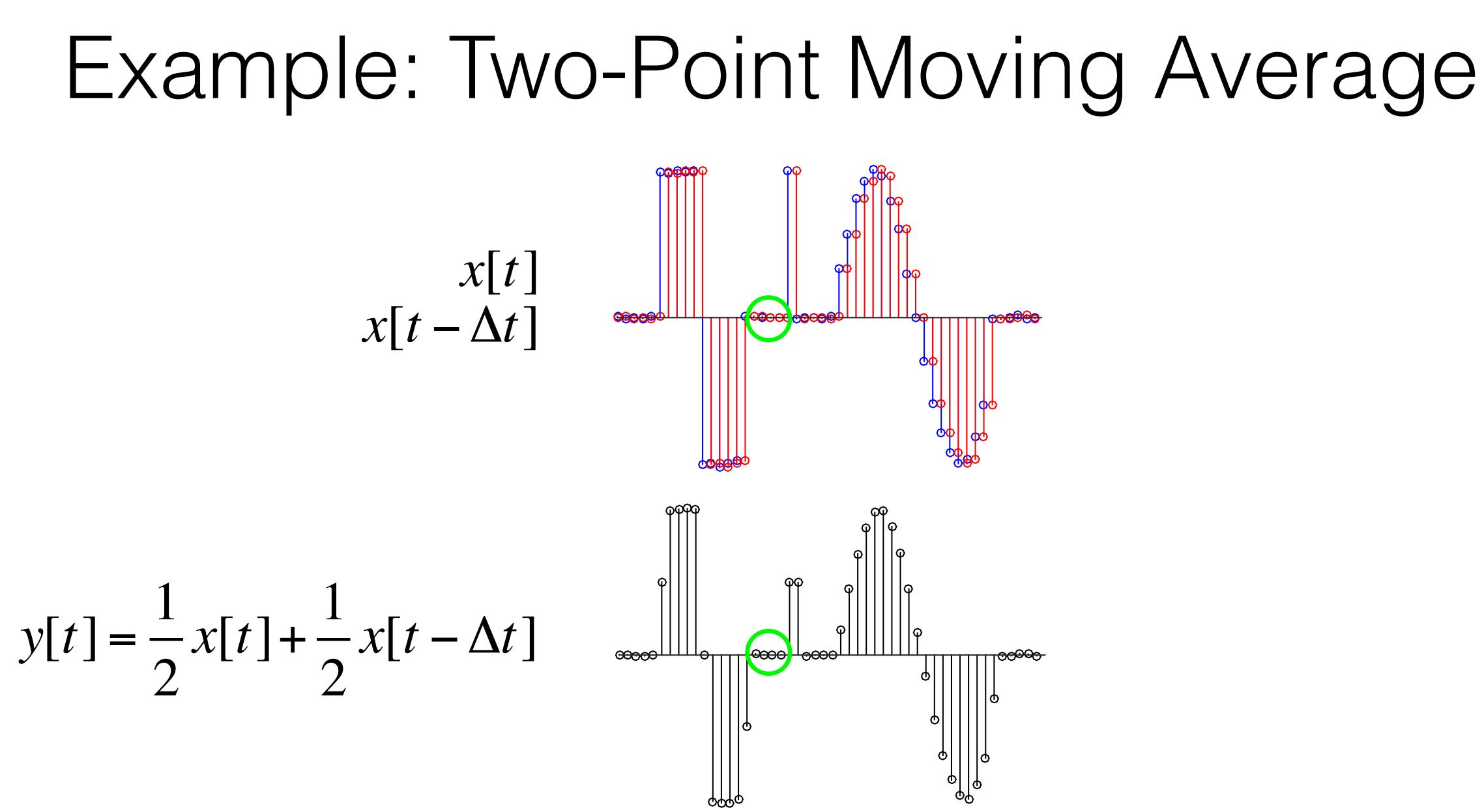


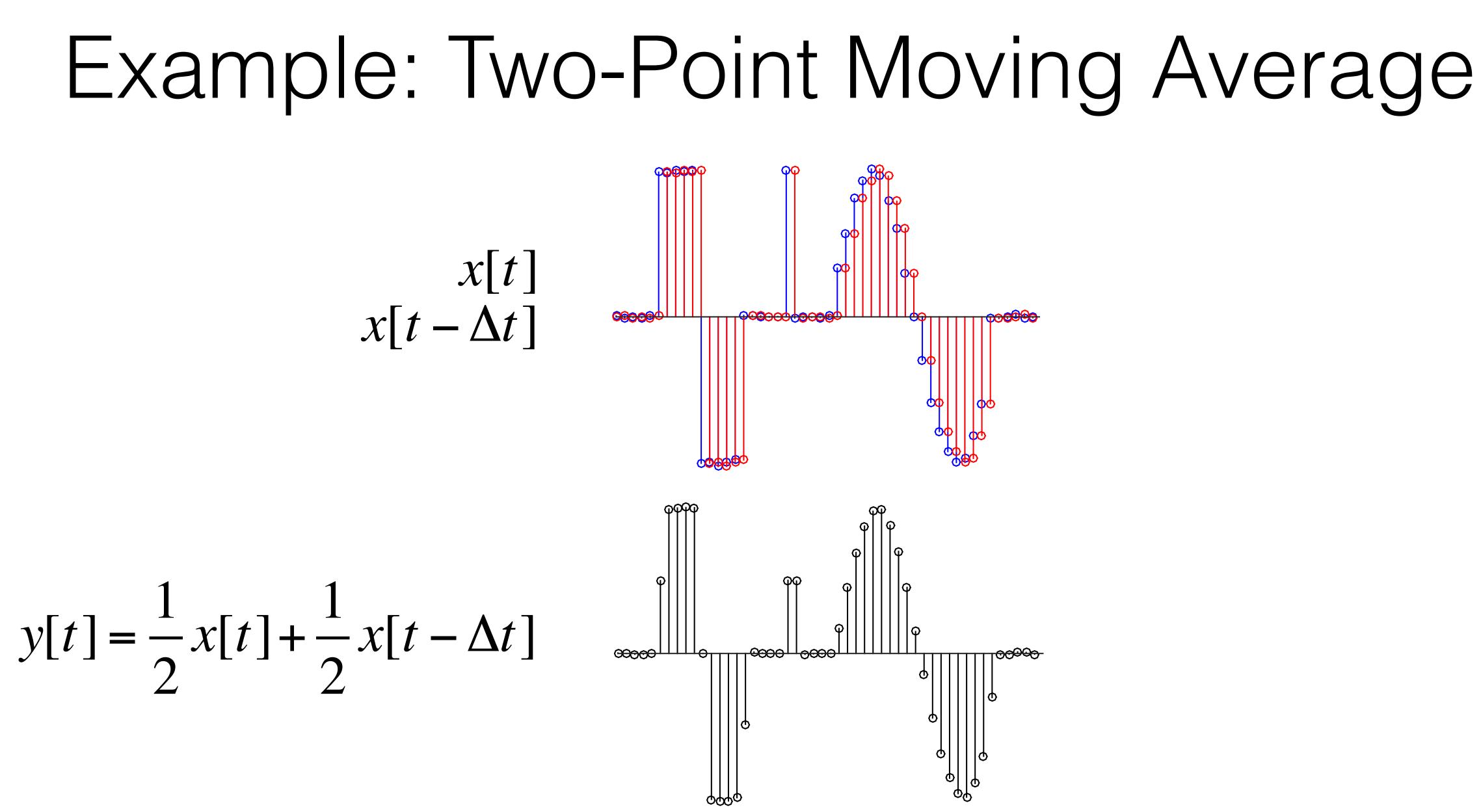


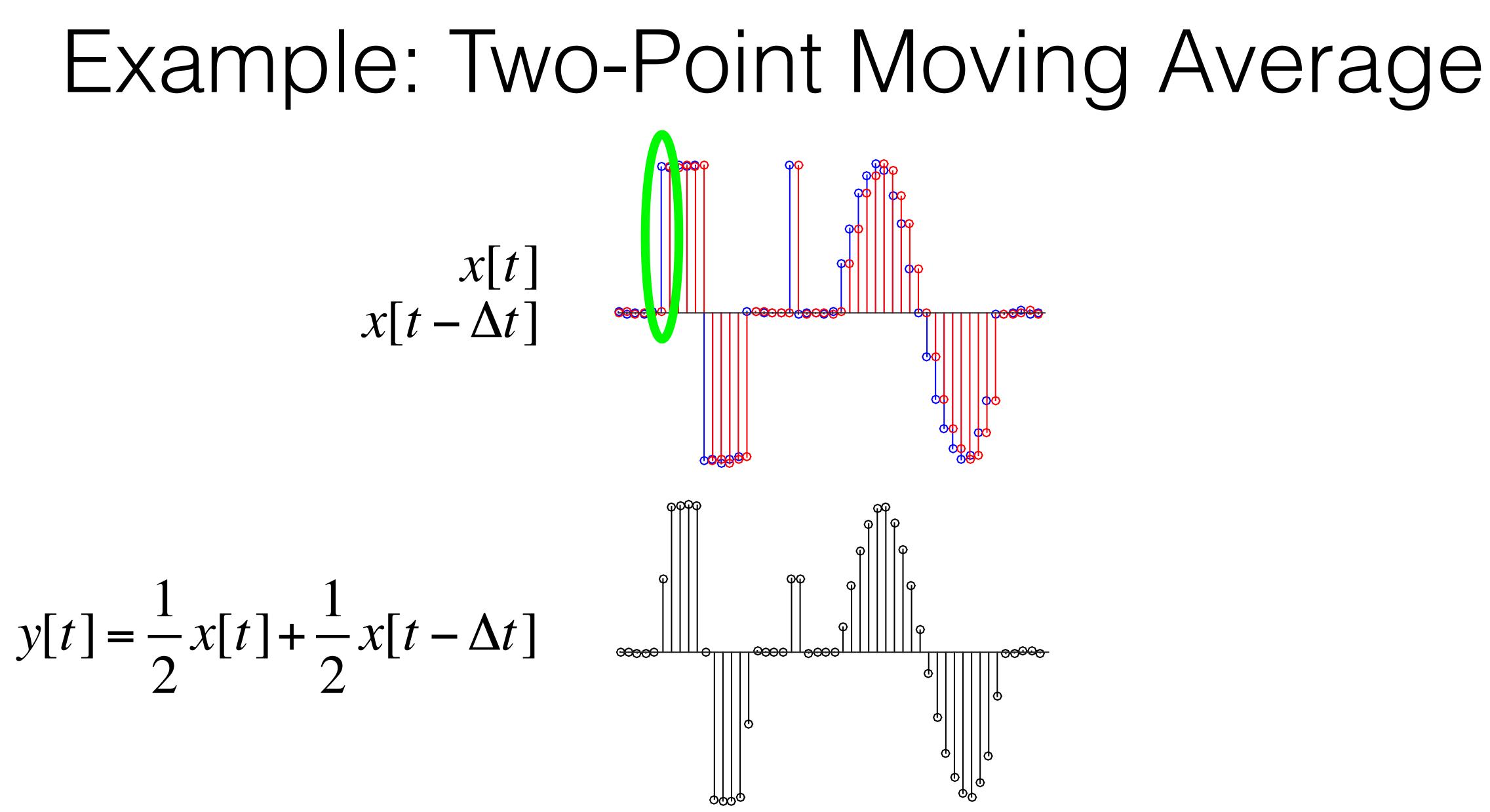


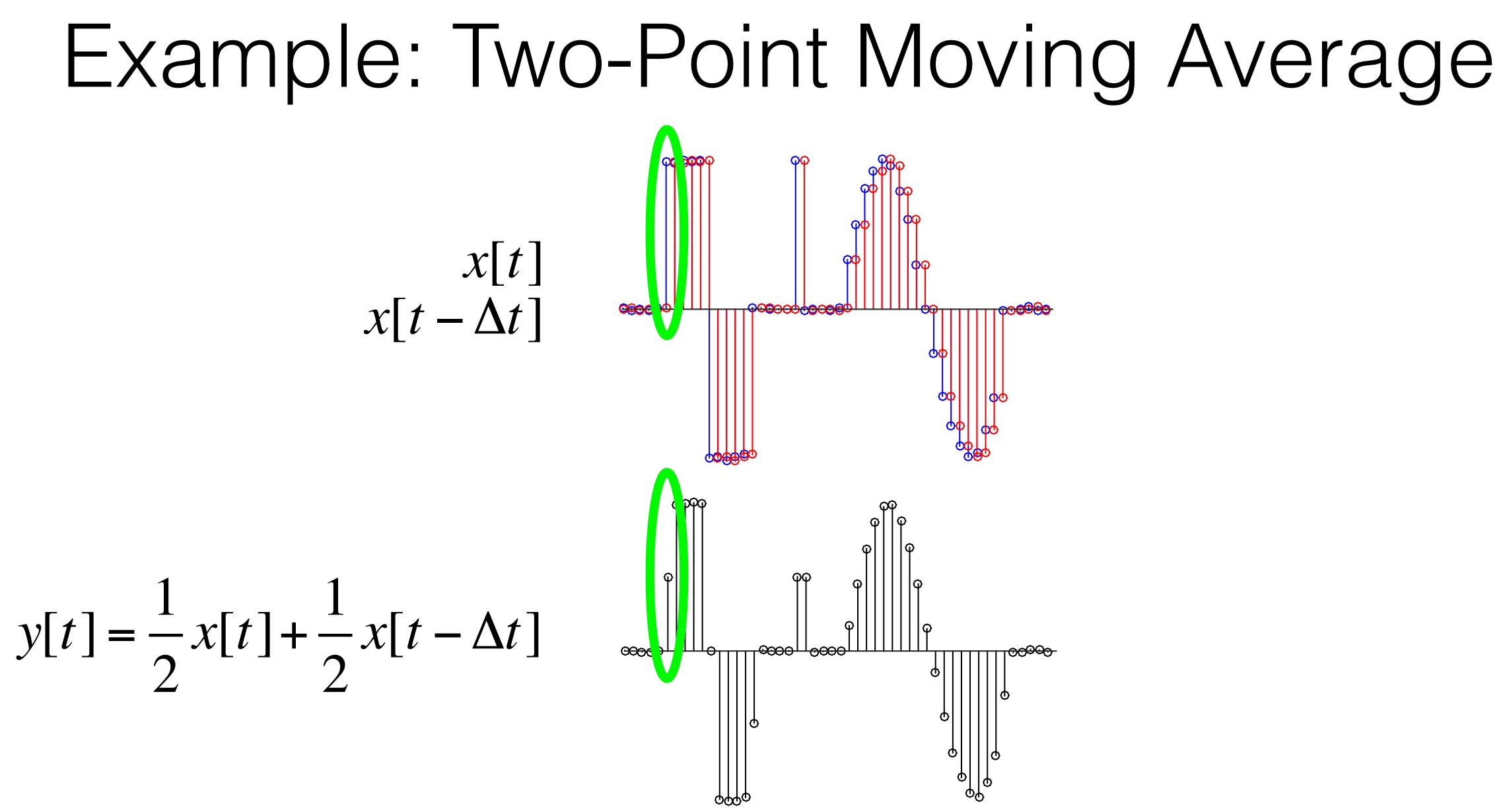


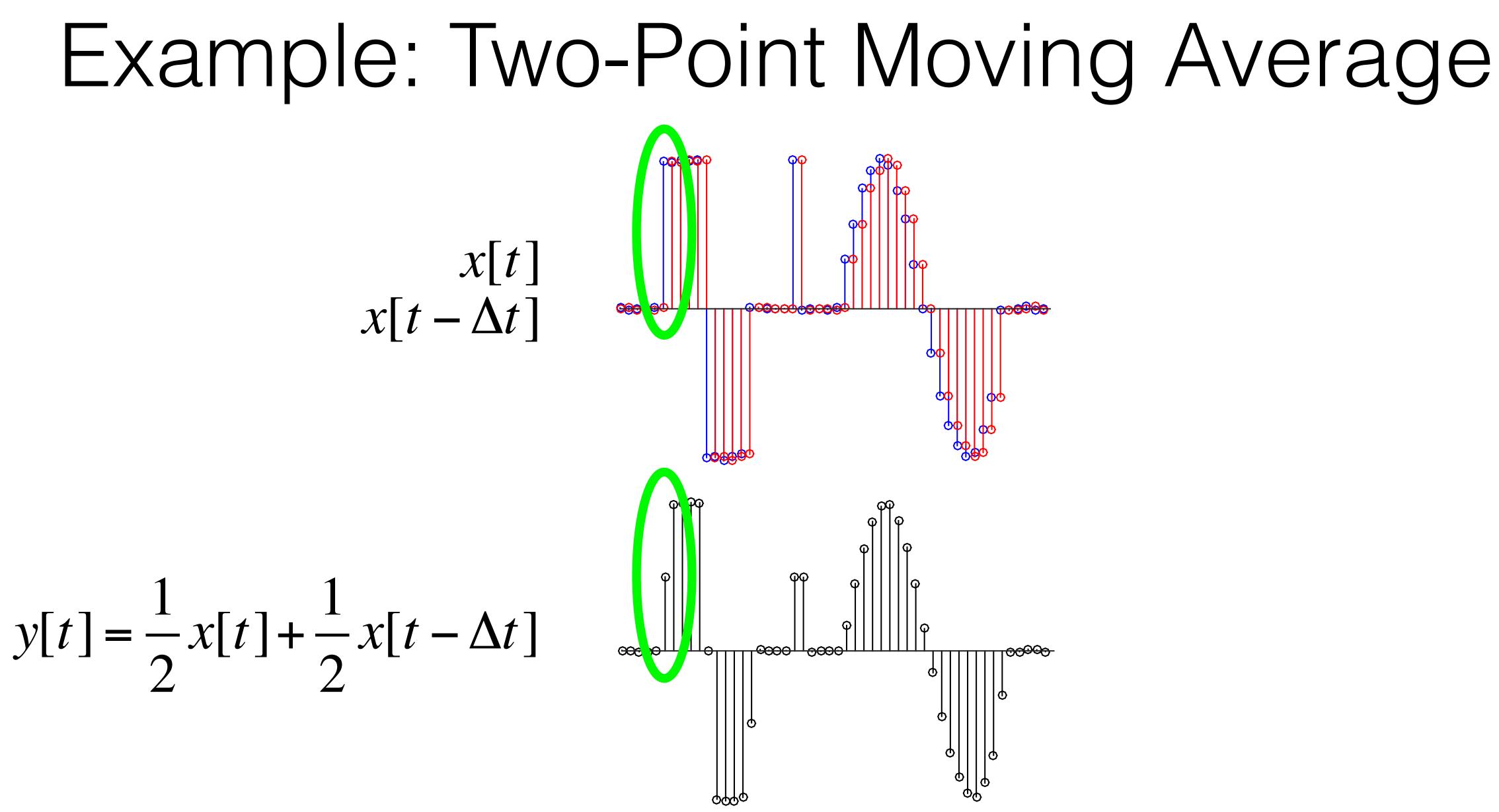


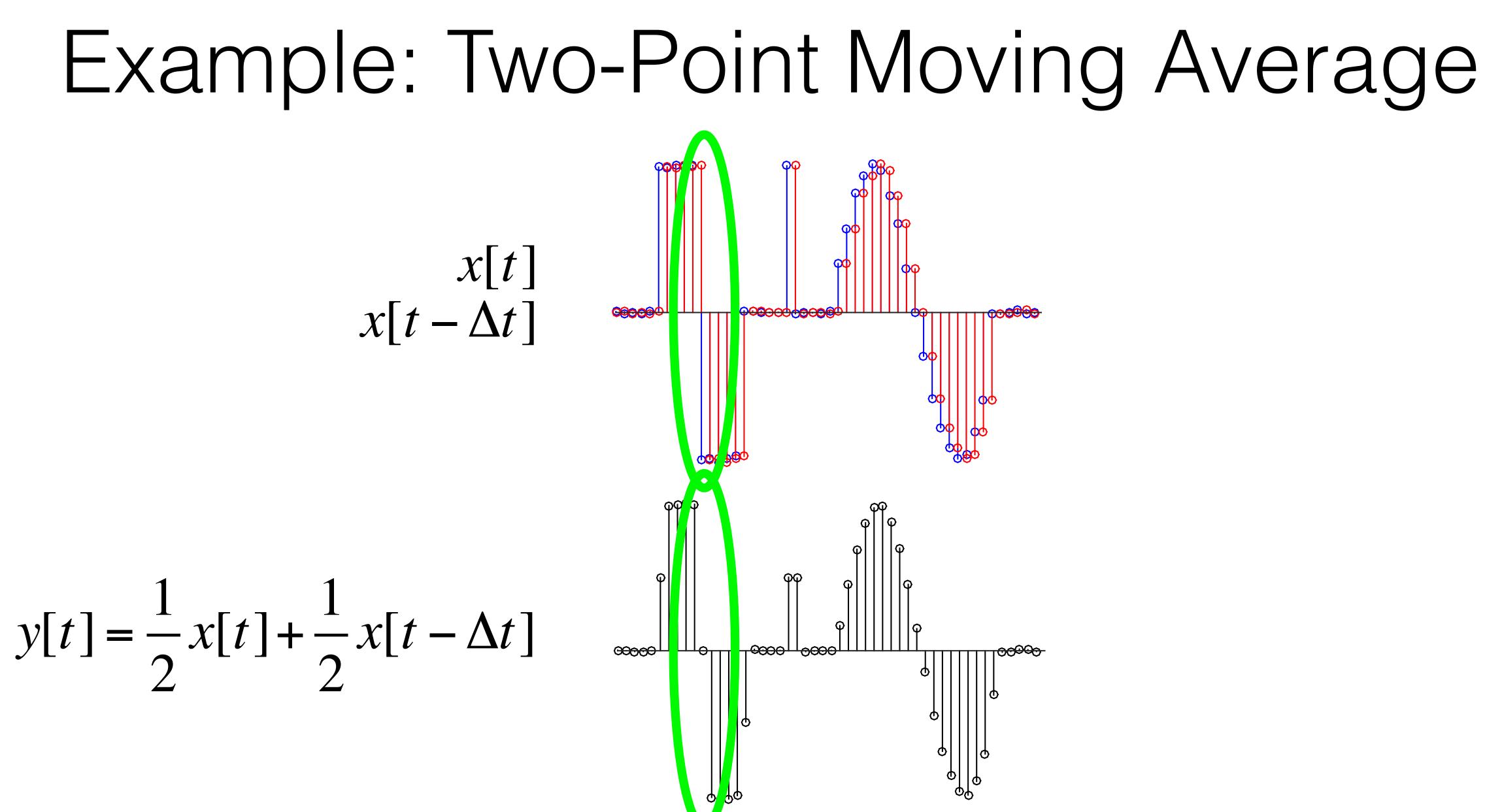


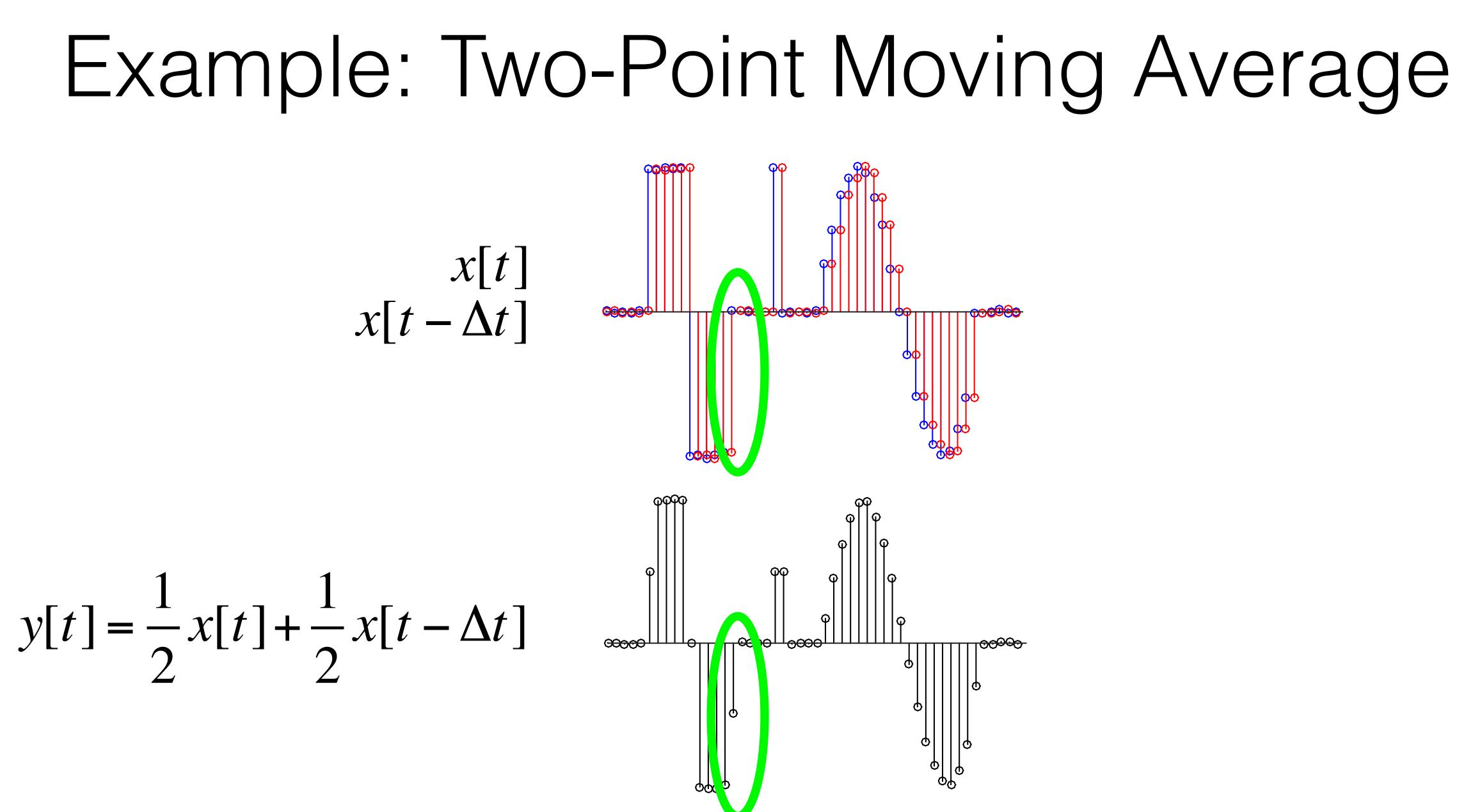


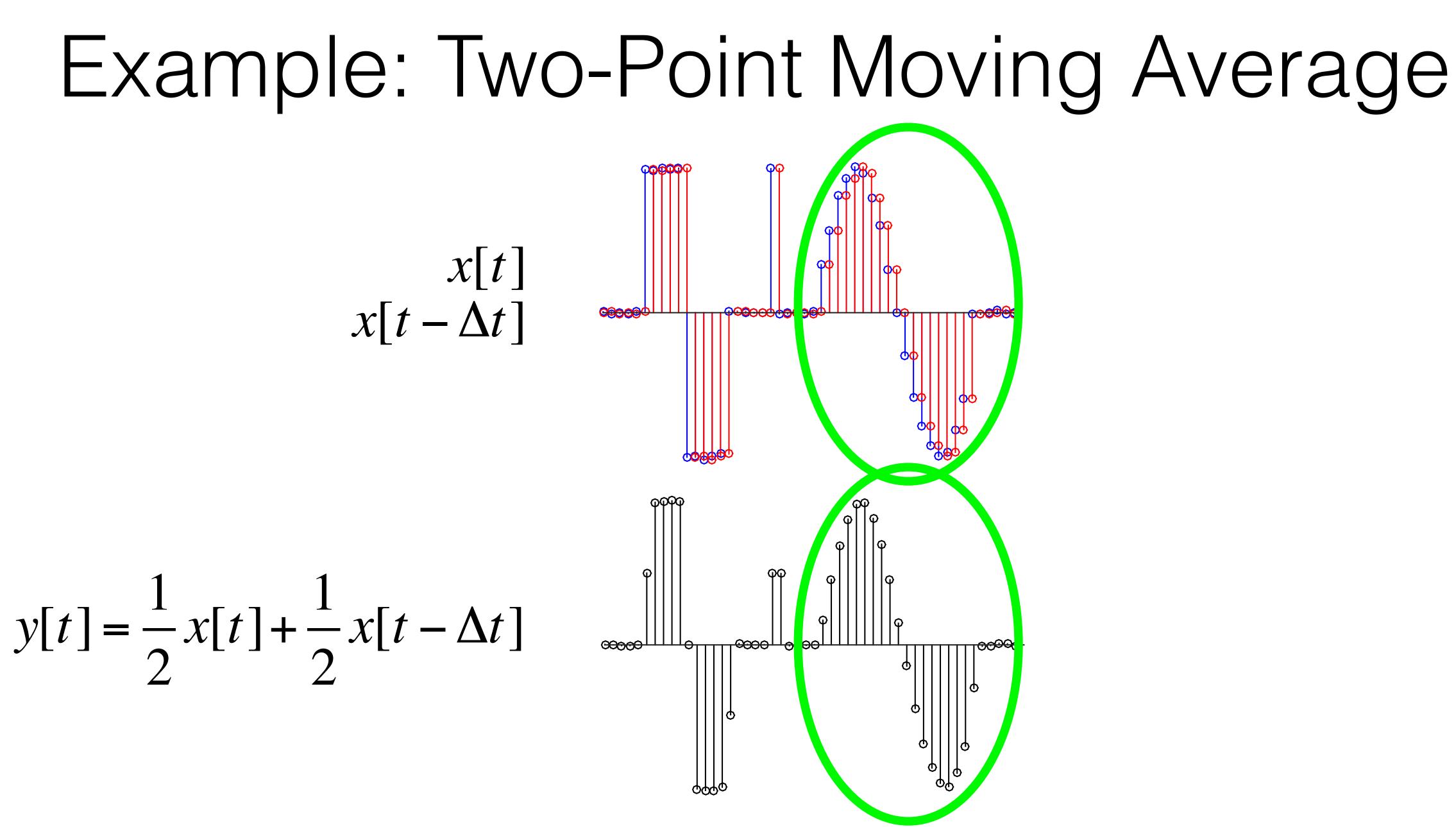


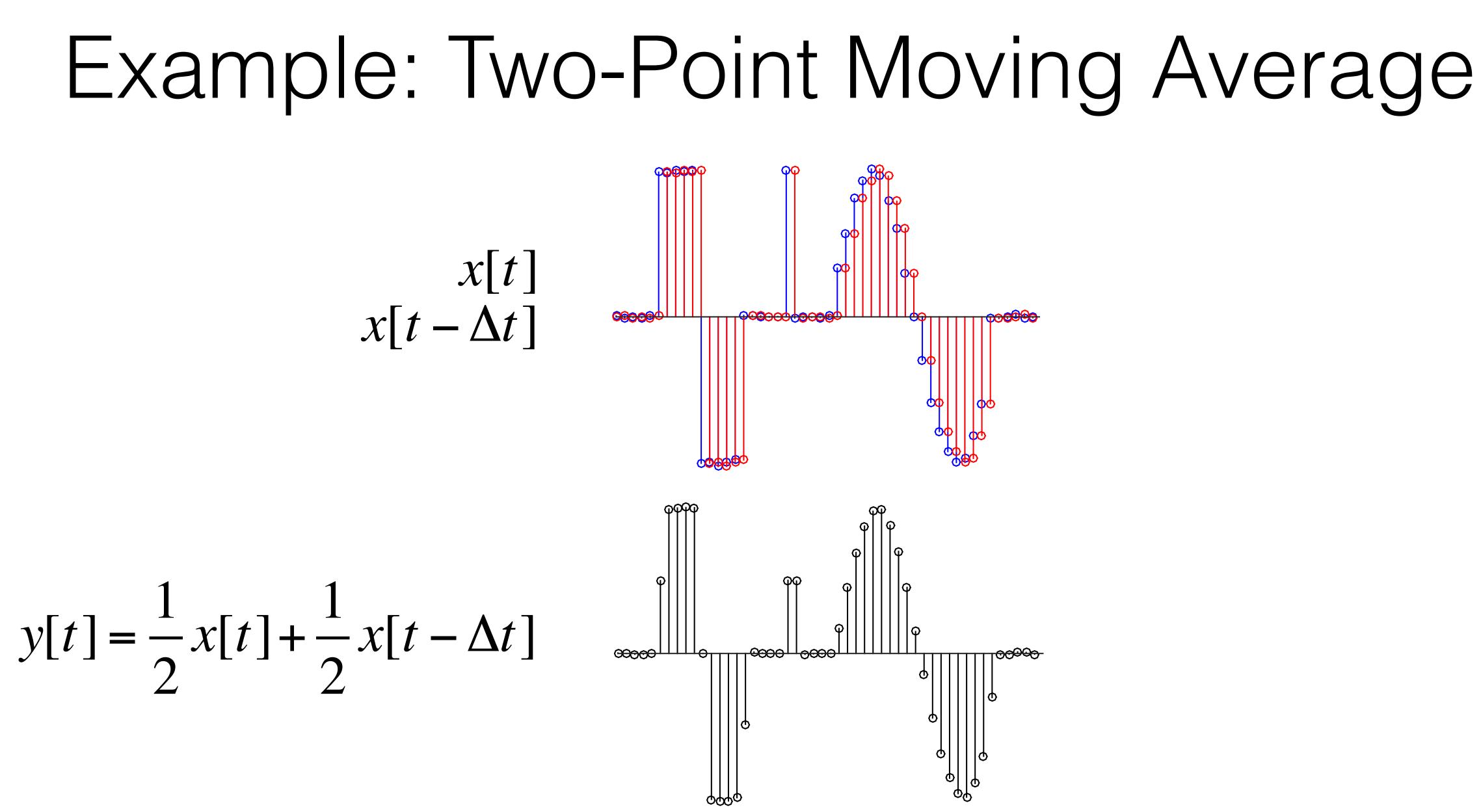


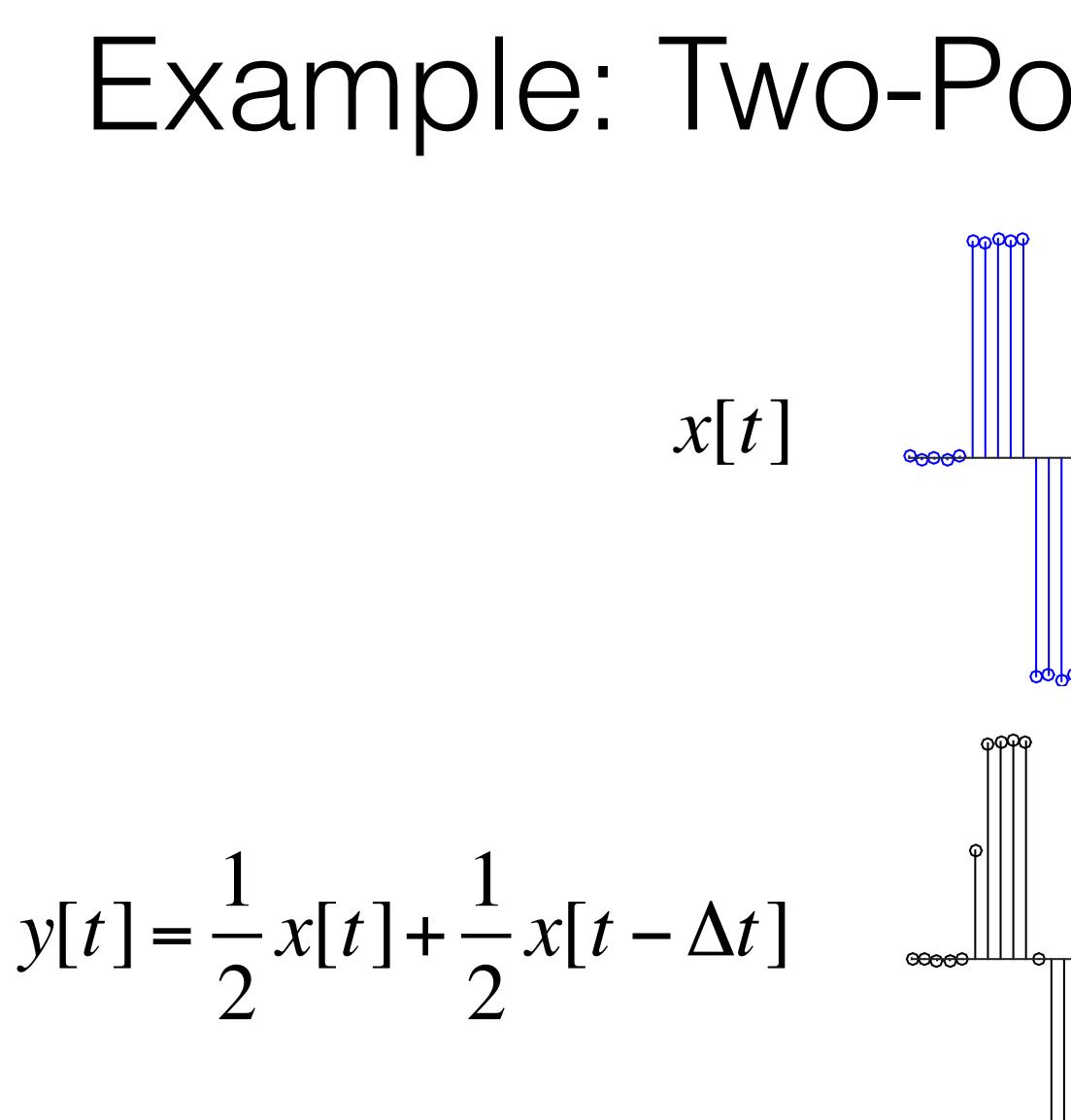








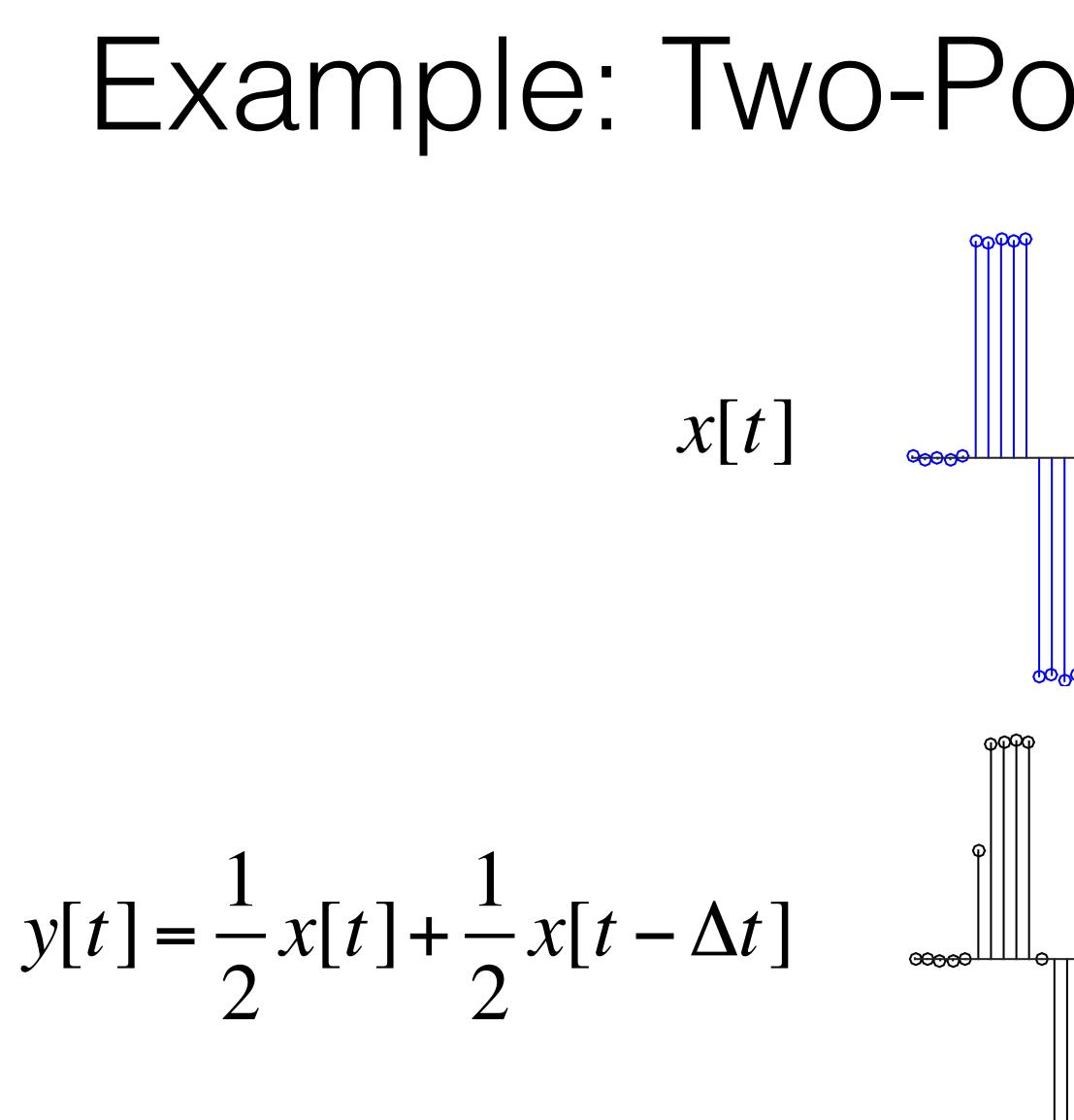




Example: Two-Point Moving Average **Results:**

<u>0000</u>

- Softens sudden changes
- Leaves slowly varying signals largely unchanged
- Slight delay in output relative to input
- _ow Pass Filter?



Example: Two-Point Moving Average **Results:**

00000

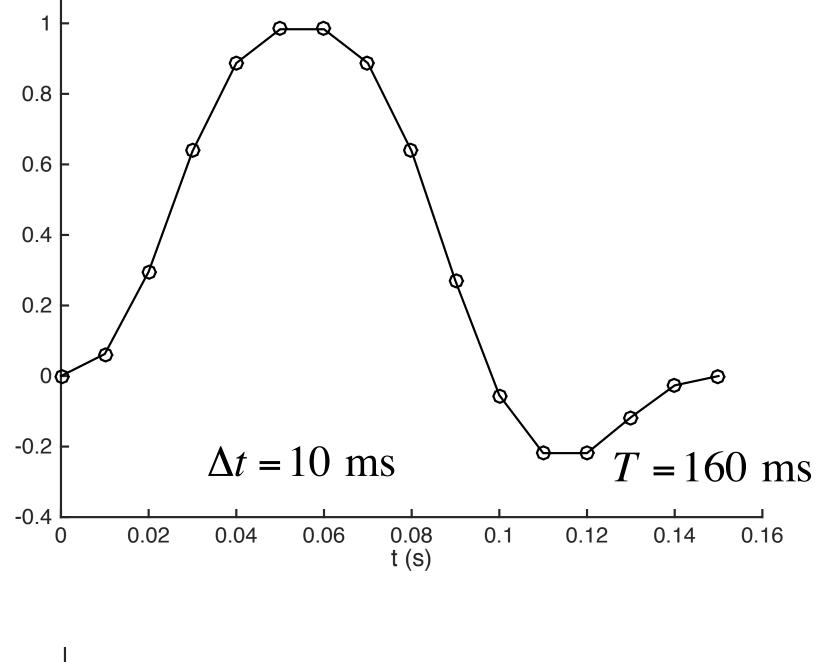
- Softens sudden changes
- Leaves slowly varying signals largely unchanged
- Slight delay in output relative to input
- ow Pass Filter?

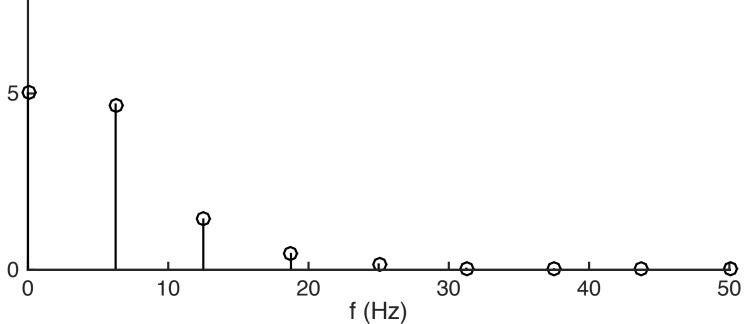




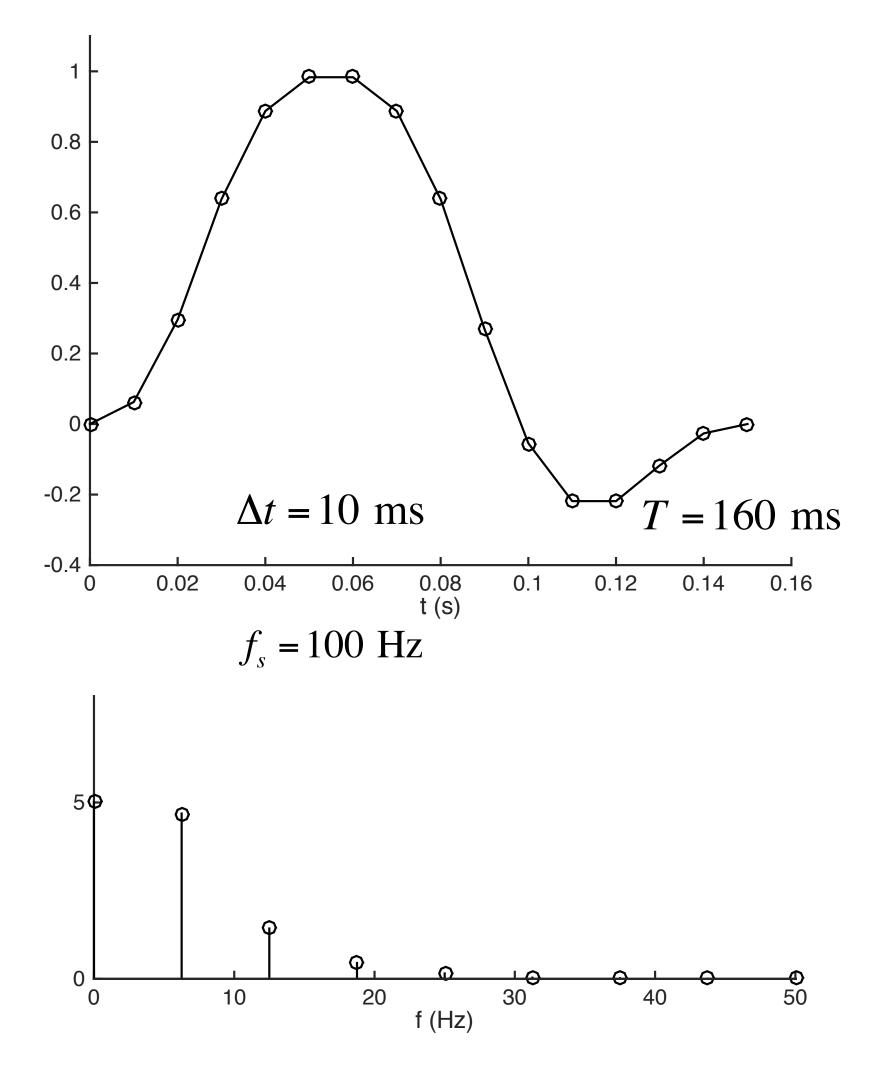


- Every Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & *frequency span* (f_s)
- Reciprocal relationship: frequency resolution (Δf) & time span (T)

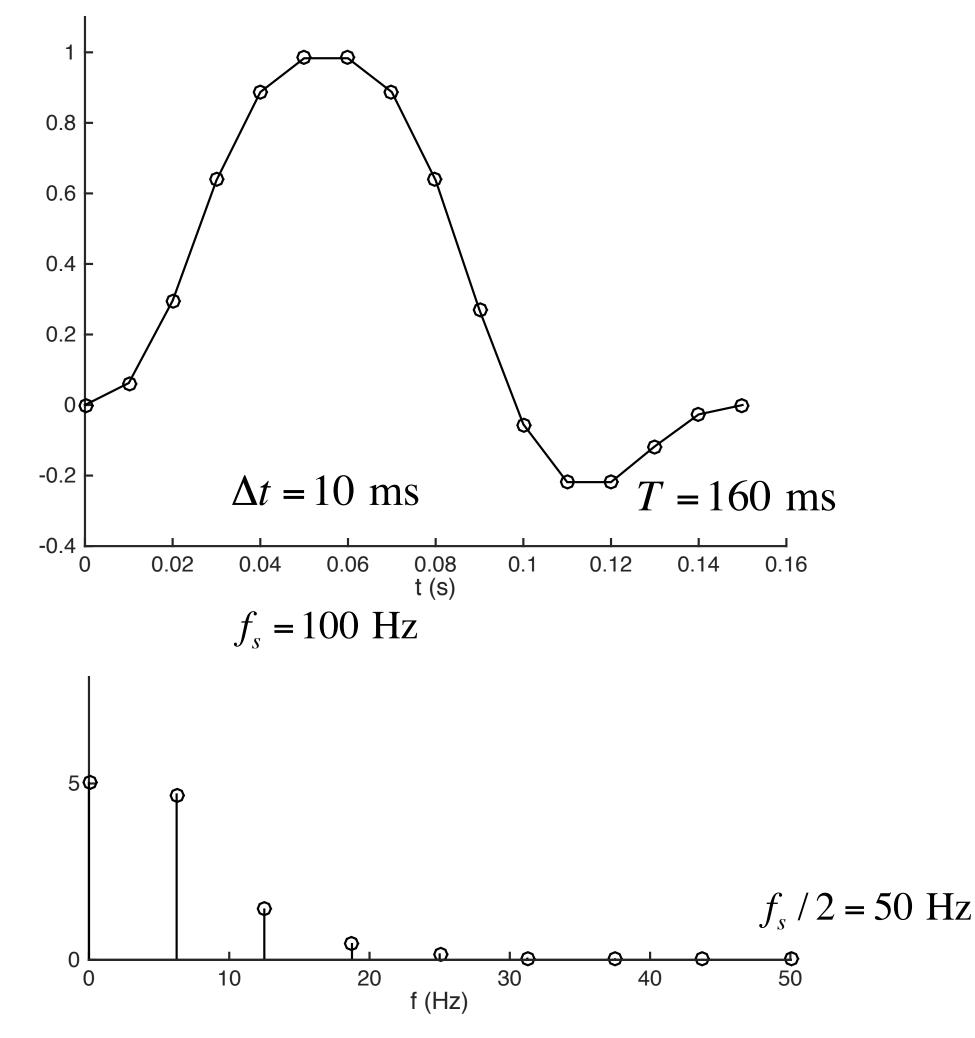




- Every Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & *frequency span* (f_s)
- Reciprocal relationship: frequency resolution (Δf) & time span (T)

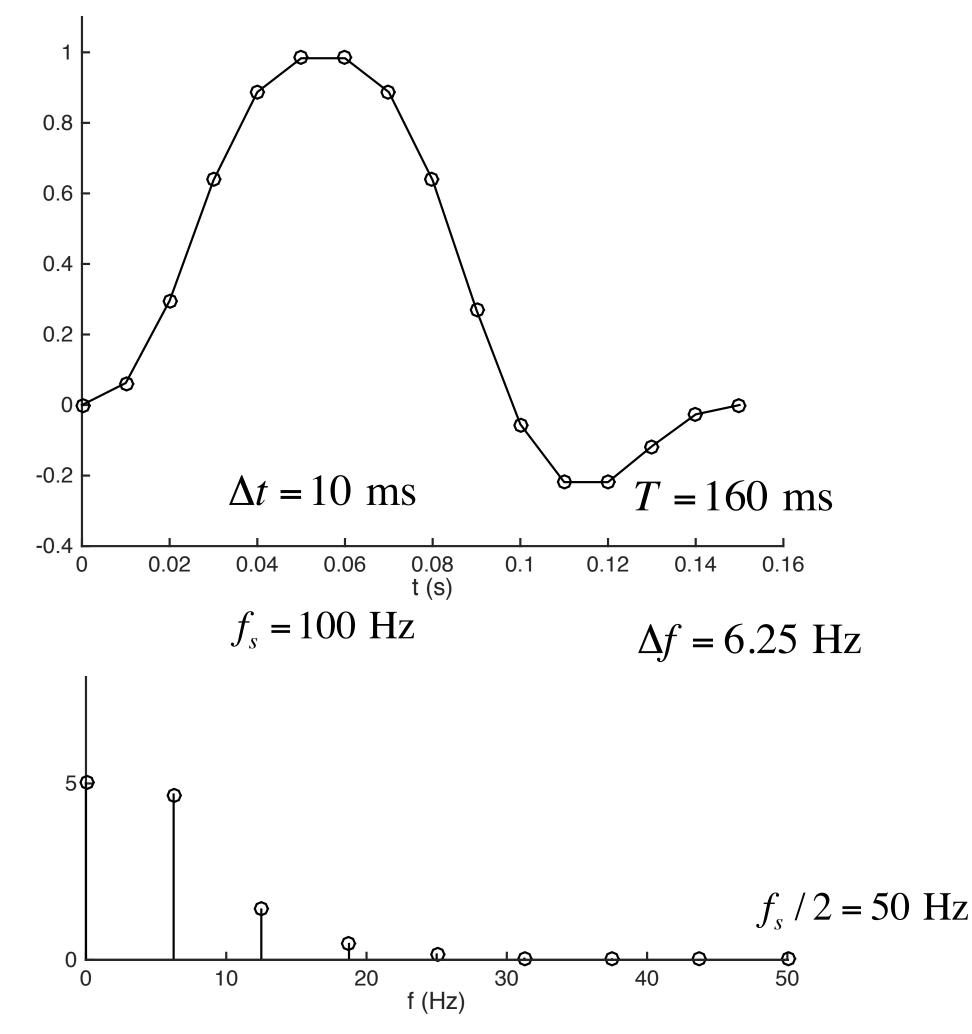


- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & frequency span (f_s)
- Reciprocal relationship: *frequency* resolution (Δf) & time span (T)





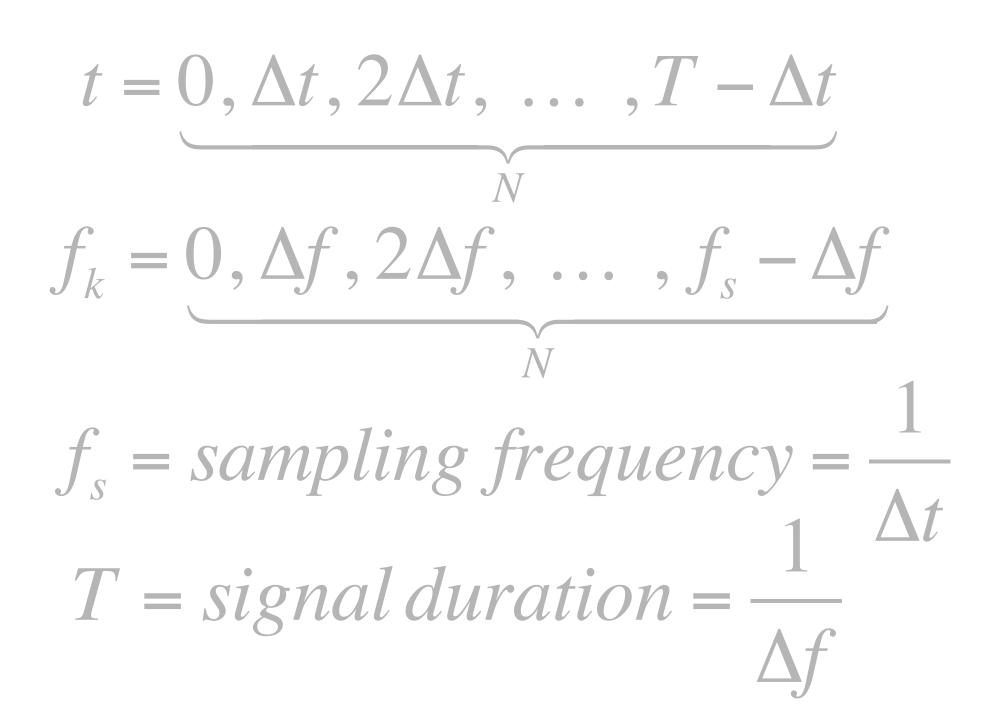
- **Every** Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & frequency span (f_s)
- Reciprocal relationship: *frequency* resolution (Δf) & time span (T)

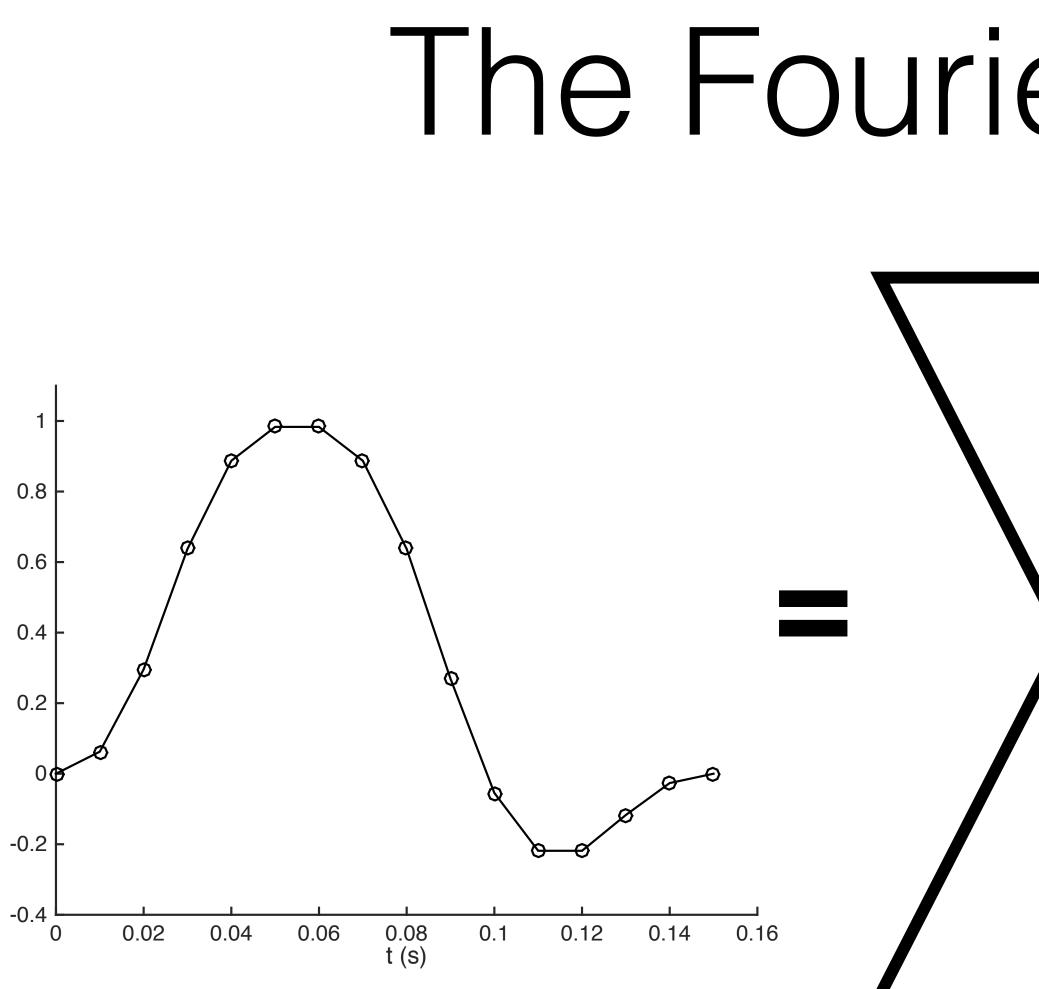


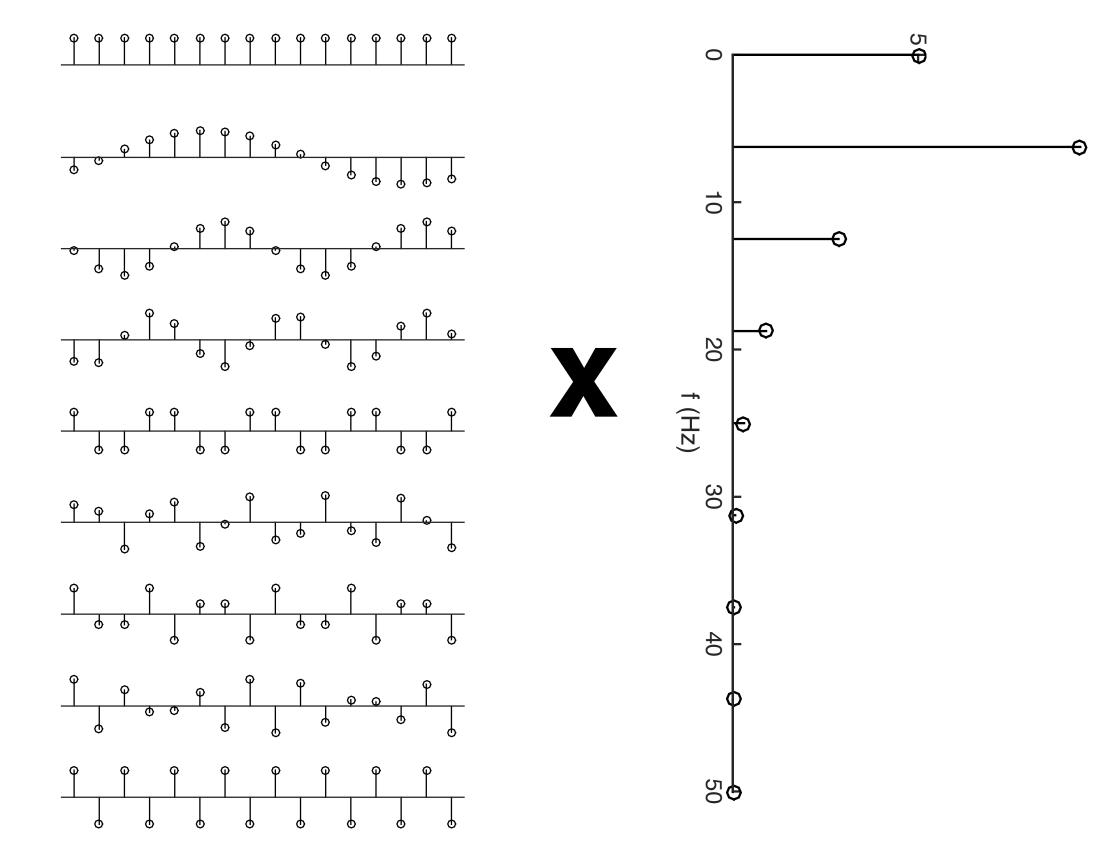


- Every Time-Domain Signal can be Re-expressed as a Sum of Sinusoids/Oscillations
- # of time points = # of frequencies
- Reciprocal relationship: *time* resolution (Δt) & *frequency span* (f_s)
- Reciprocal relationship: frequency resolution (Δf) & time span (T)

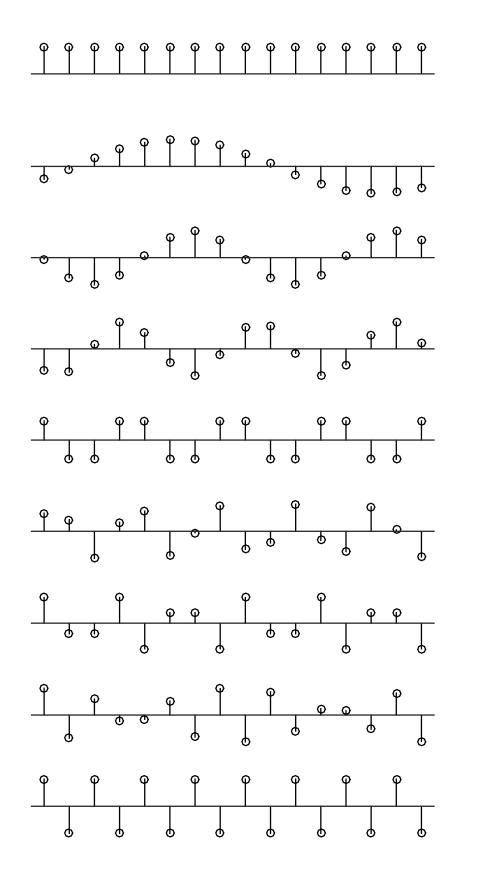






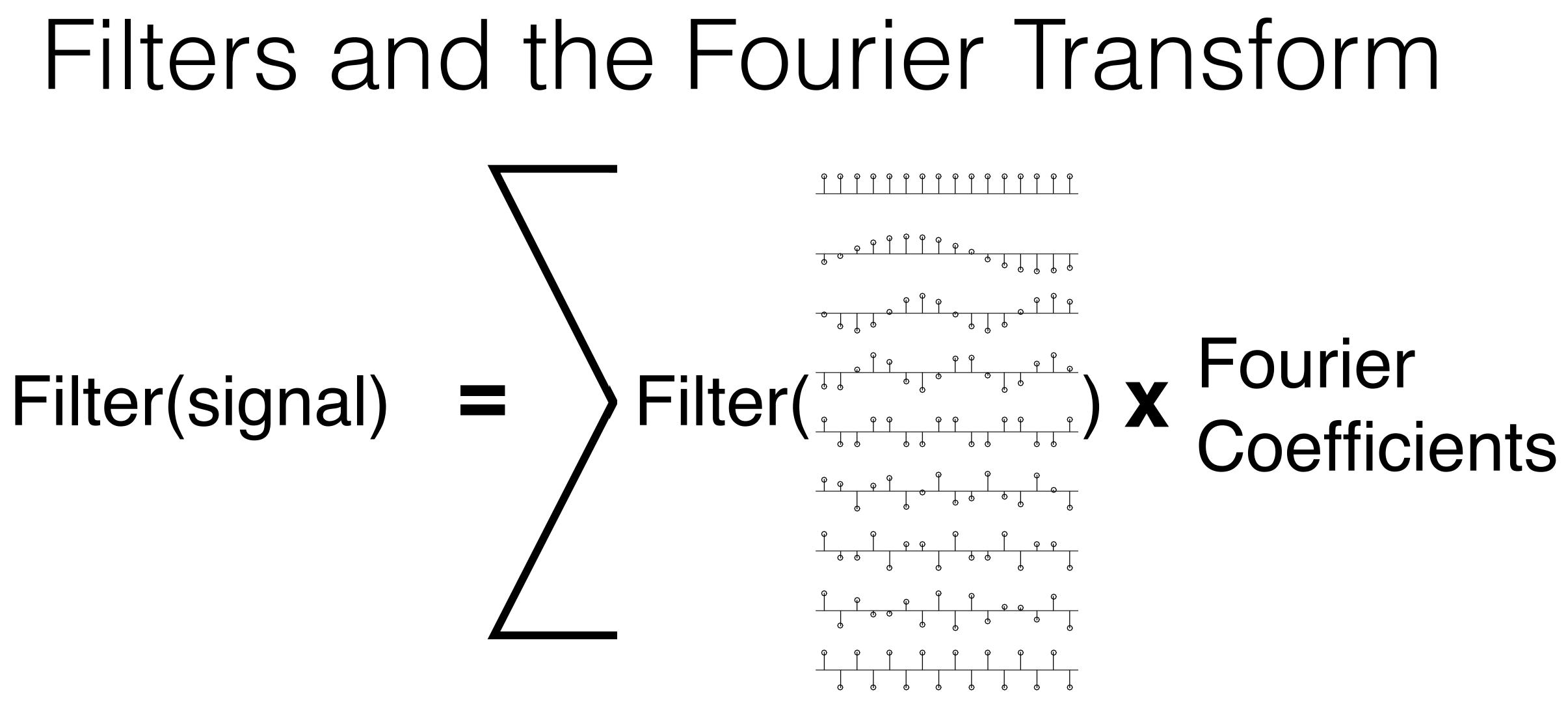


Every Signal

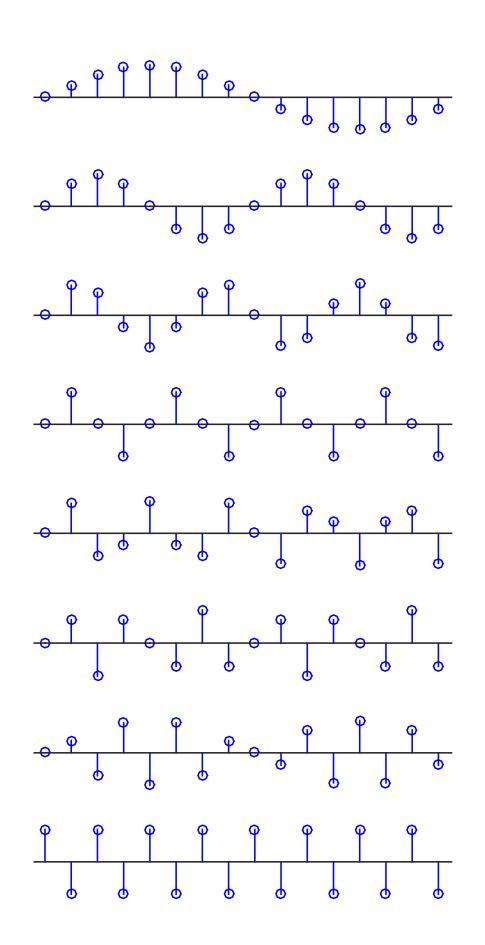


Fourier X Coefficients



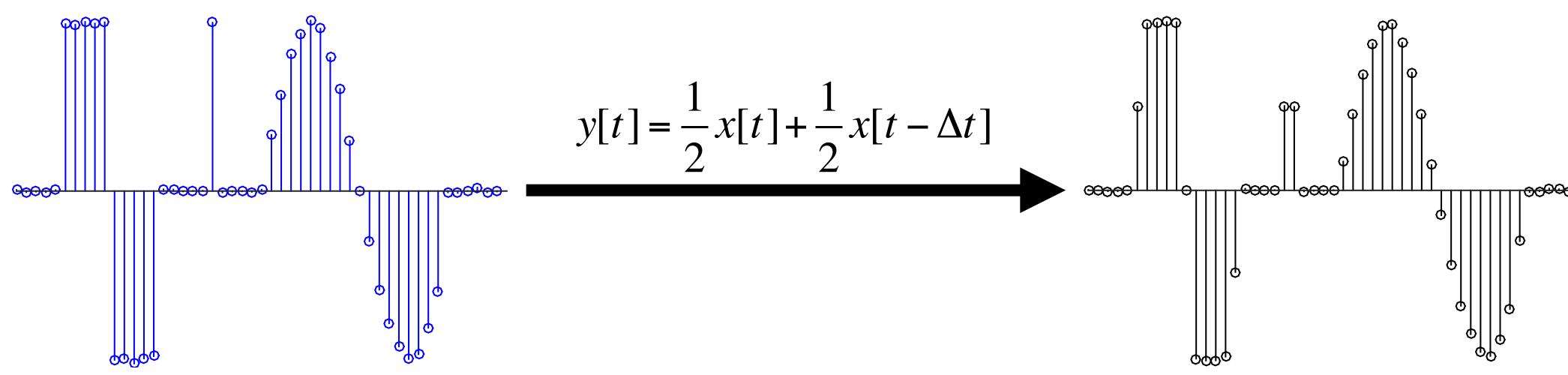


So it's really important what the filter does to these:

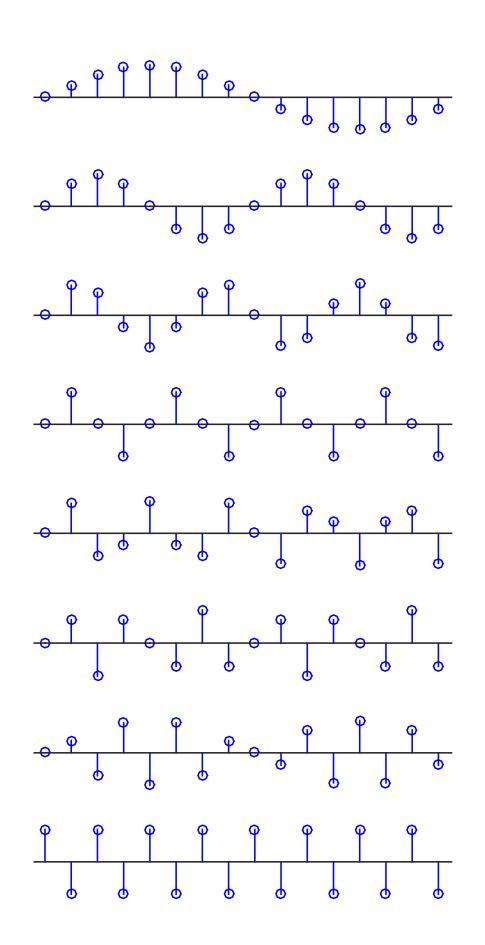


 $y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

Recall:

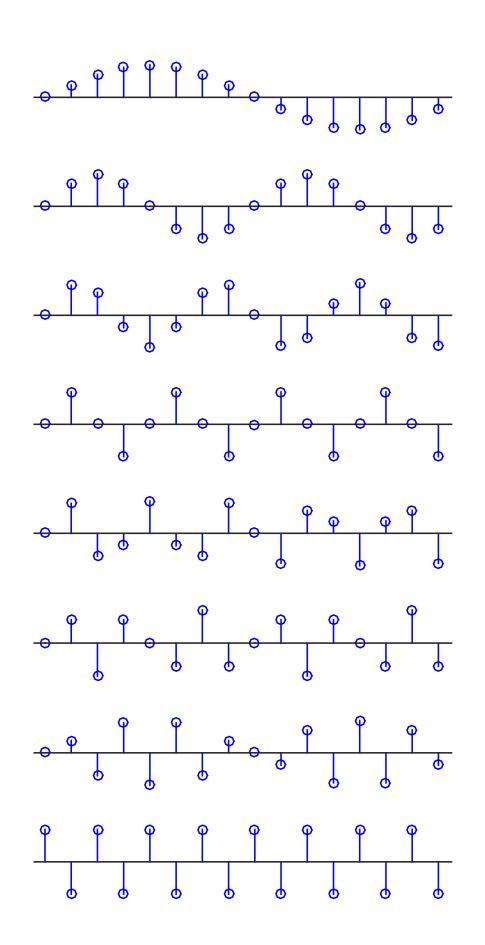


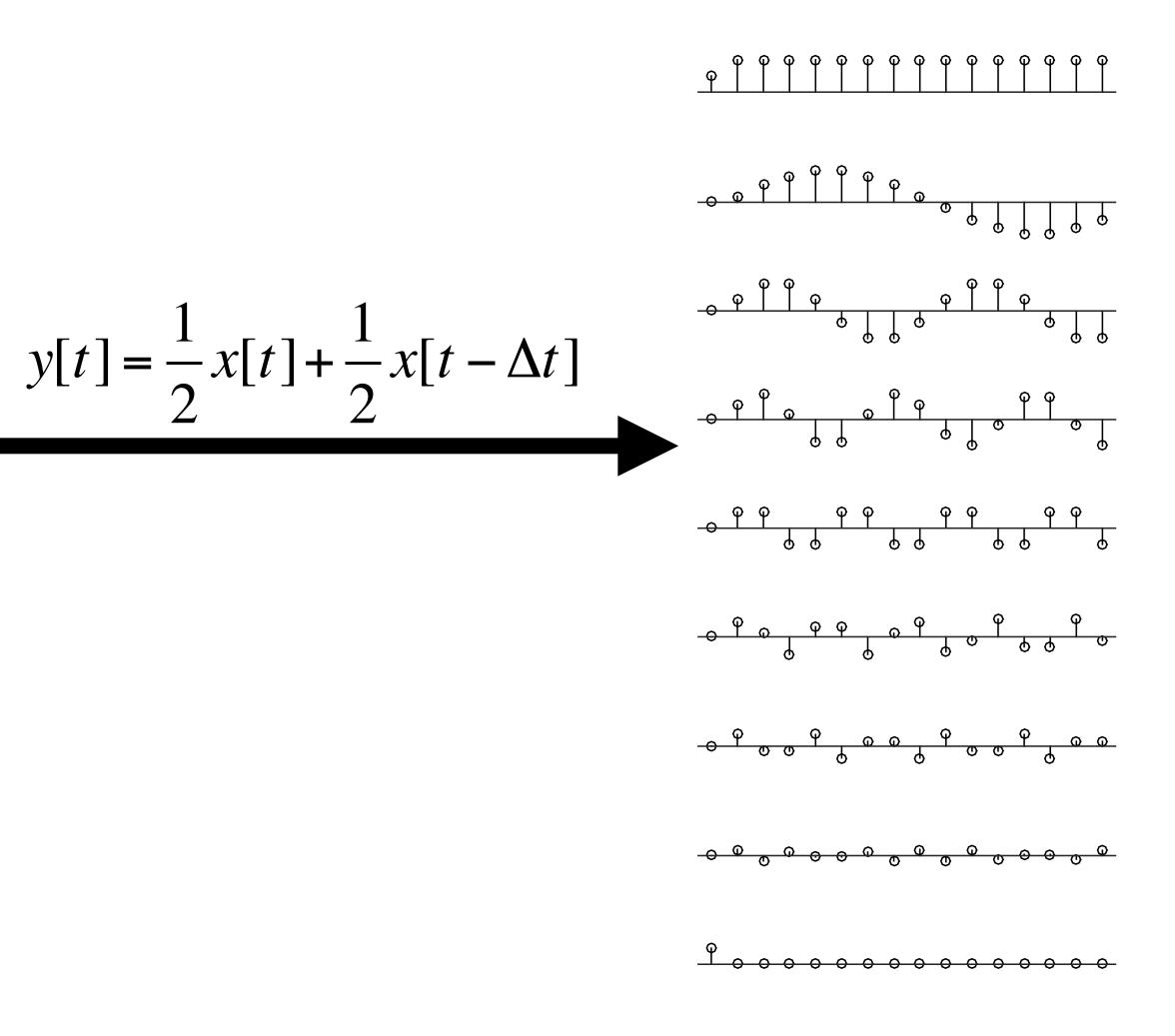
So it's really important what the filter does to these:



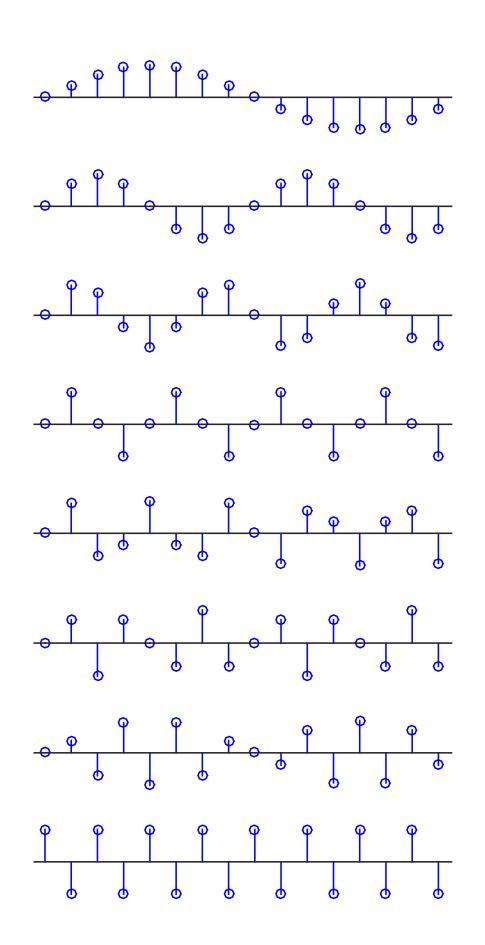
 $y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

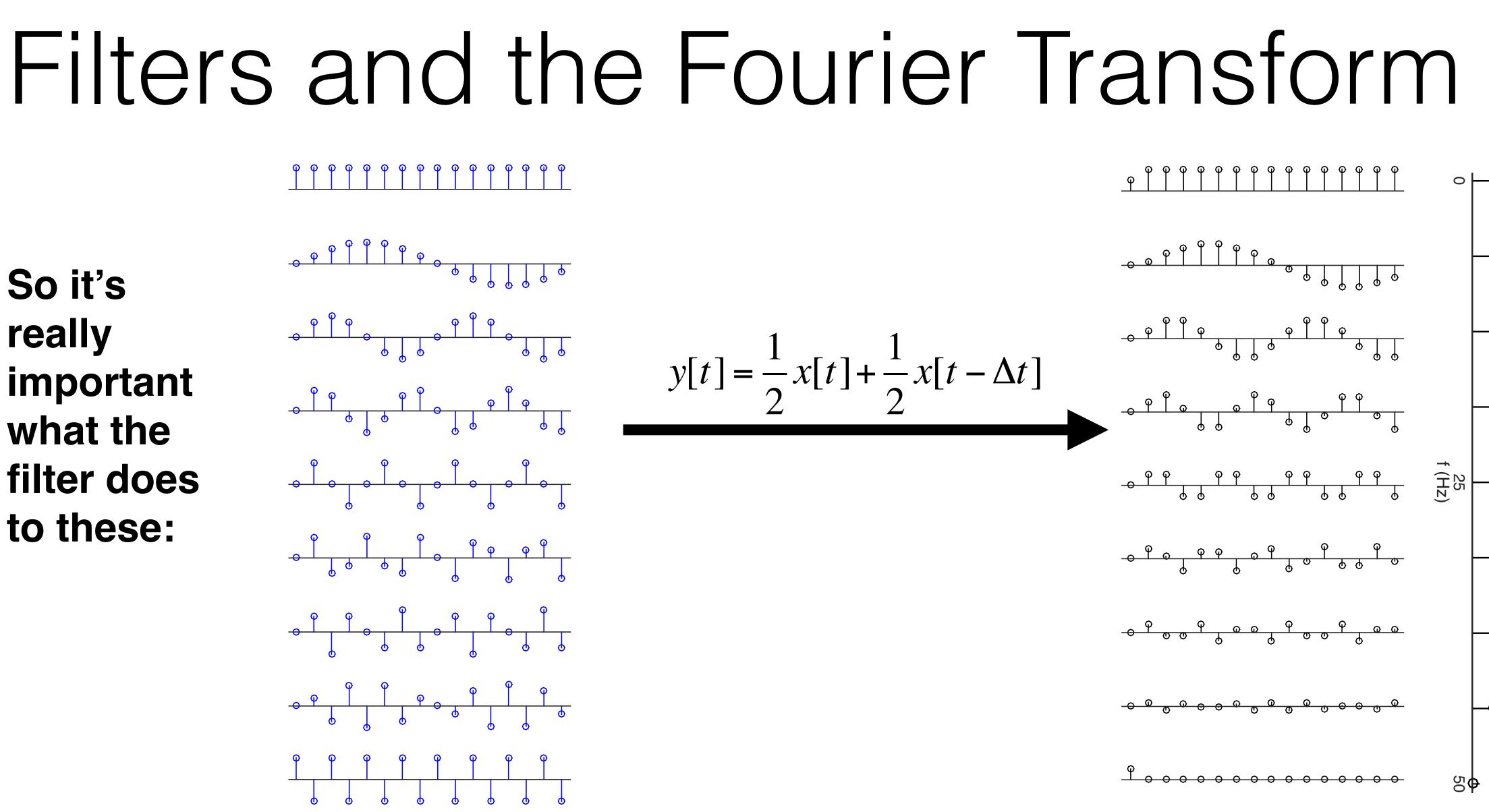
So it's really important what the filter does to these:

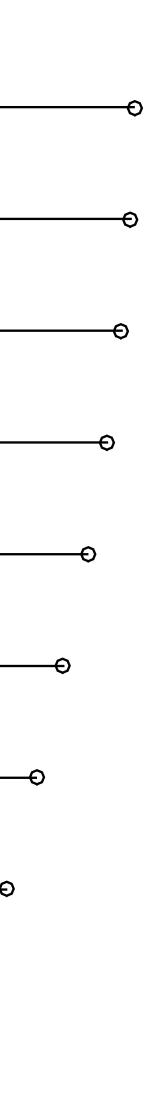


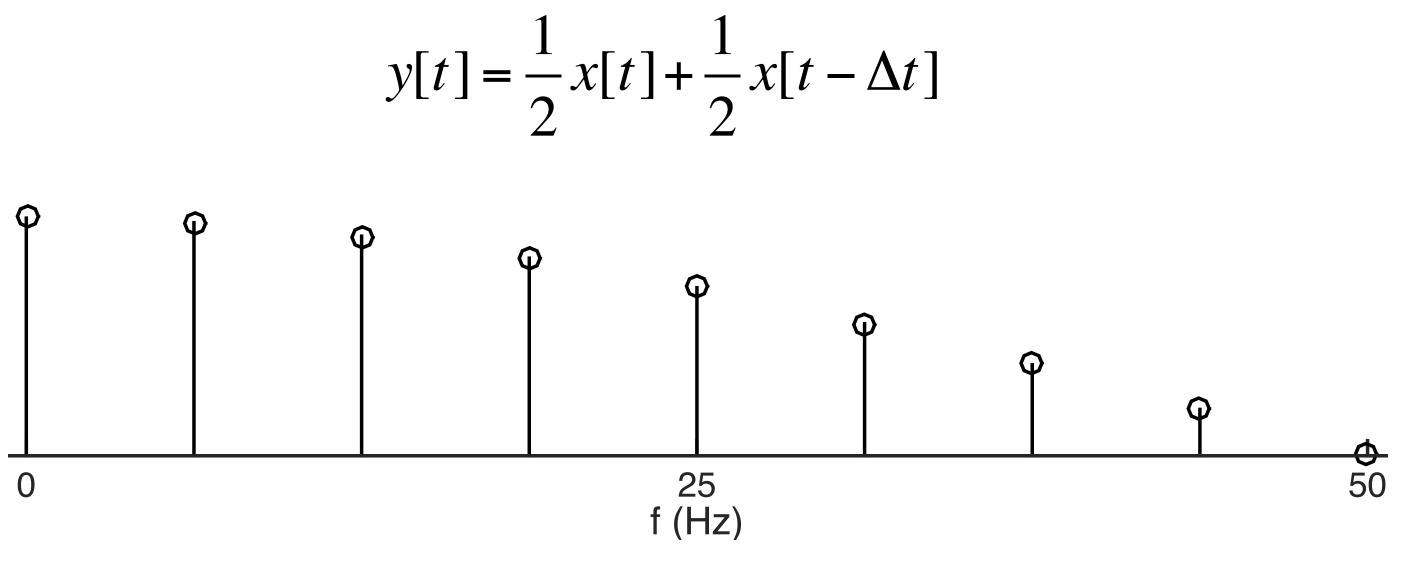


So it's really important what the filter does to these:









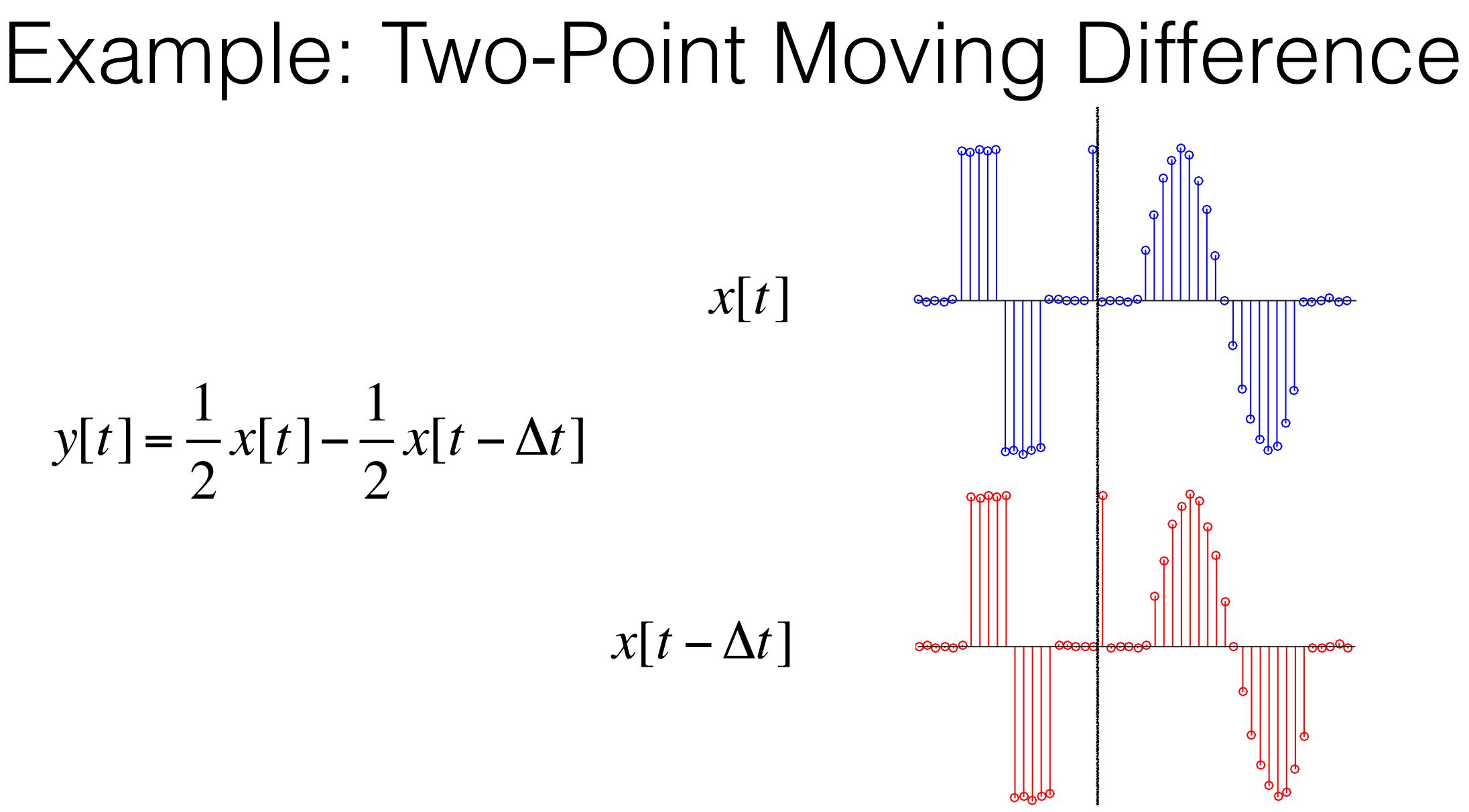
Low Pass Filter

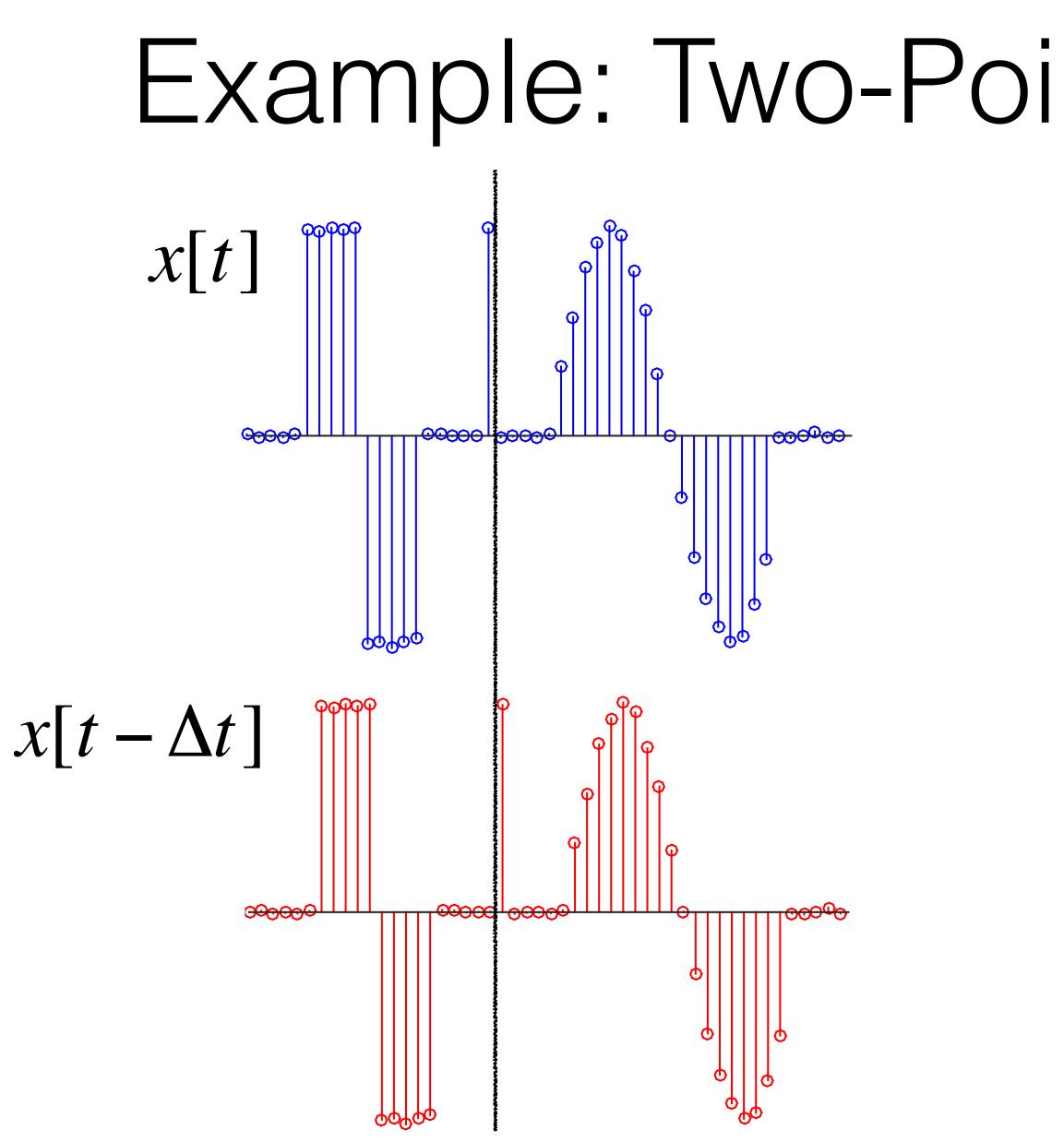
$y[t] = \frac{x[t] - x[t - \Delta t]}{2}$

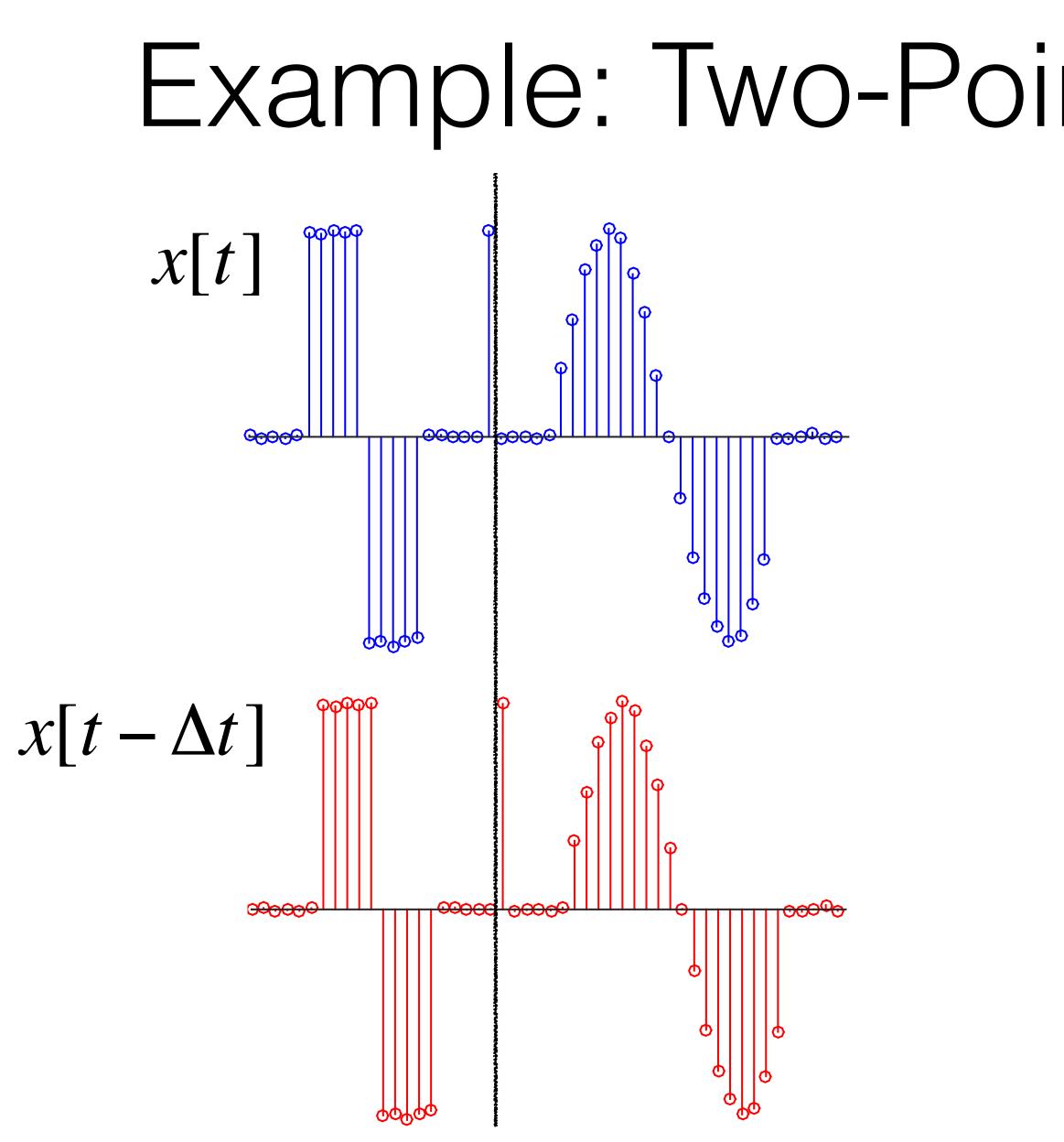
What to Expect:

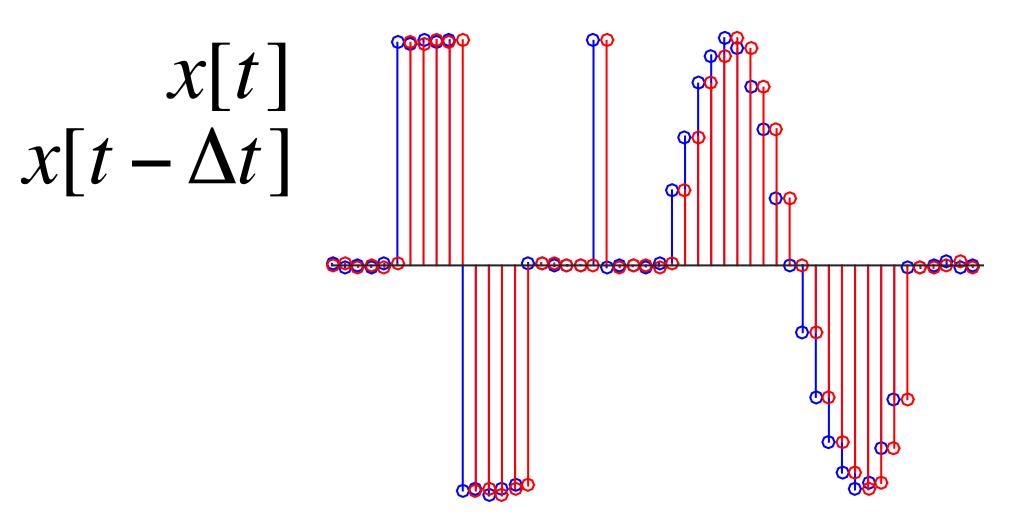
- Exaggerate differences
- Amplify quickly varying signals
- Attenuate slowly varying signals
- High Pass Filter?

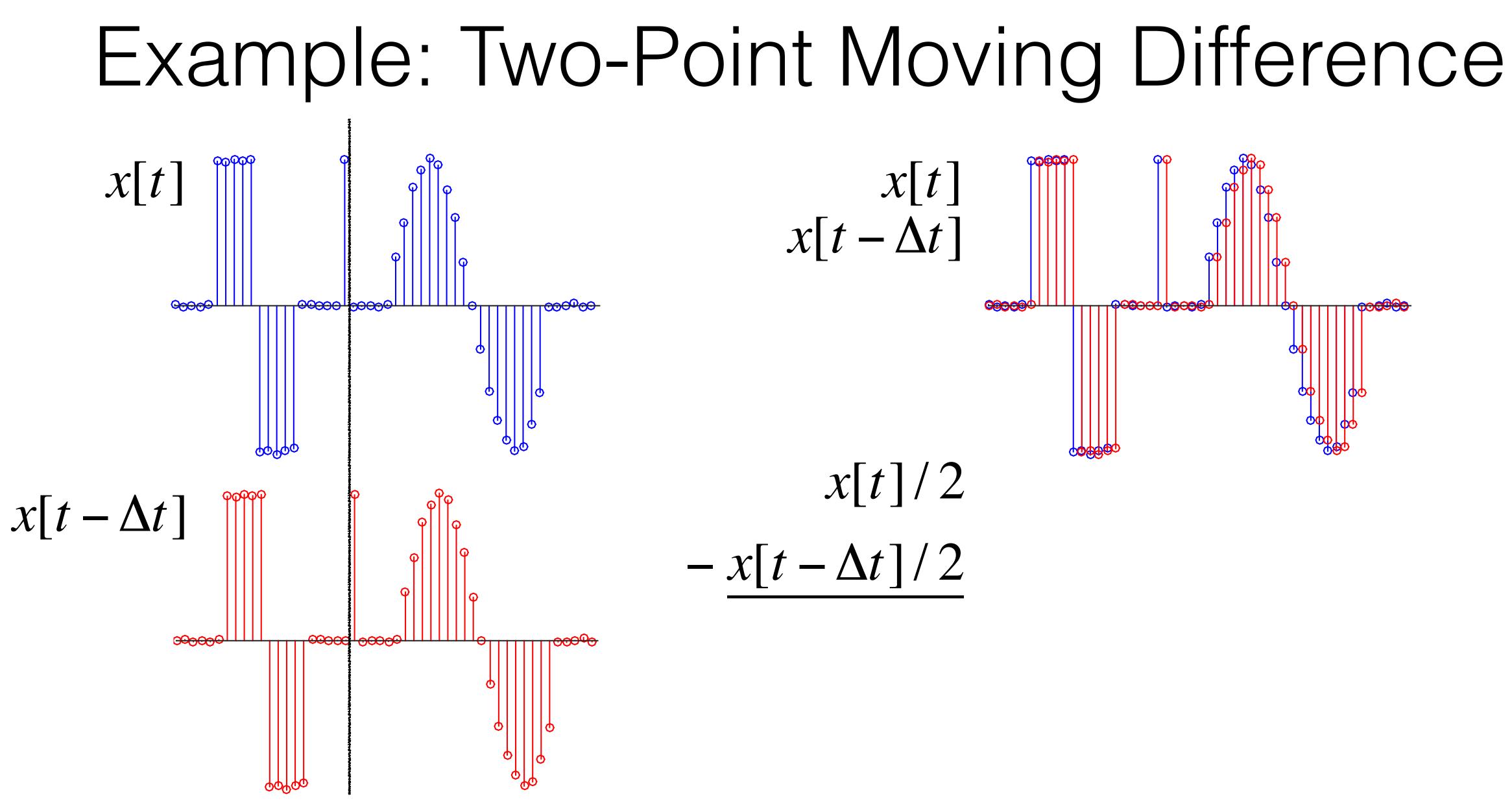
$y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

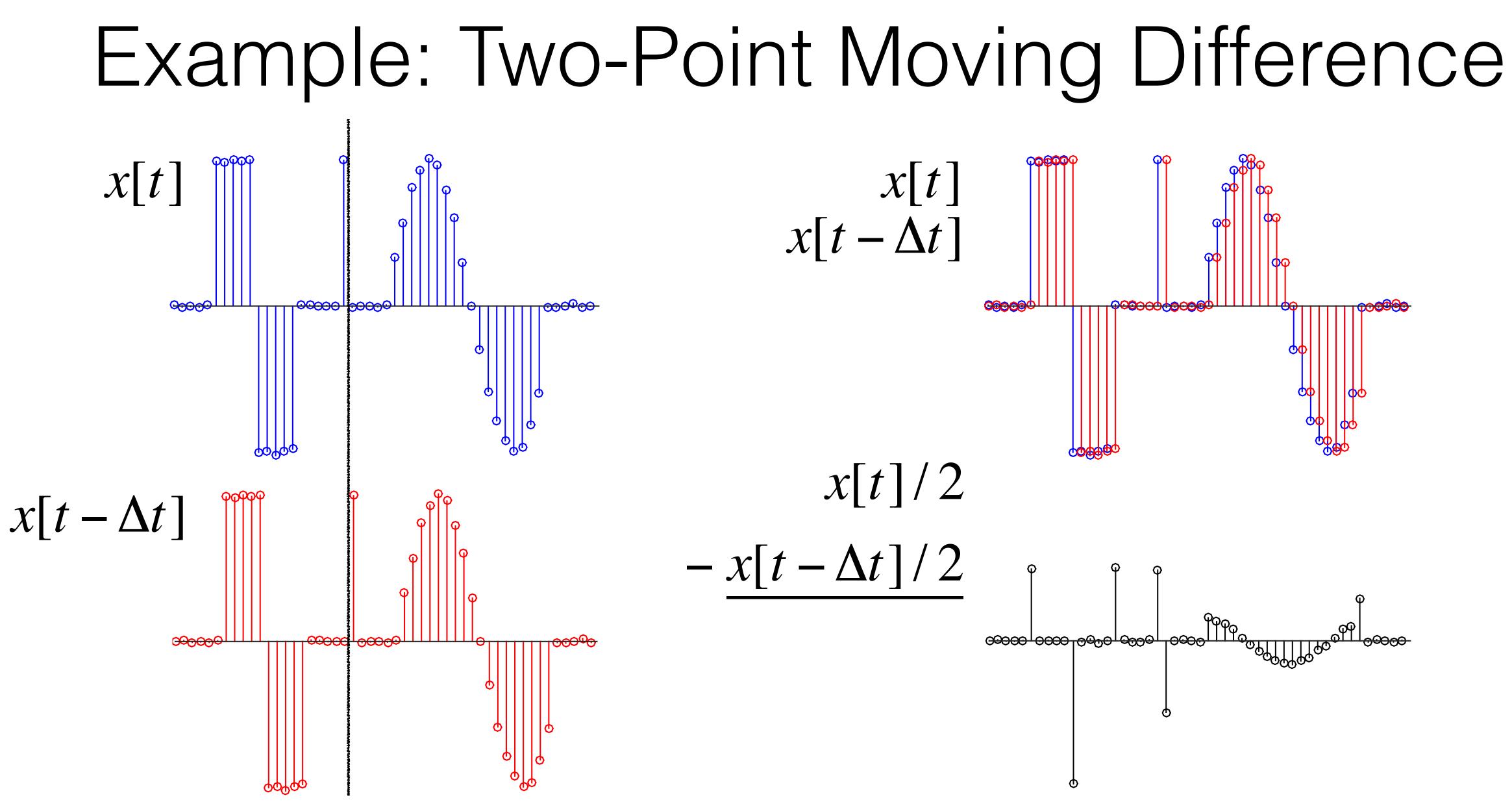




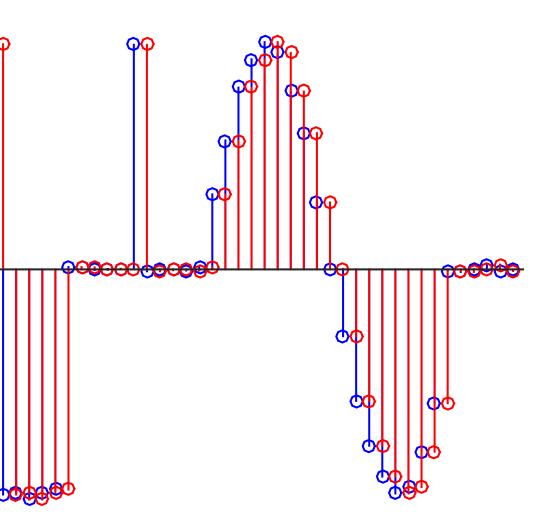


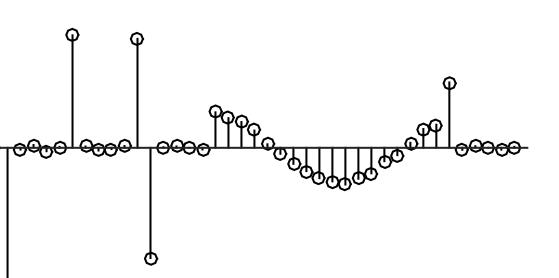




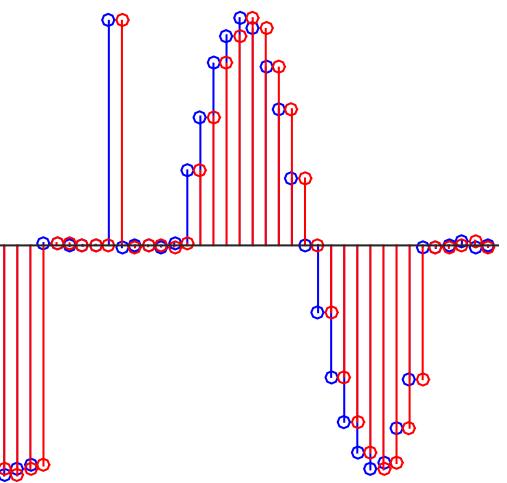


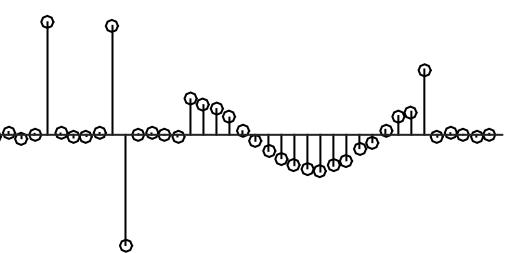
 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$



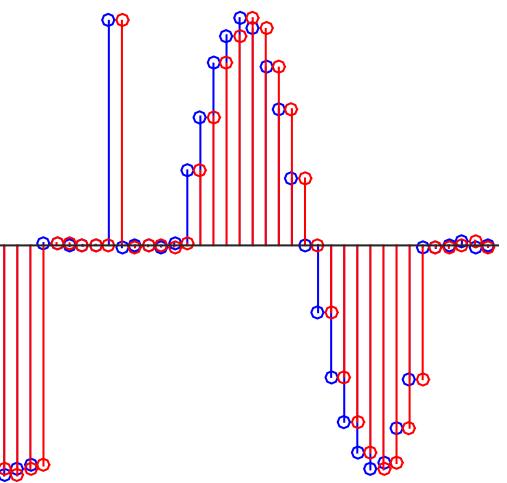


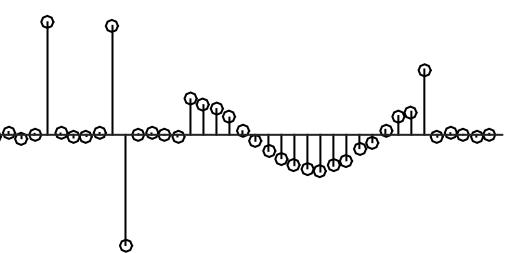
Example: Two-Point Moving Difference $x[t] \\ x[t - \Delta t]$

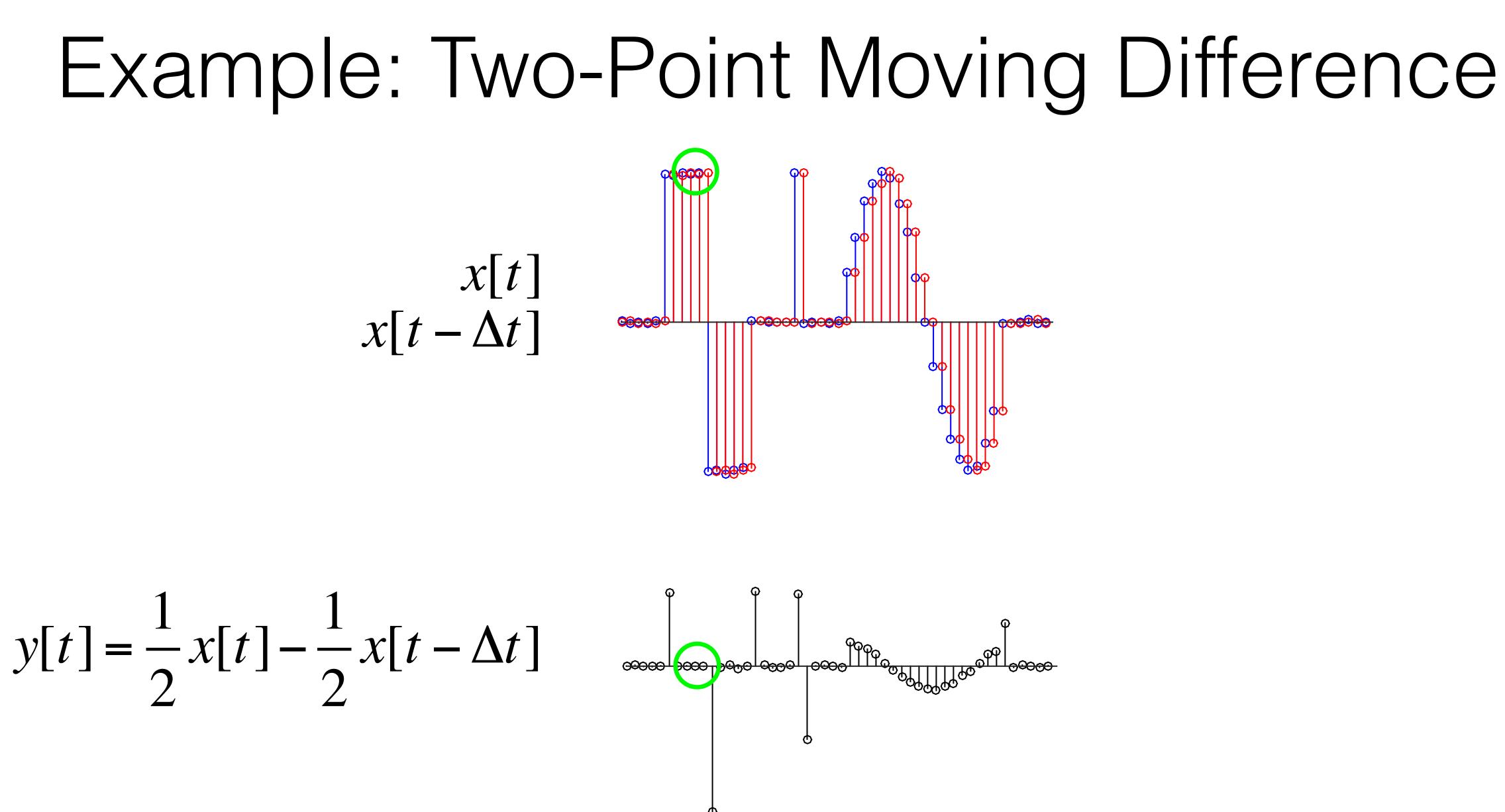




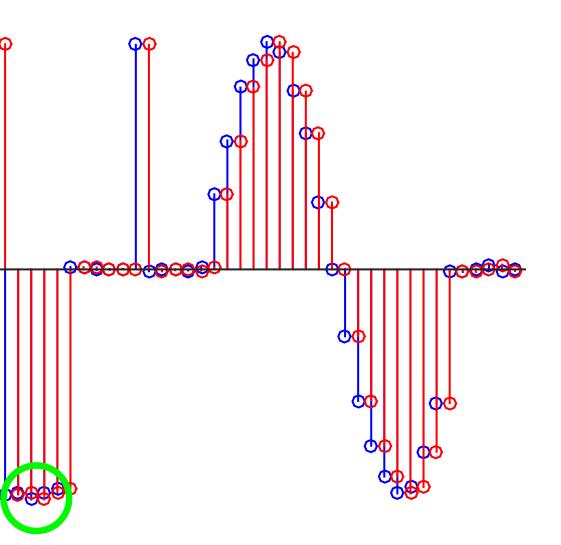
Example: Two-Point Moving Difference x[t] $x[t - \Delta t]$

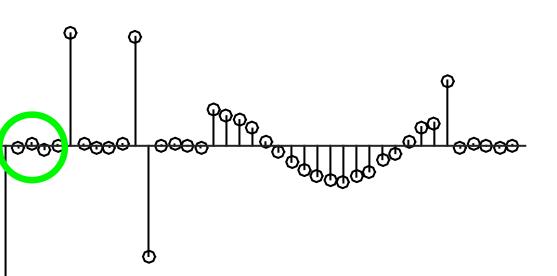




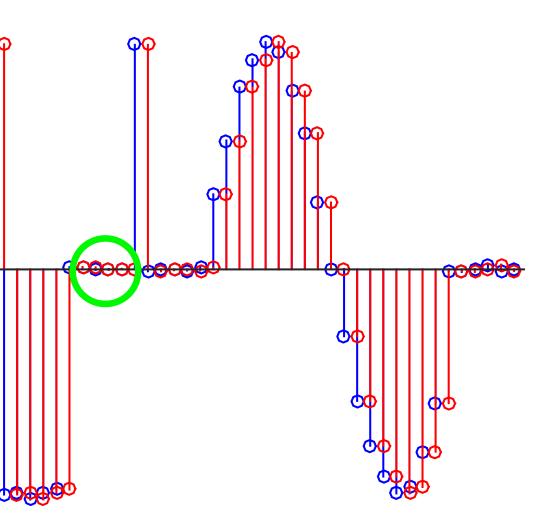


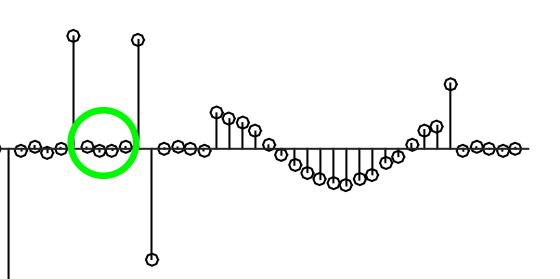
 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$



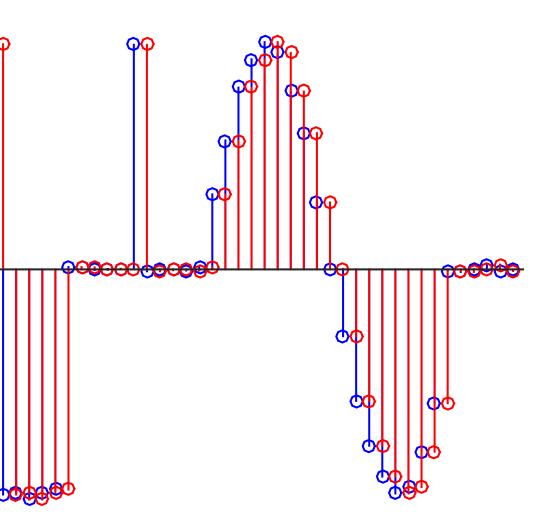


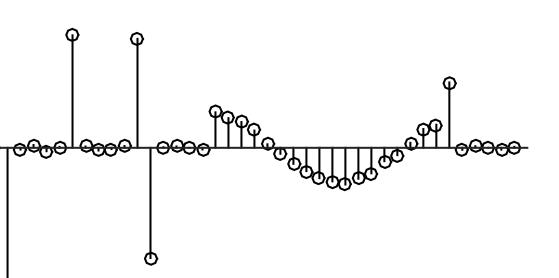
 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

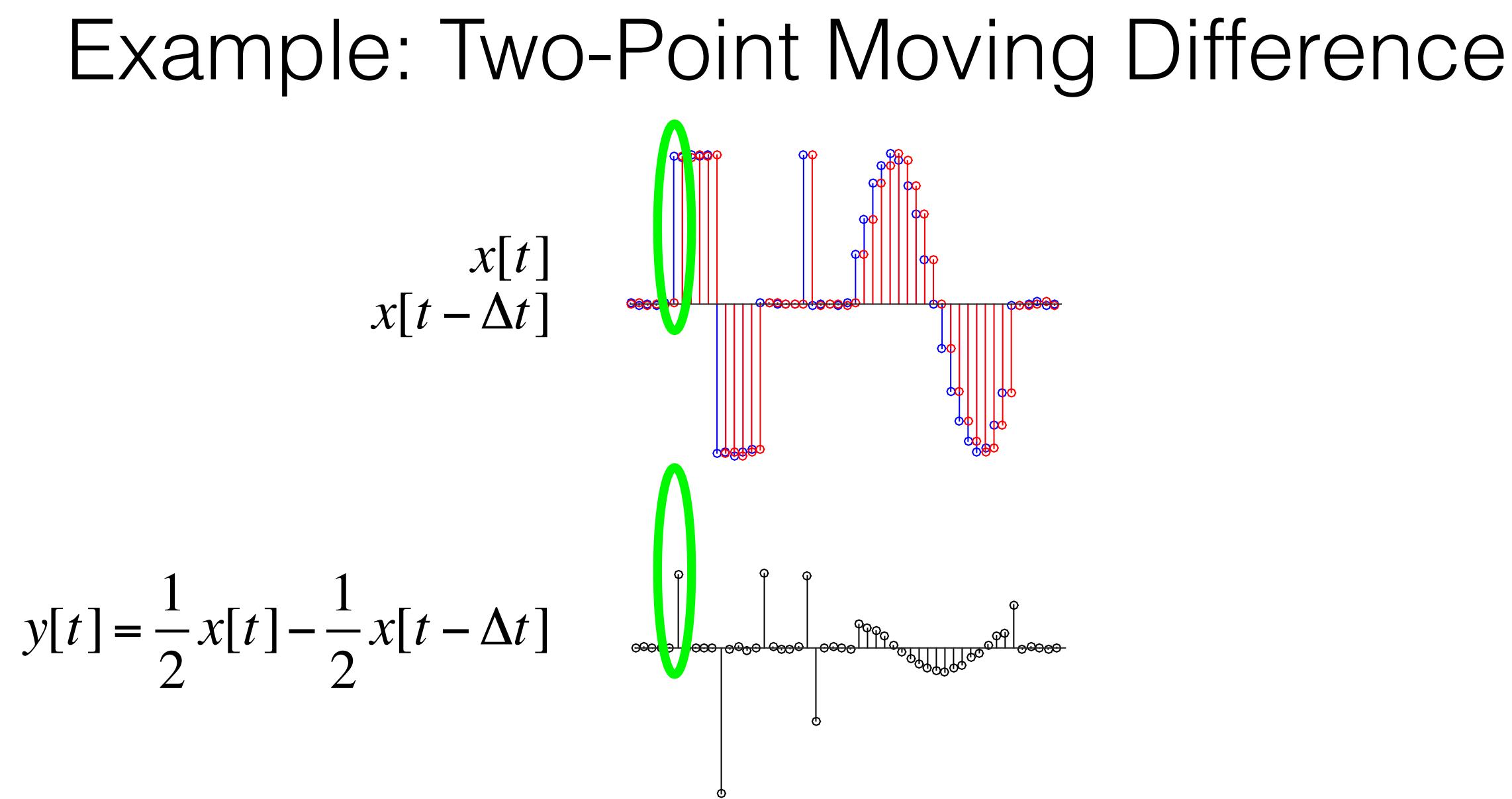


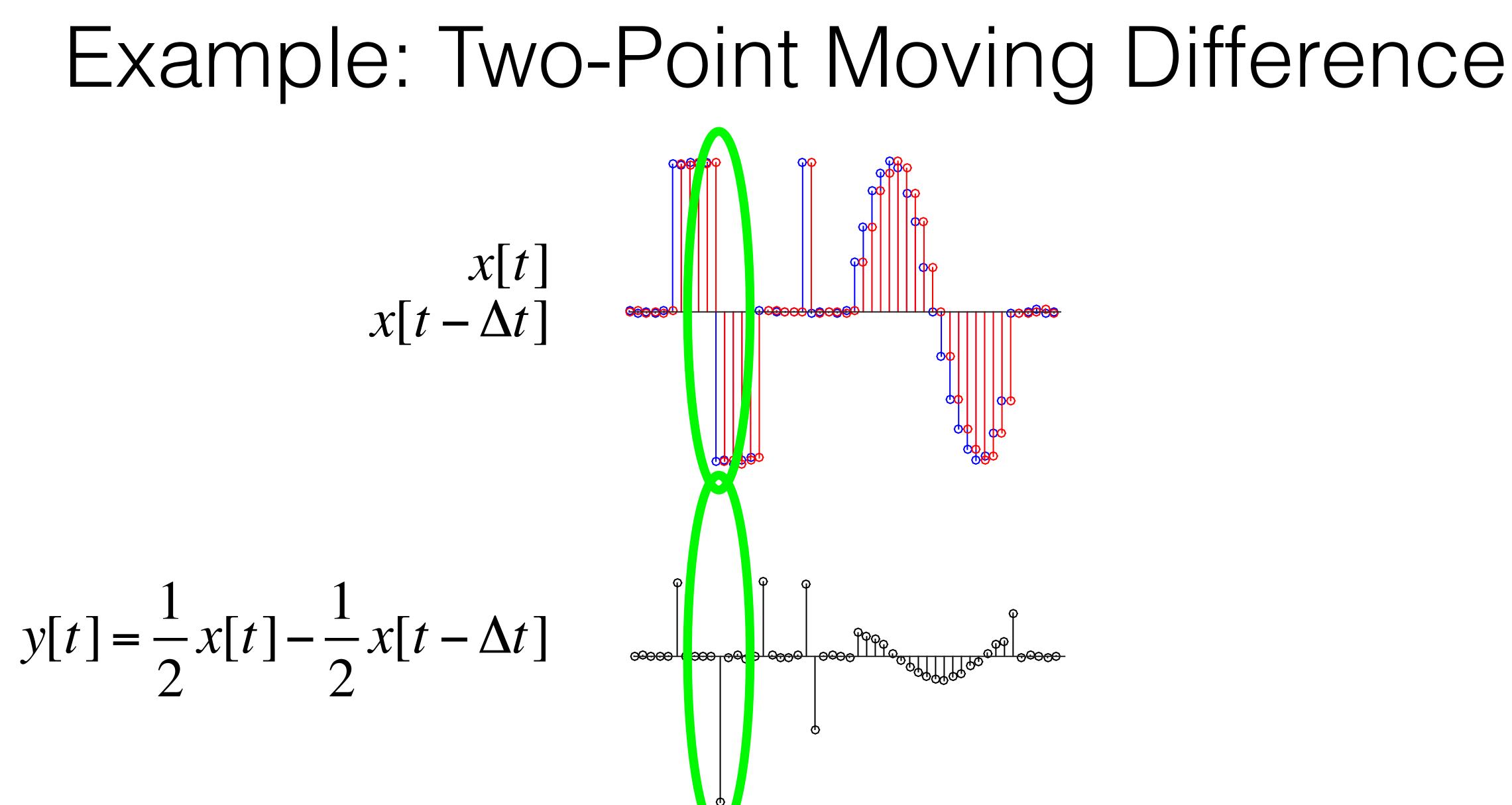


 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

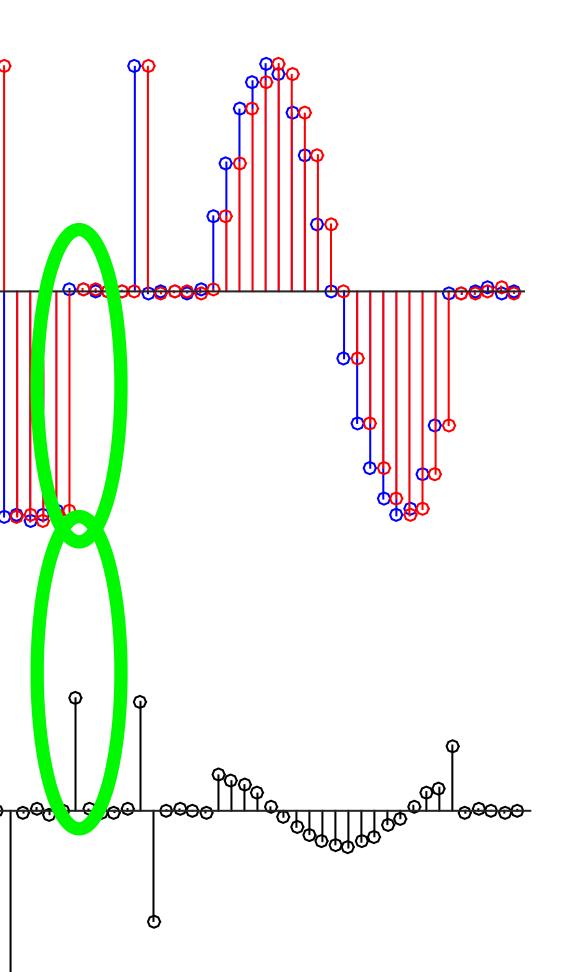




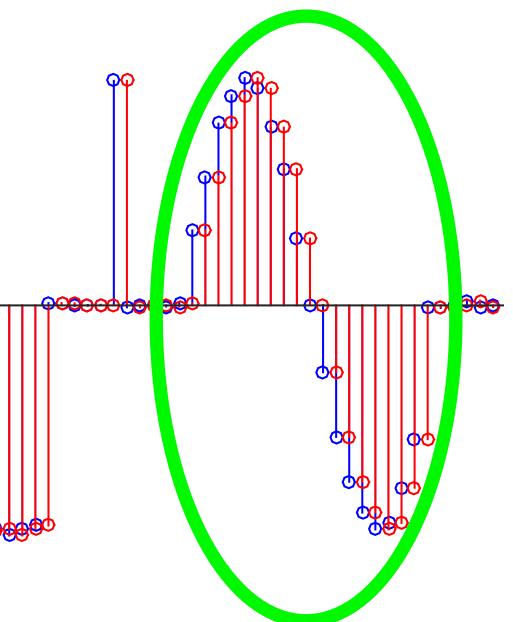


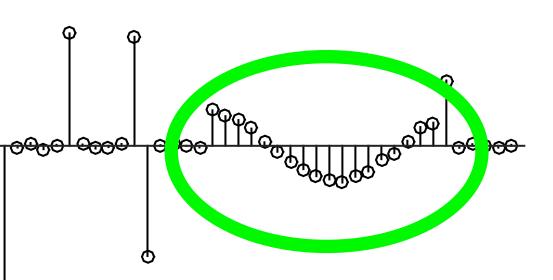


 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

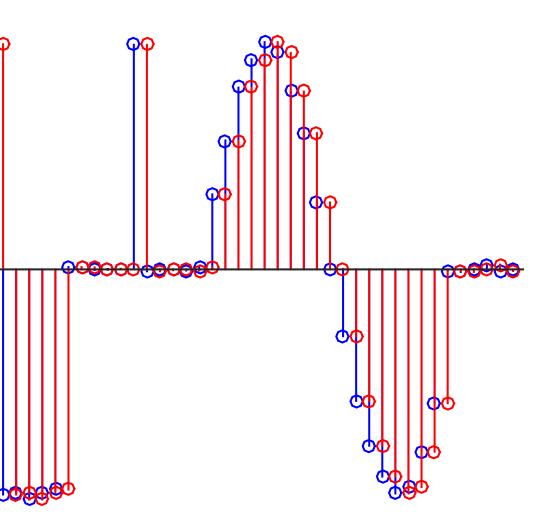


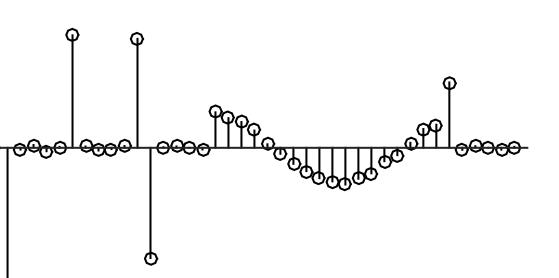
 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

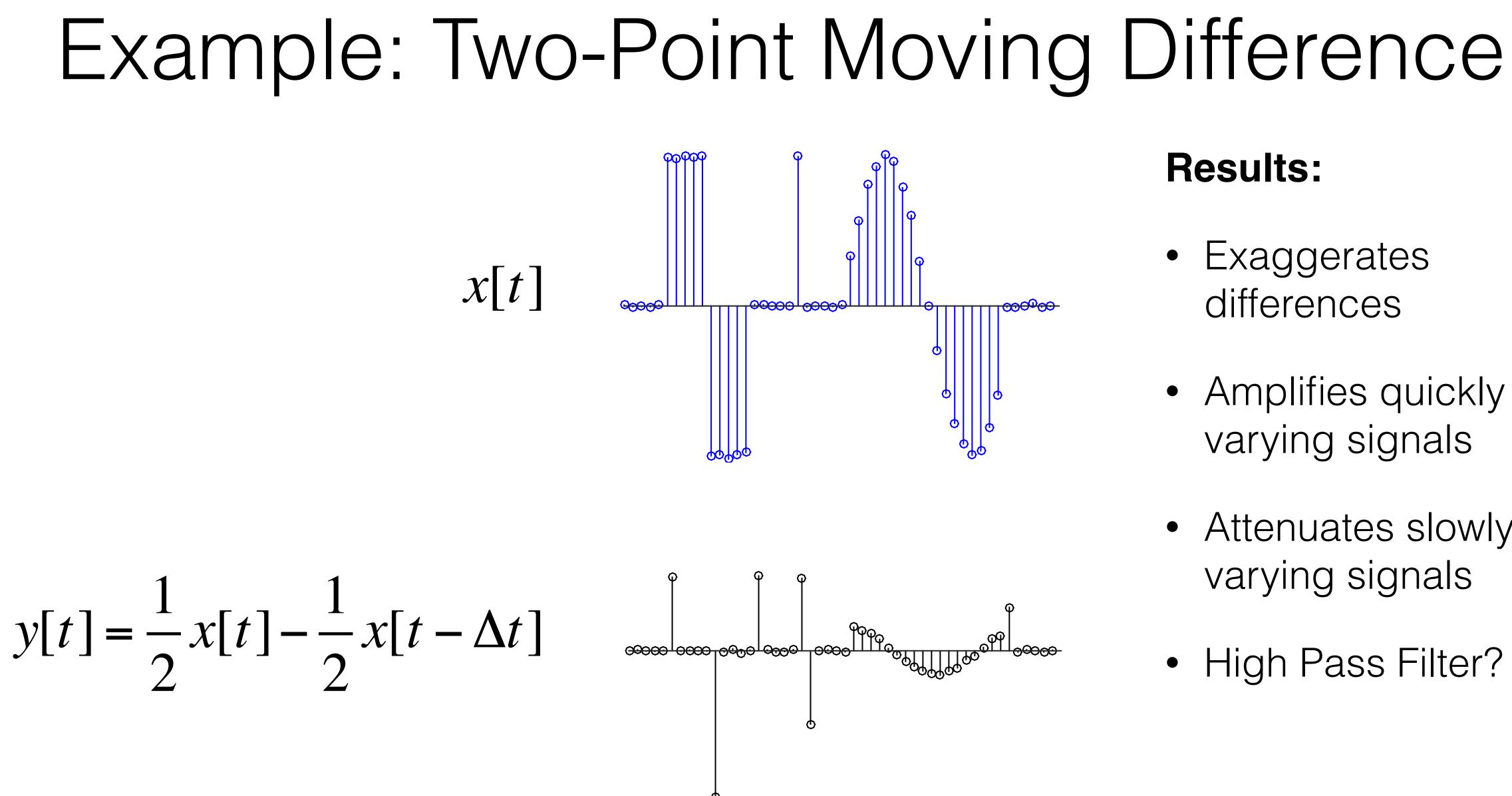




 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

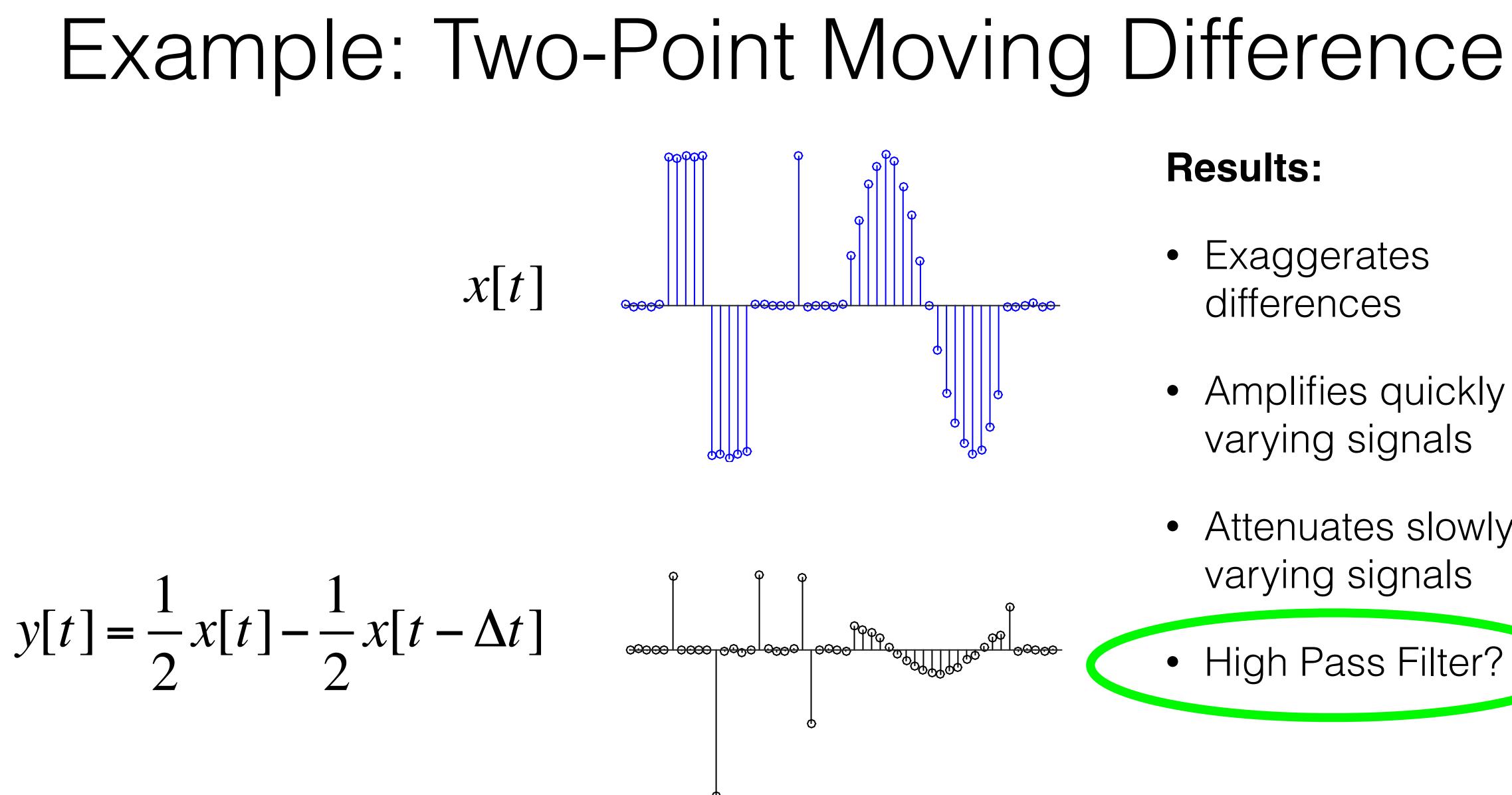






Results:

- Exaggerates differences
- Amplifies quickly varying signals
- Attenuates slowly varying signals
- High Pass Filter?

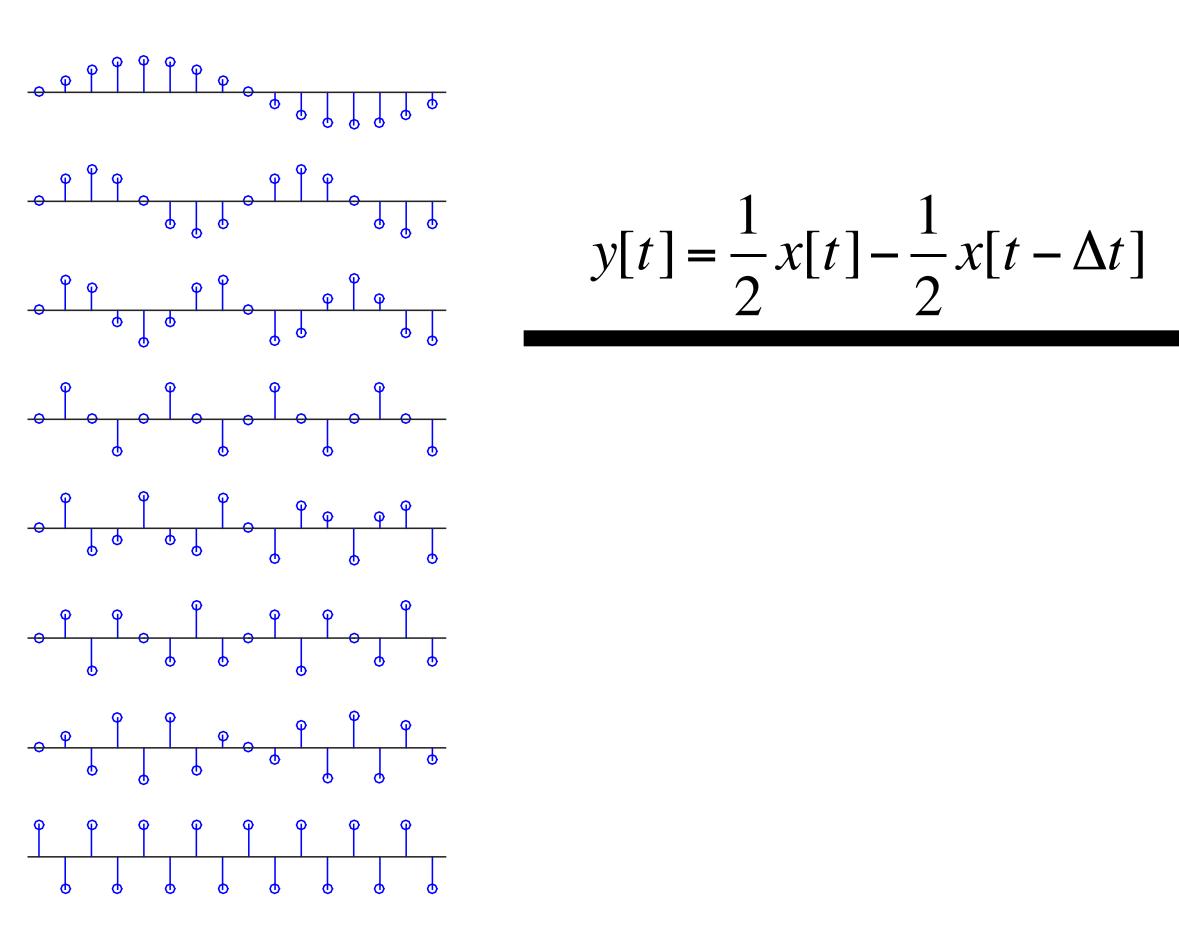


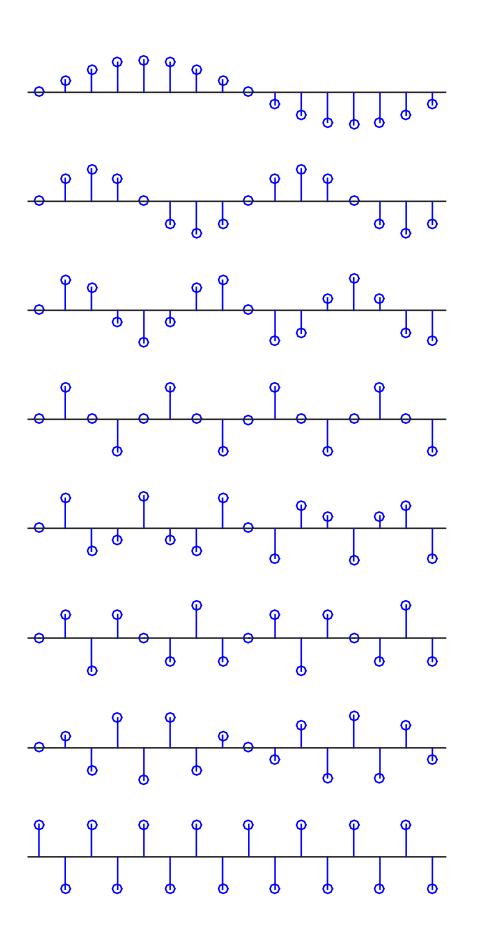
Results:

- Exaggerates differences
- Amplifies quickly varying signals
- Attenuates slowly varying signals

High Pass Filter?

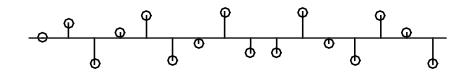


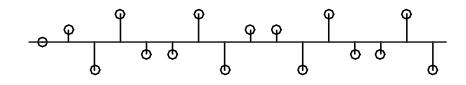


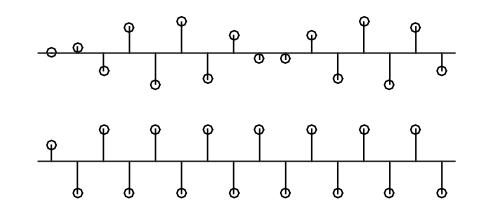


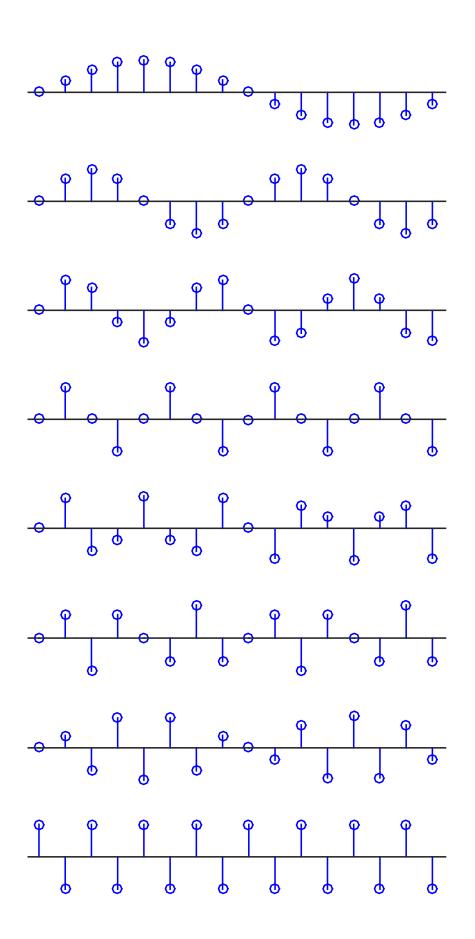
Filters and the Fourier Transform

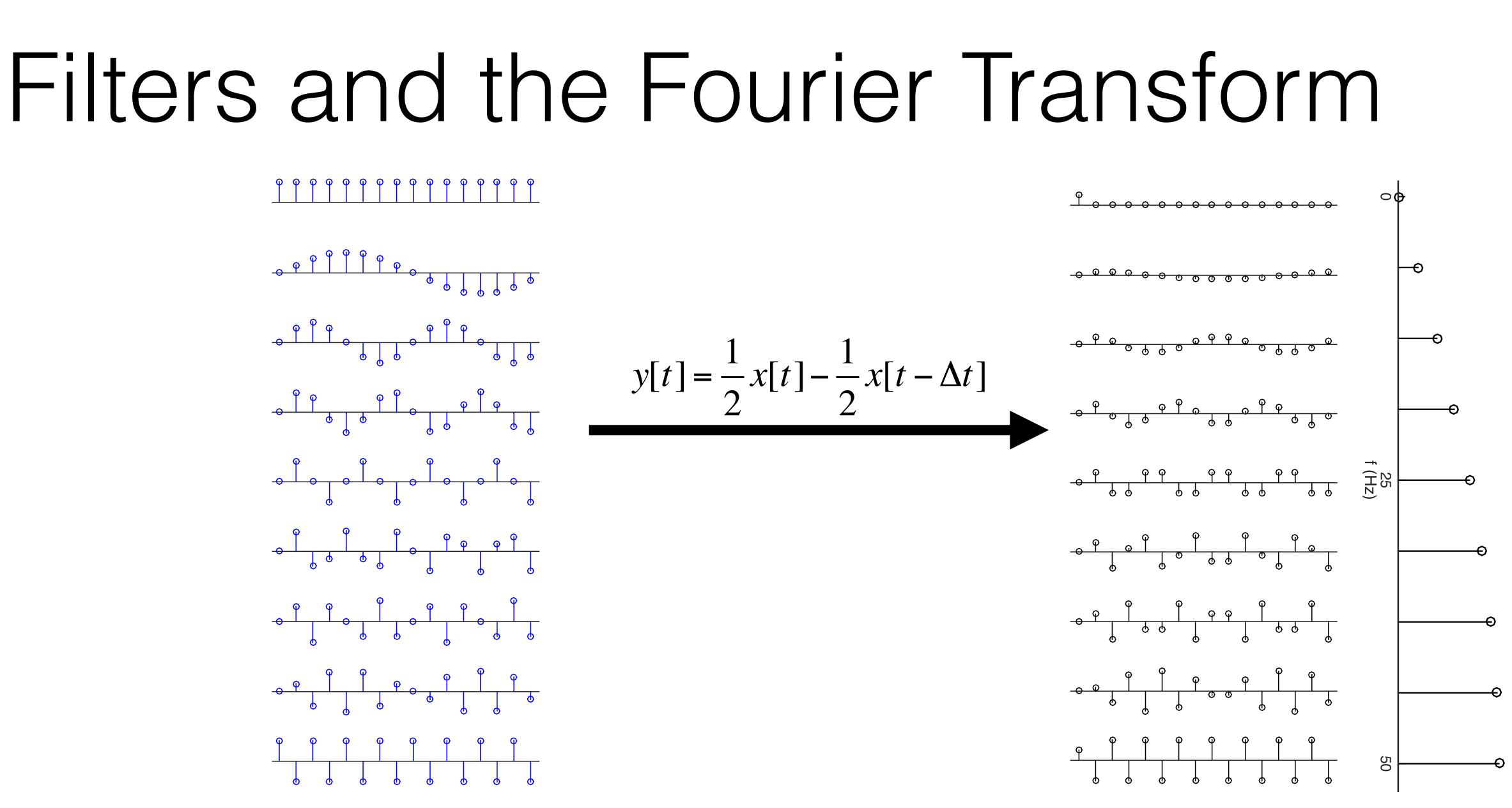
 $y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]$

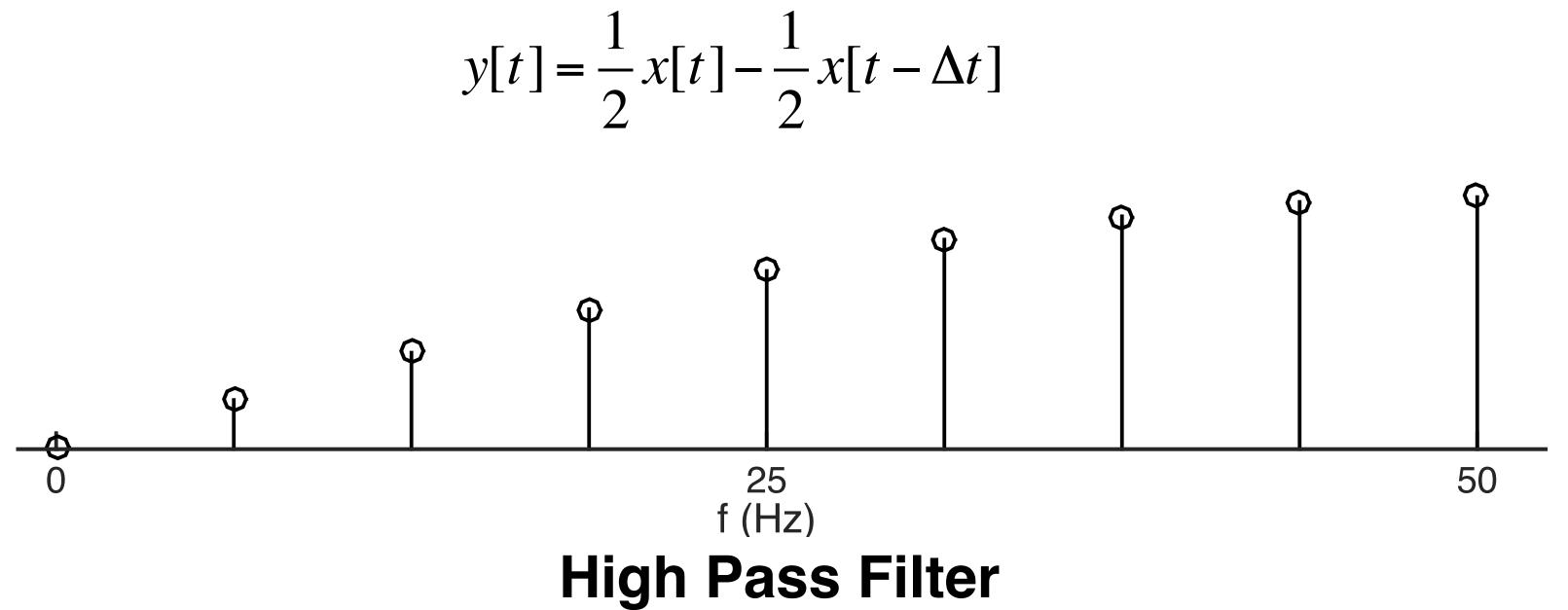






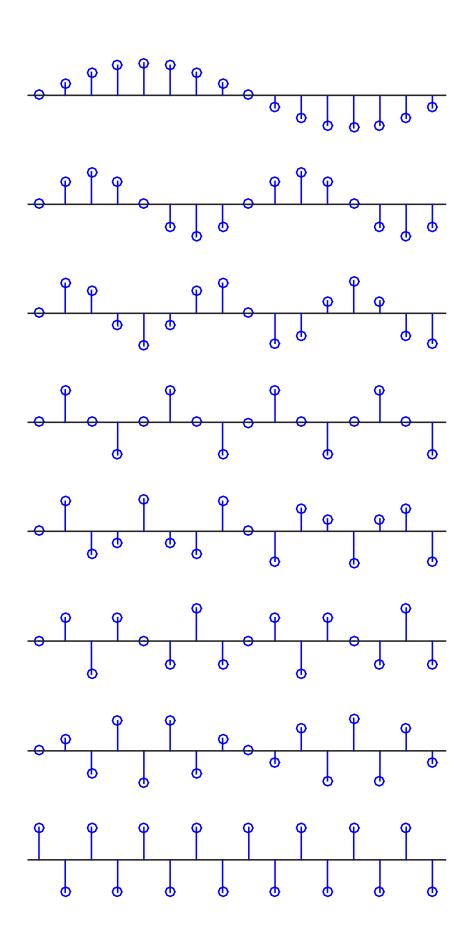






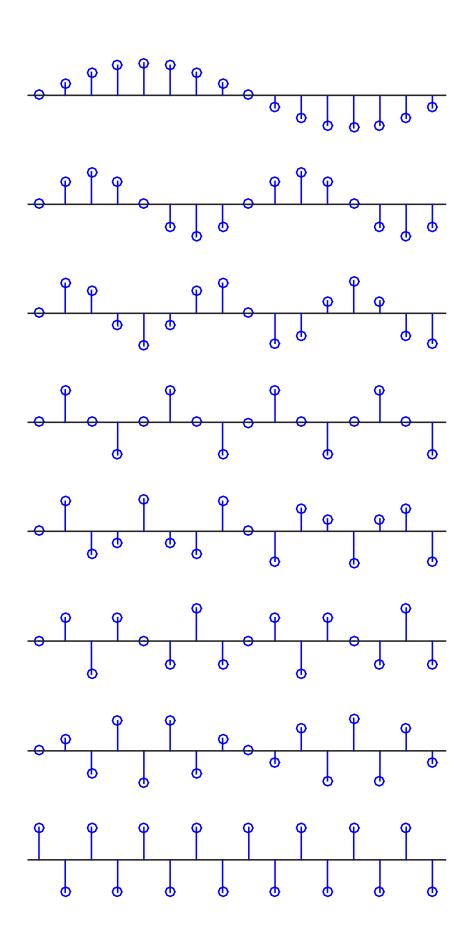
Filters and the Fourier Transform

• • • • • • • • • • • • • • • • •



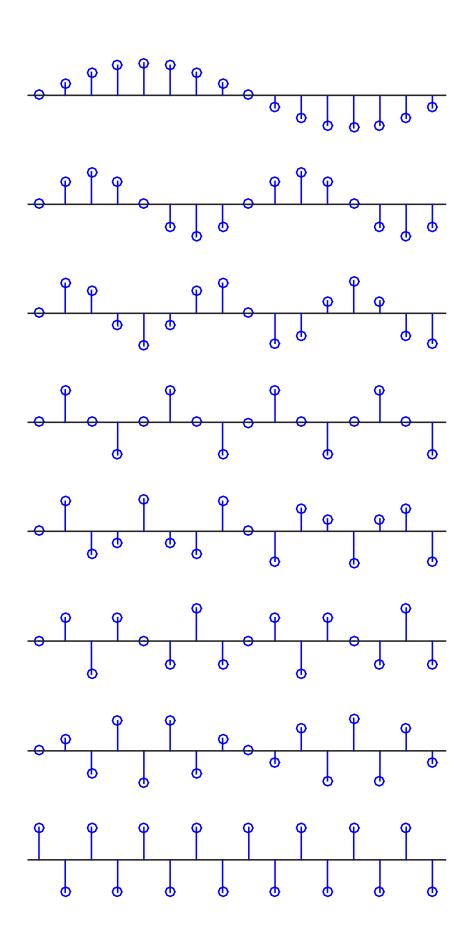
Filters and the Fourier Transform

 $y[t] = \frac{1}{10}x[t] - \frac{9}{10}y[t - \Delta t]$



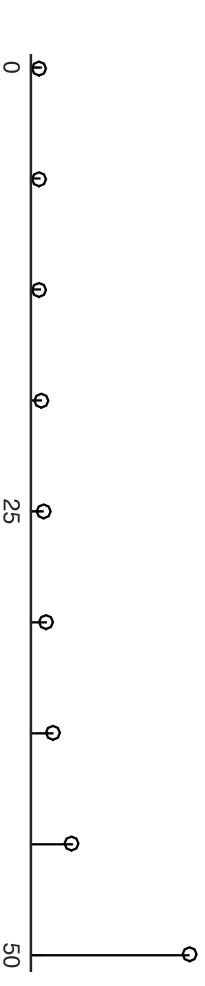
Filters and the Fourier Transform

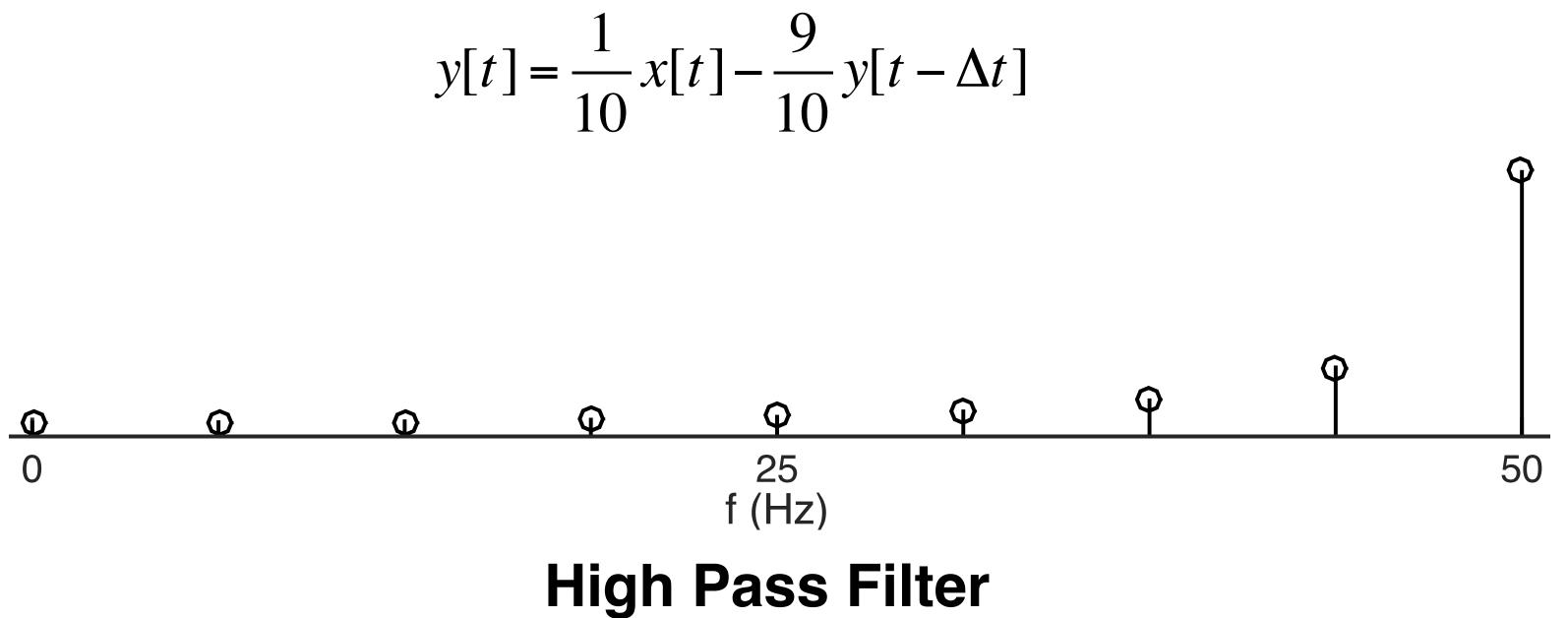
 $y[t] = \frac{1}{10} x[t] - \frac{9}{10} y[t - \Delta t]$



Filters and the Fourier Transform

 $y[t] = \frac{1}{10} x[t] - \frac{9}{10} y[t - \Delta t]$ (Н_z 25 Ю -Ð





Filters and the Fourier Transform

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

"Which Filter Should I Use?"

-Every student I've ever worked with

Many Filter Decisions

- Frequency Selectivity: Sharp vs. Soft Frequency Transition
- Feedforward Only/Feedback: FIR vs. IIR
- Filter Order: Low order vs. High Order
- Causality: Causal vs. non-Causal (e.g. "zero-phase" filters)
- and more (e.g., FIR: moving average vs. Parks-McClellan, IIR: Butterworth vs. elliptic)

Ideas to Keep in Mind

- Filters modify signals, by design.
- There is no such thing as a filter that leaves signals (or signal components) unaltered
- Most filter decisions involve considering a valid tradeoff
 - Don't go overboard one way or the other
- Some filter decisions allow one to avoid artifacts without any tradeoff

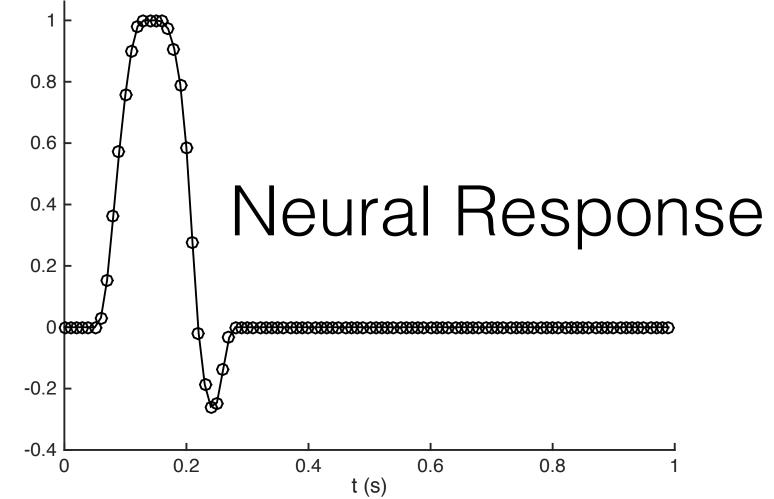
Frequency Selectivity/Transitions • Time and Frequency are inextricably linked.

- - content of the signal.
 - Low-Pass Filters will lengthen fast temporal changes
 - another
 - "ringing".

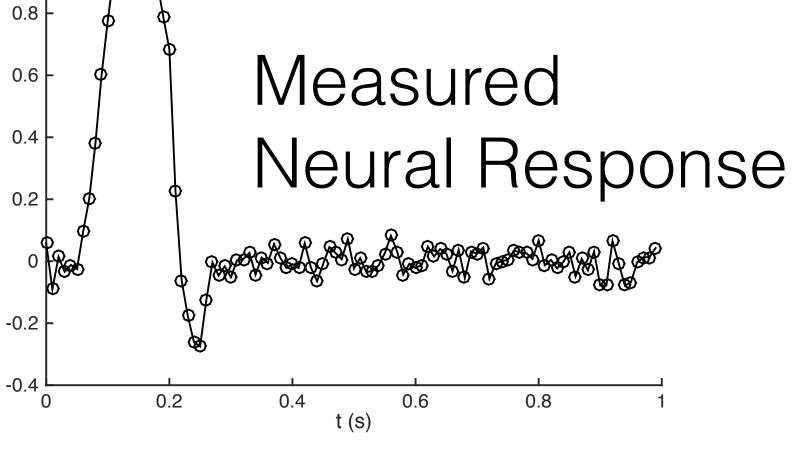
• Changing the frequency content of a signal will change the temporal

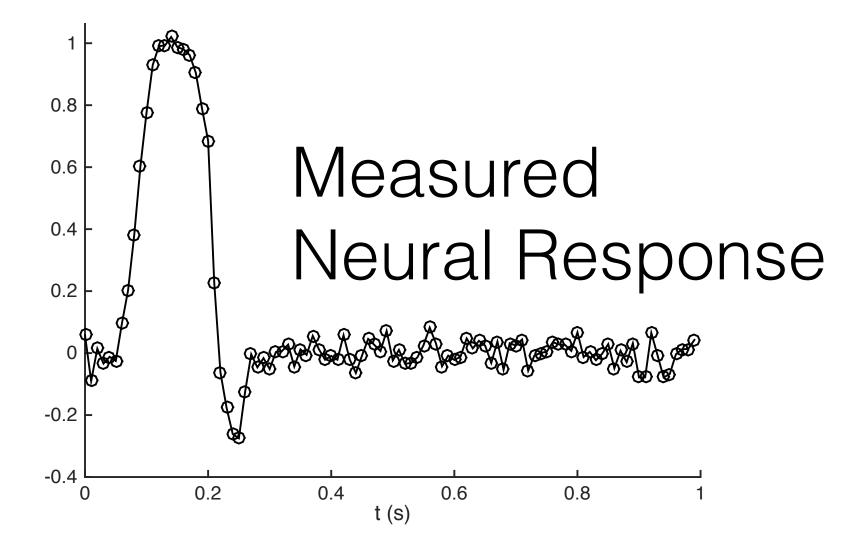
High-Pass Filters will remove slow transitions from one baseline to

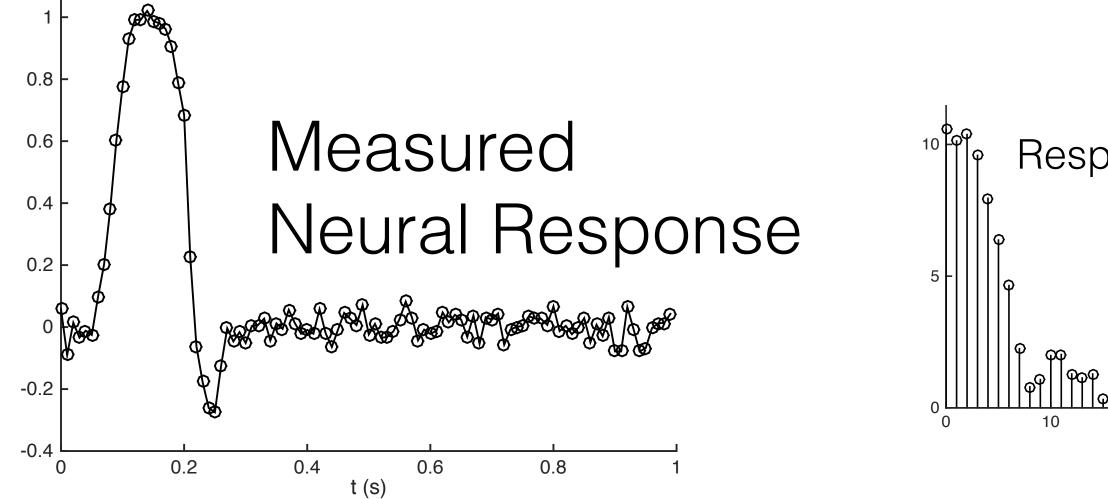
Sharp frequency transitions produce artificial temporal elongation:



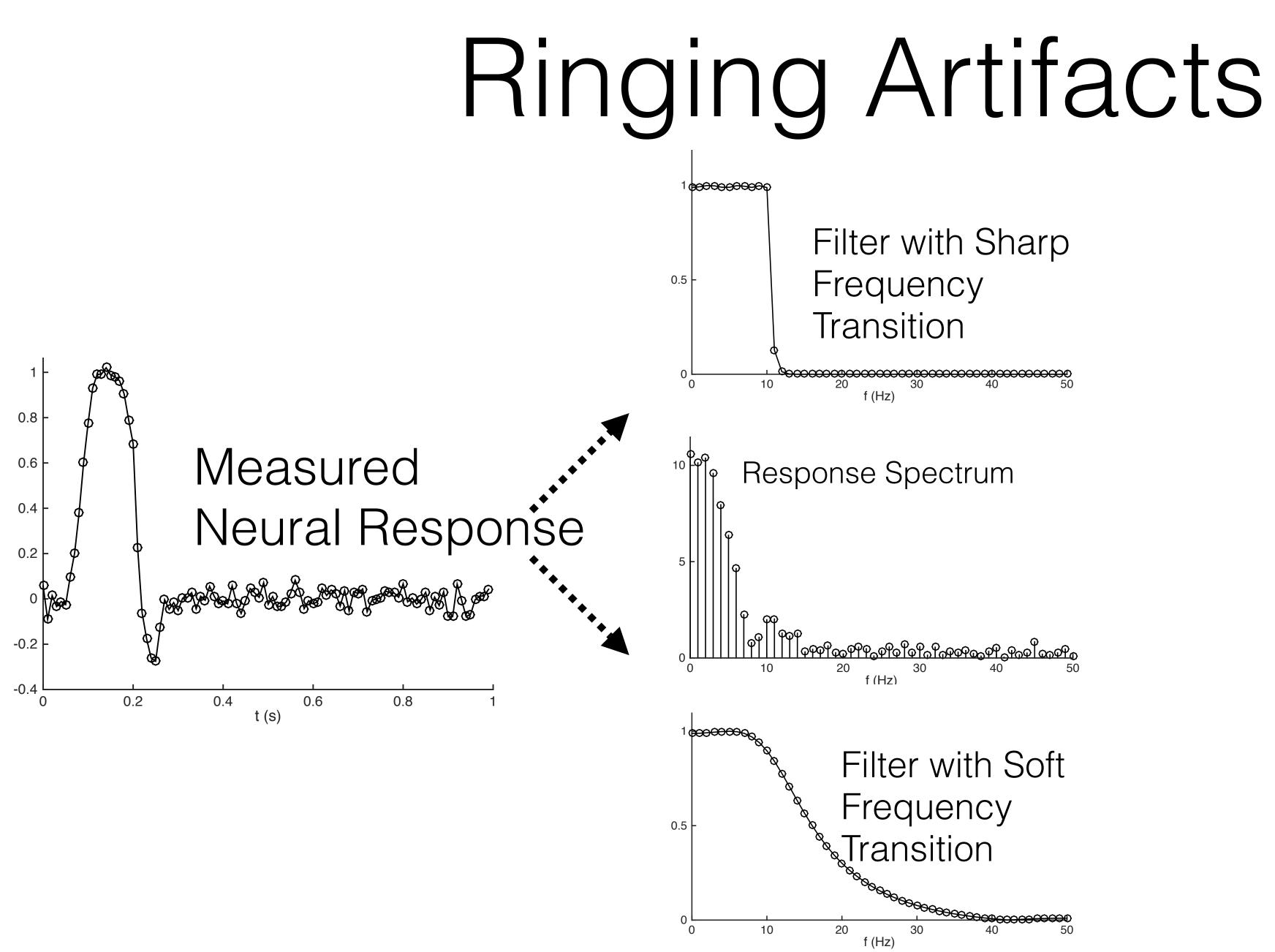
Ringing Artifacts 1 0.8 0.6 Neural Response 0.4 0.2 -0.2 -0.4 0.2 0.4 0.6 0.8 0 t (s) 1

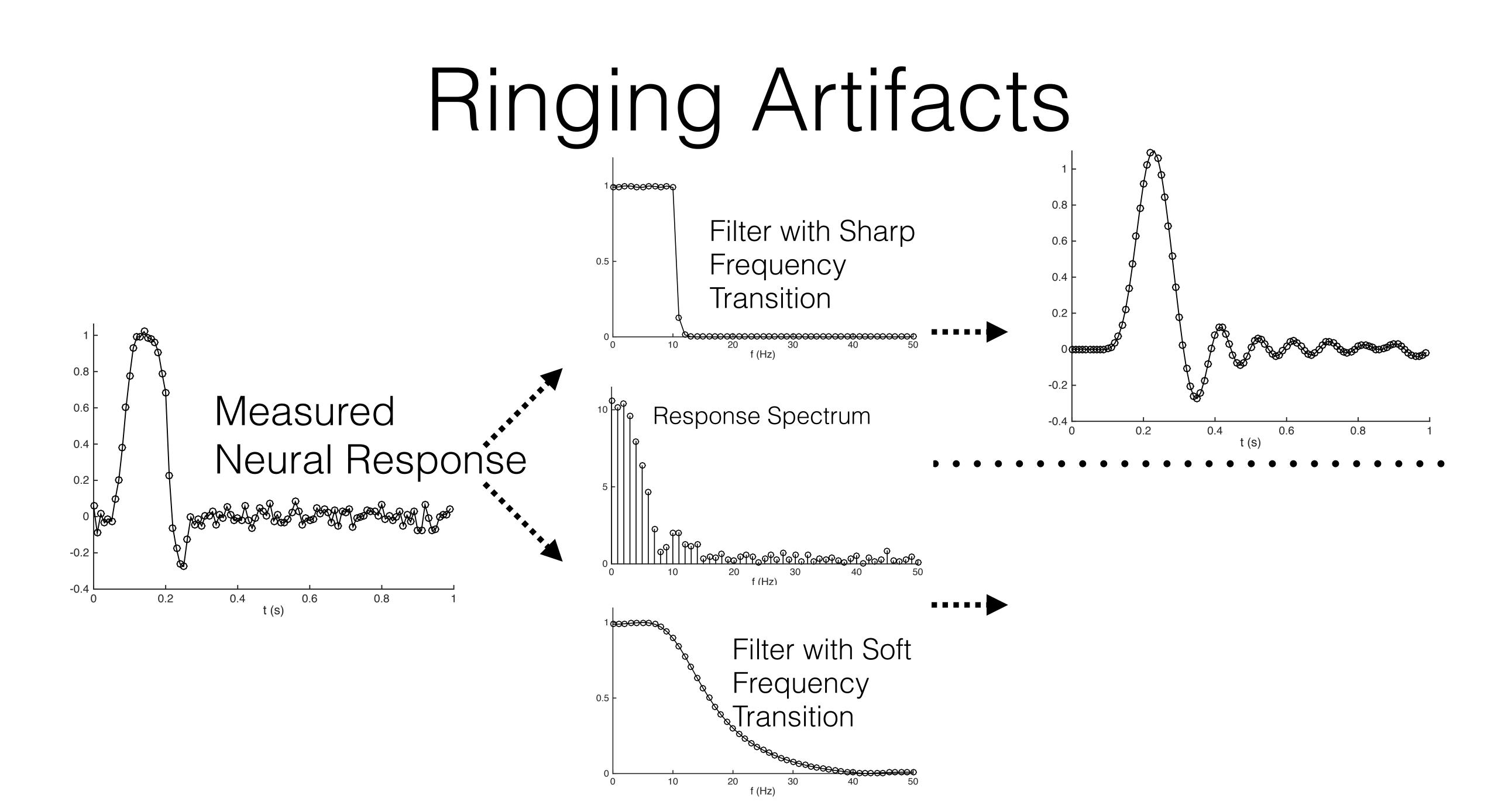


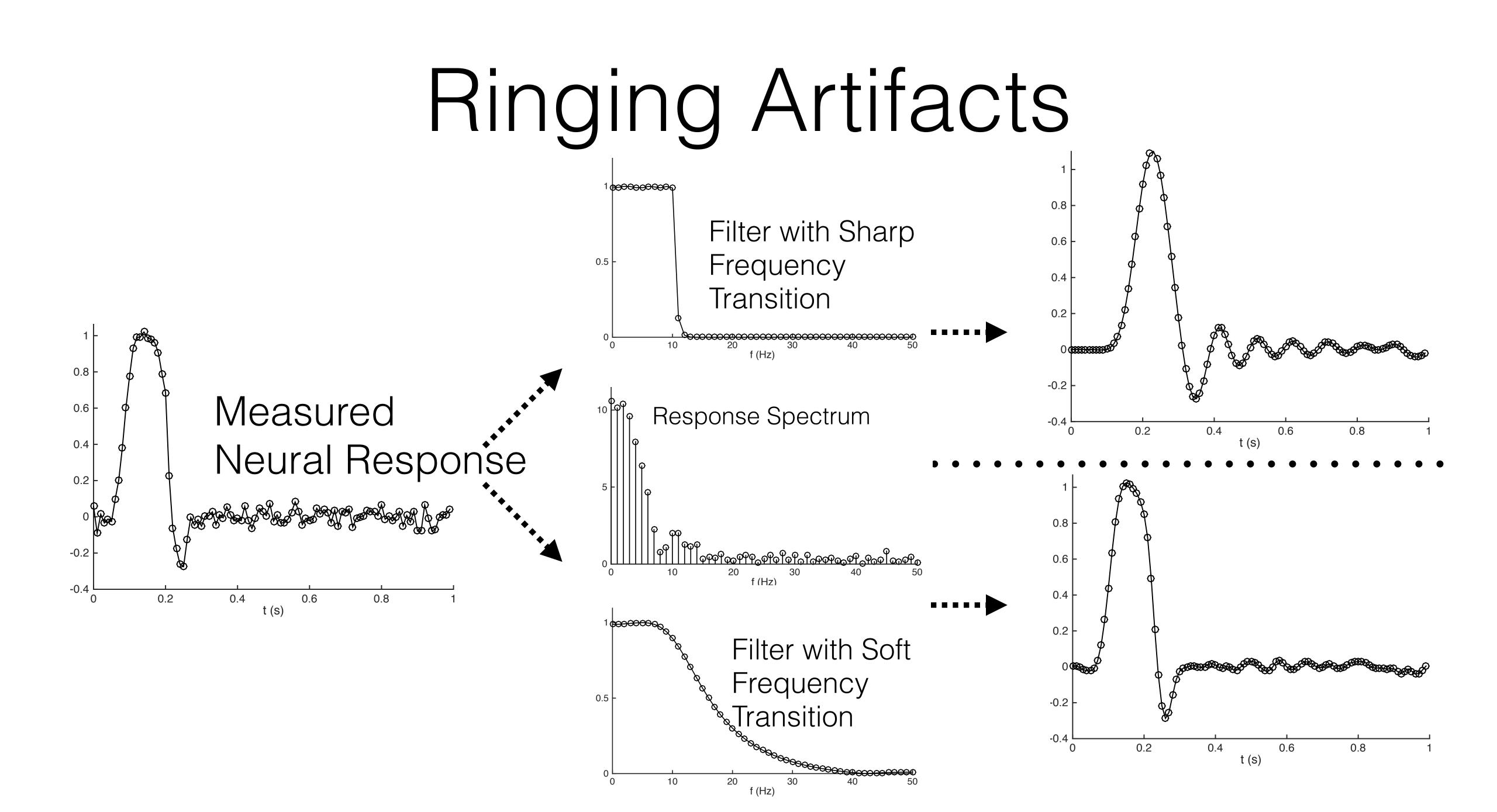




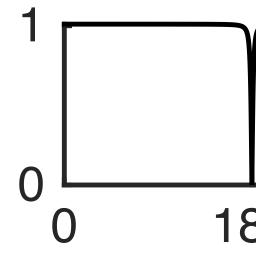
Response Spectrum





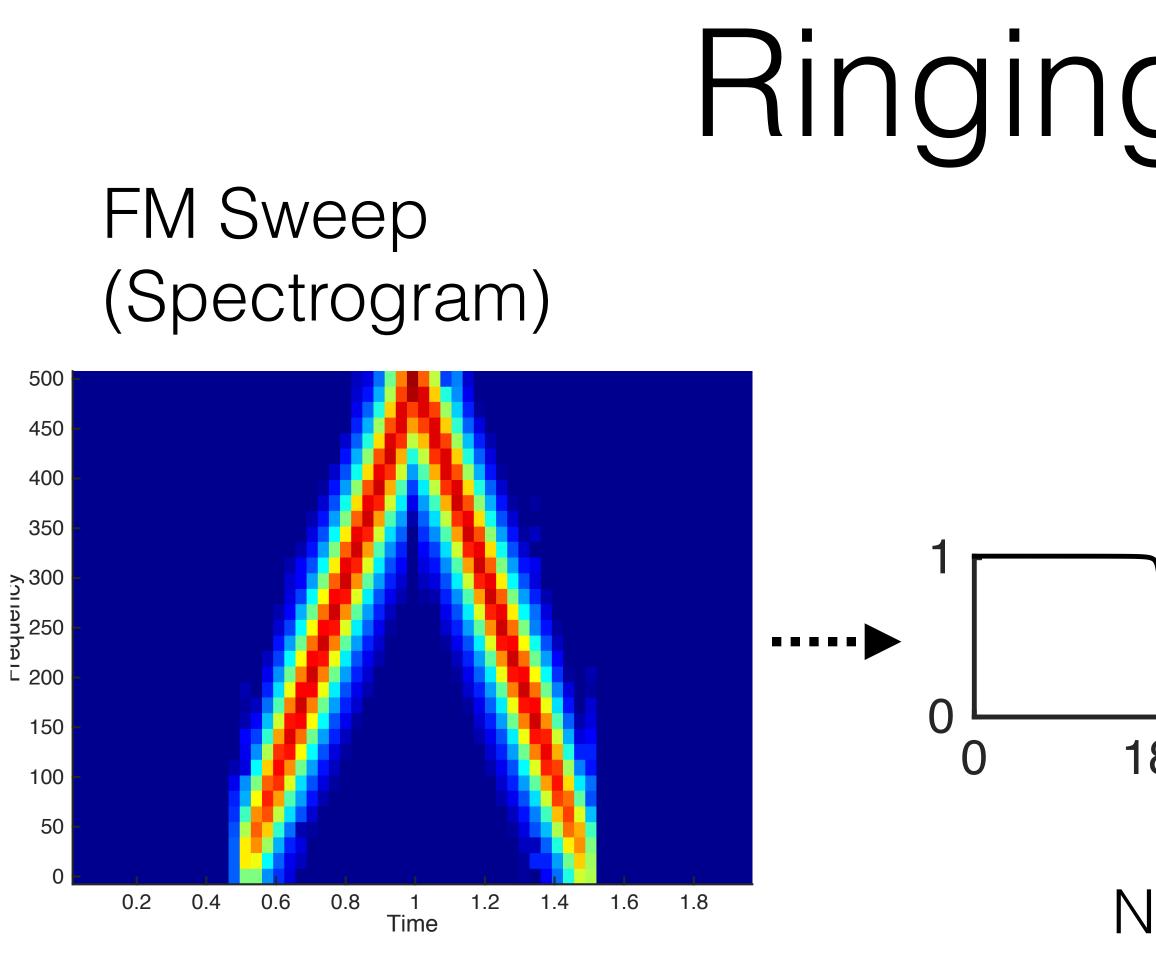


- Sharp Frequency Transitions are sometimes Necessary
 - e.g., Notch filters (and related filters, such as Comb filters)
- In these cases there will be unavoidable ringing



Notch Filter (Sharp Frequency Transition)

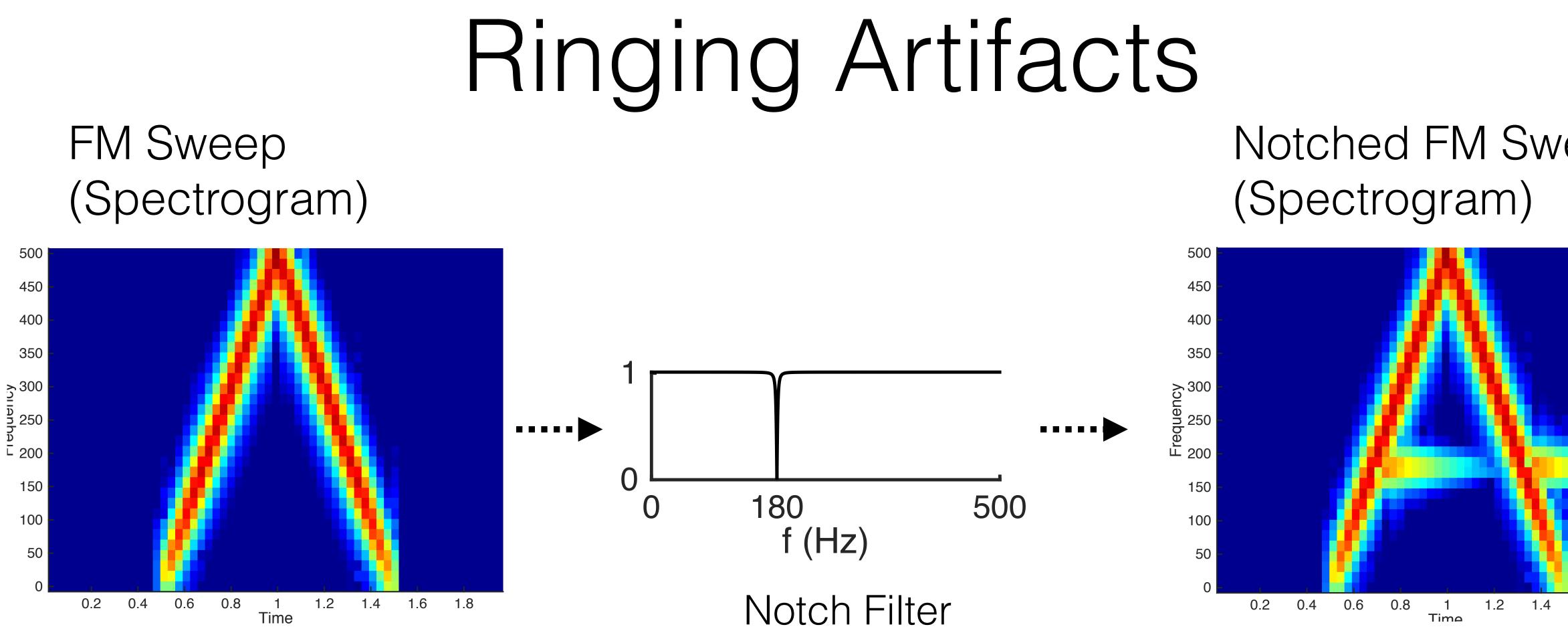
180 500 f (Hz)



Notch Filter (Sharp Frequency Transition)

Ringing Artifacts

180 500 f (Hz)



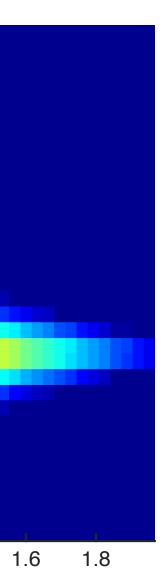
(Sharp Frequency Transition)

Notched FM Sweep

Notch too brief to see But ringing clear:

- narrowband
- extended in time



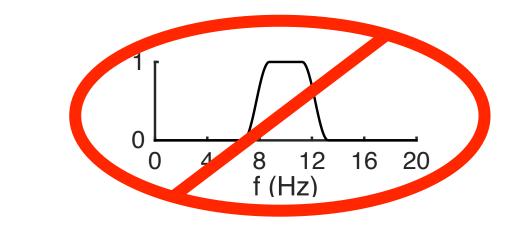


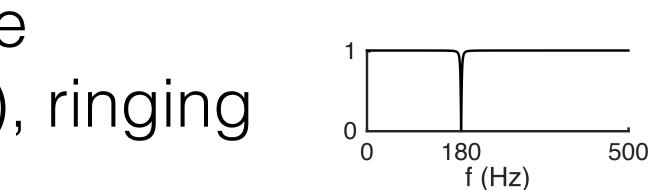


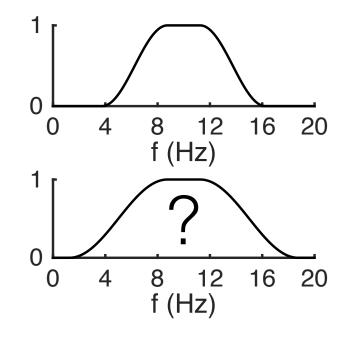


Take care, but don't overreact

- Avoid Ringing by avoiding sharp frequency transitions
- If sharp frequency transitions are necessary (as for notch filtering), ringing may follow
- Don't overly soften frequency transitions or you'll lose frequency selectivity







- FIR (finite impulse response): Feedforward only
 - Examples: Moving Average (*avoid, in general*), Parks-McClellan ("Optimal"), others
- IIR (infinite impulse response): Feedback also incorporated
 - Instability a potential issue
 - Examples: Butterworth (*not awful, but not great*), Chebyshev, Elliptic (very good), others

FIR VS. IIR

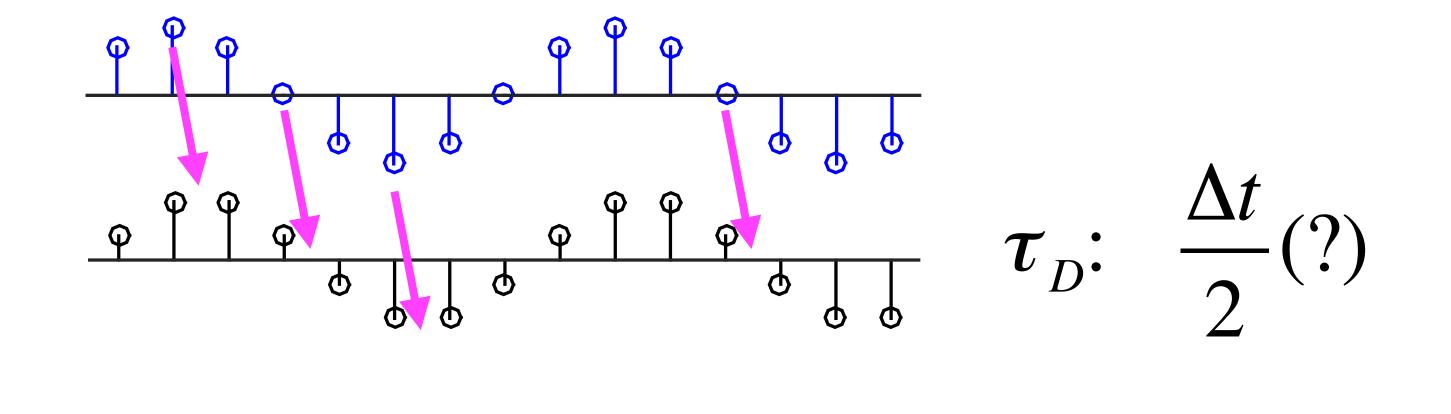
FIR vs. IIR: How to choose?

- No universal answer. It may depend on:
 - group delay (signal delay intrinsic to filter): group delay value and group delay frequency dependence
 - signal loss due to filter startup (dependence on signal values before signal starts)
 - stability concerns (if IIR filter)
 - more...

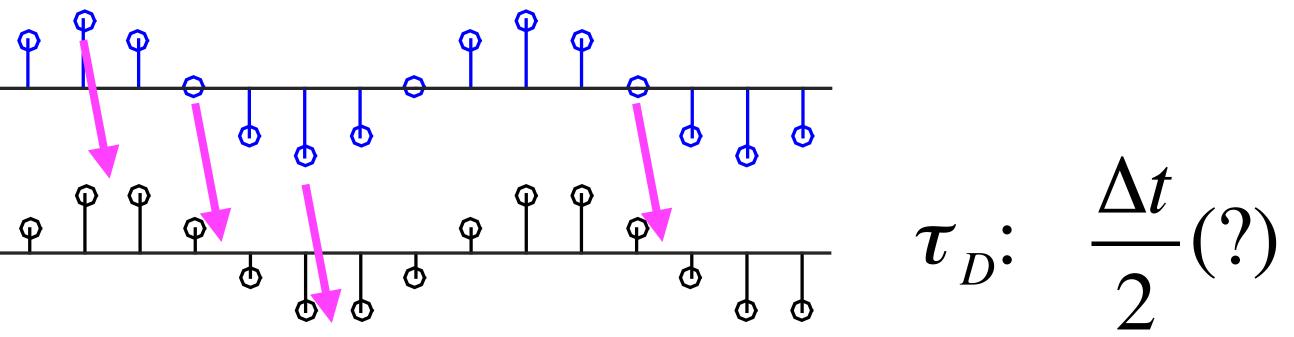
Group Delay

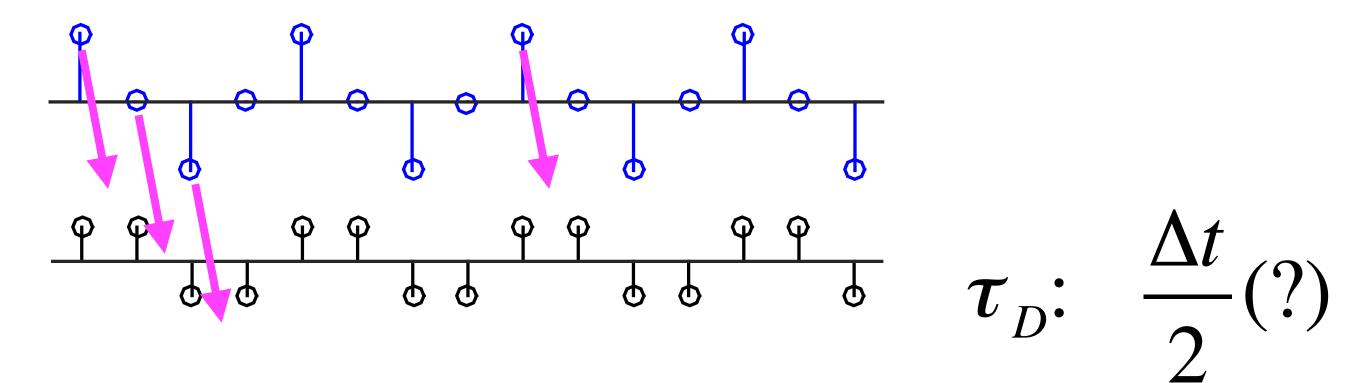
- Intrinsic to filtering—cannot be removed
- Filtering changes signals by design—all filters change temporal features of the signal
- Causal filters always incur delay

Group Delay Examples $y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

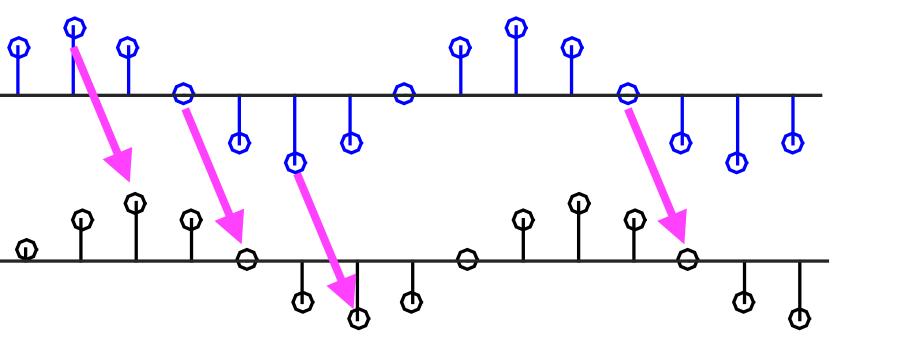


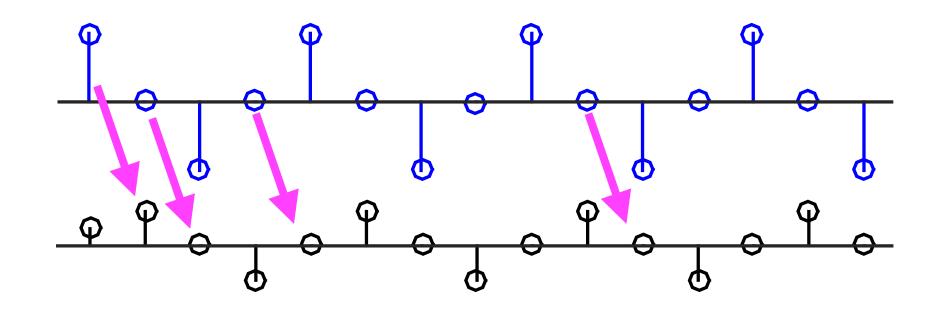
Group Delay Examples $y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$





 $y[t] = \frac{1}{4}x[t] + \frac{1}{2}x[t - \Delta t] + \frac{1}{4}x[t - 2\Delta t]$





Group Delay Examples

 τ_D : Δt

 τ_D :

Group Delay: FIR filters

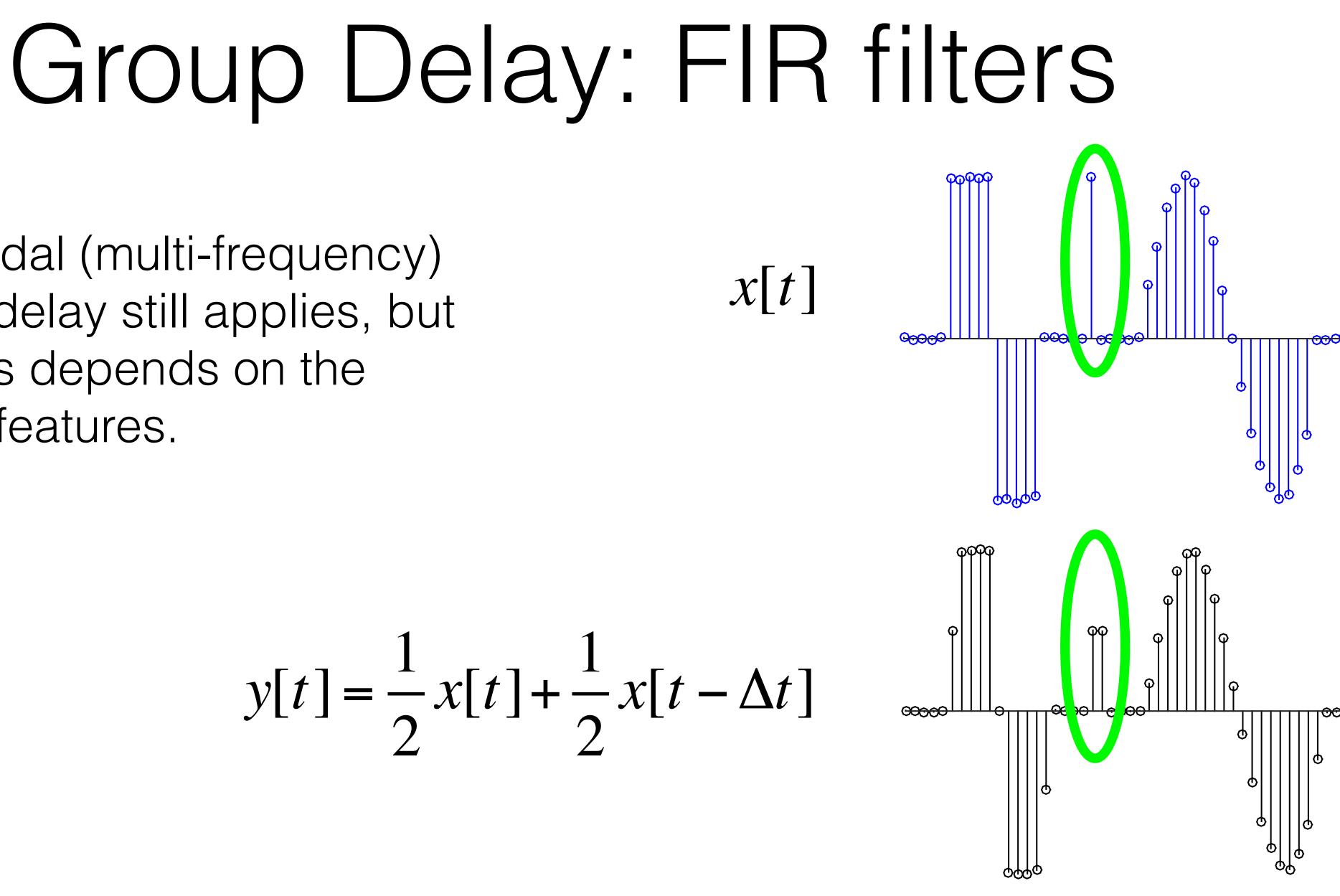
- Group delay corresponds to "average" delay imparted by time-shifted filter terms.
- The group delay of an FIR filter does not depend on frequency.
- The *order* of an FIR filter, *N_{order}*, is the number of time shifts used by the most delayed component, minus 1 (i.e. one less than the length of the filter).
- The group delay of an FIR filter is $\Delta t \times N_{order}/2$.
 - The higher the order, the longer the group delay
 - Calculating latencies? You may need to compensate (OK for *peak* latencies).
 - The smaller Δt , the smaller the delay, so if possible filter at high sampling frequency.

Group Delay: FIR filters x[t] $y[t] = \frac{1}{2}x[t] + \frac{1}{2}x[t - \Delta t]$

For non-sinusoidal (multi-frequency) signals, group delay still applies, but how it manifests depends on the specific signal features.

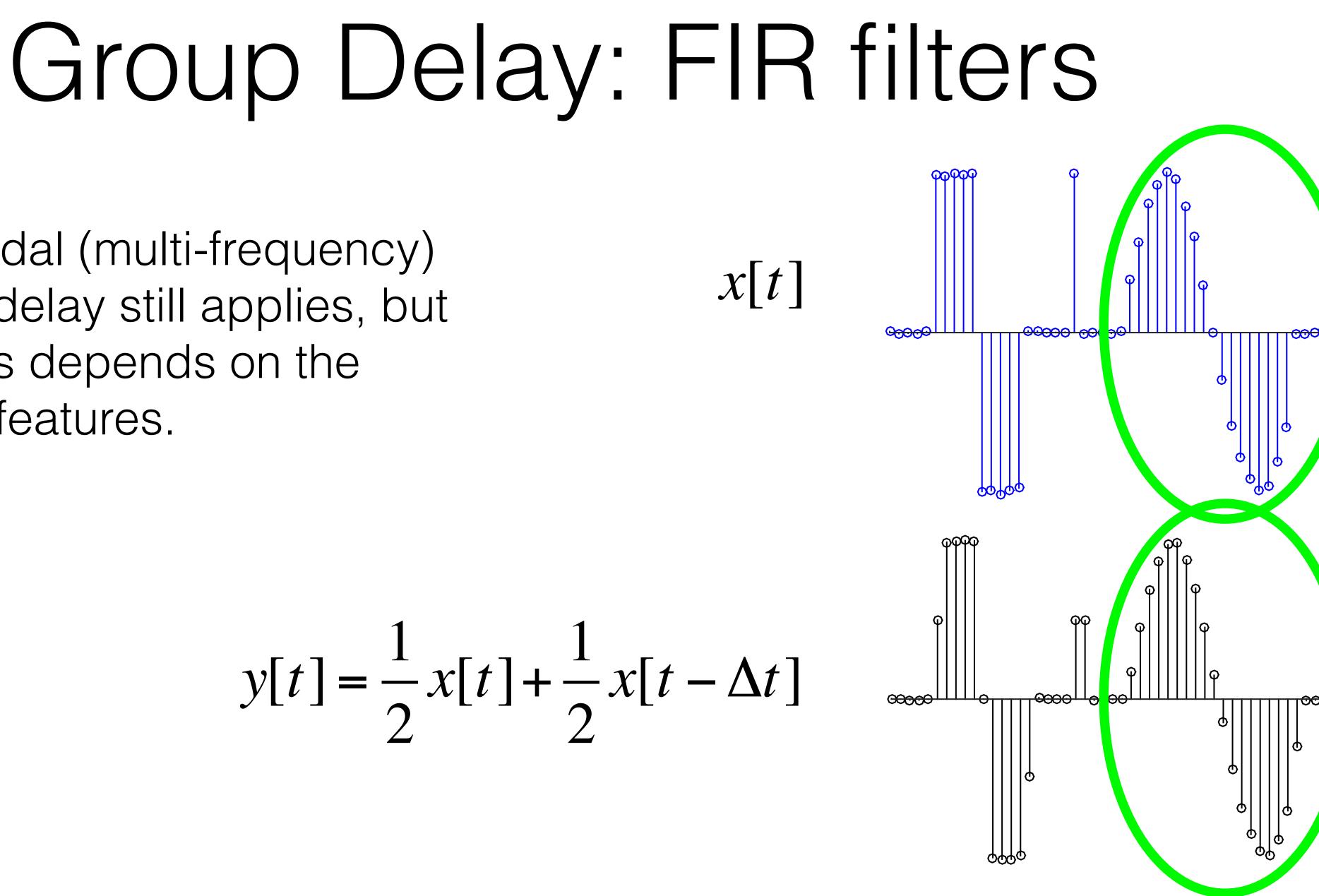


For non-sinusoidal (multi-frequency) signals, group delay still applies, but how it manifests depends on the specific signal features.





For non-sinusoidal (multi-frequency) signals, group delay still applies, but how it manifests depends on the specific signal features.

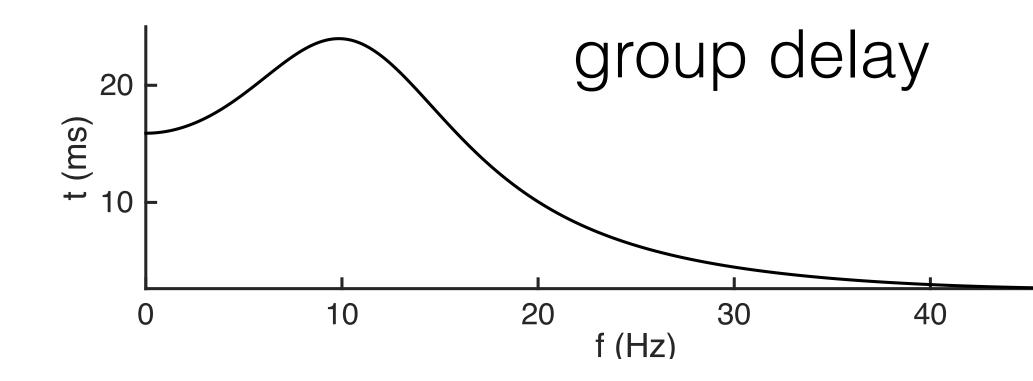






Group Delay: IIR filters

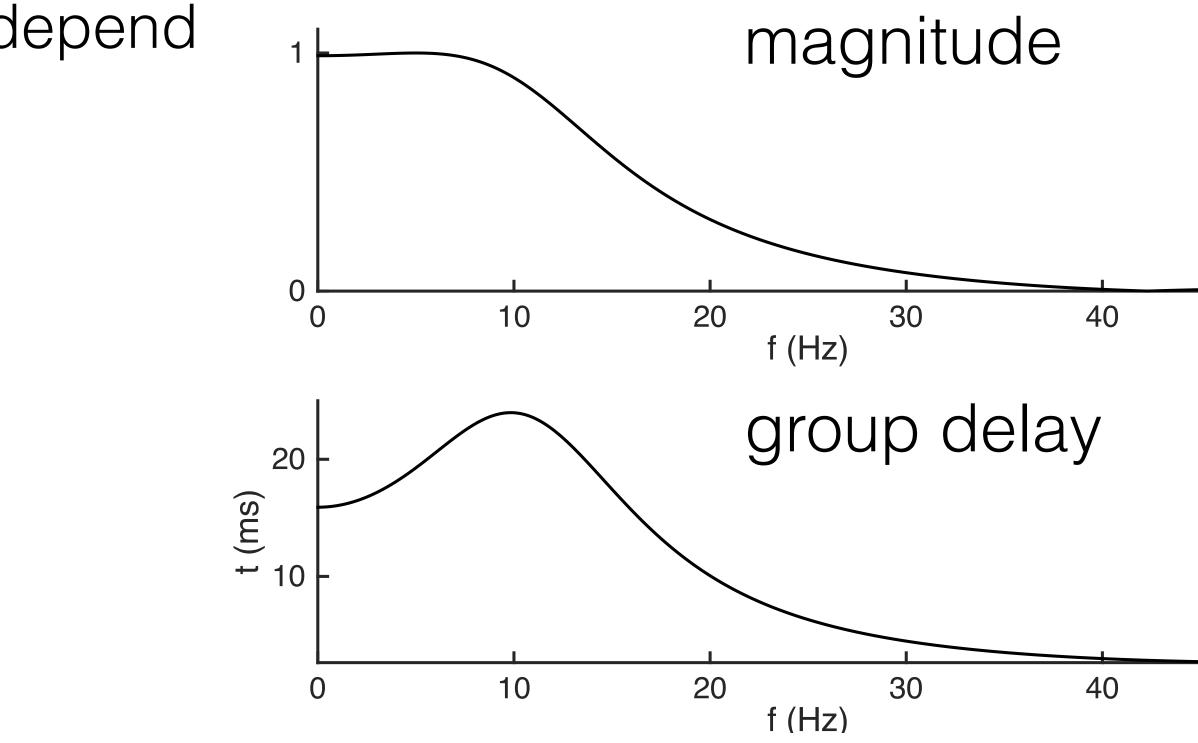
• The group delay of an IIR filter **does** depend on frequency.

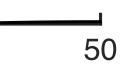




Group Delay: IIR filters

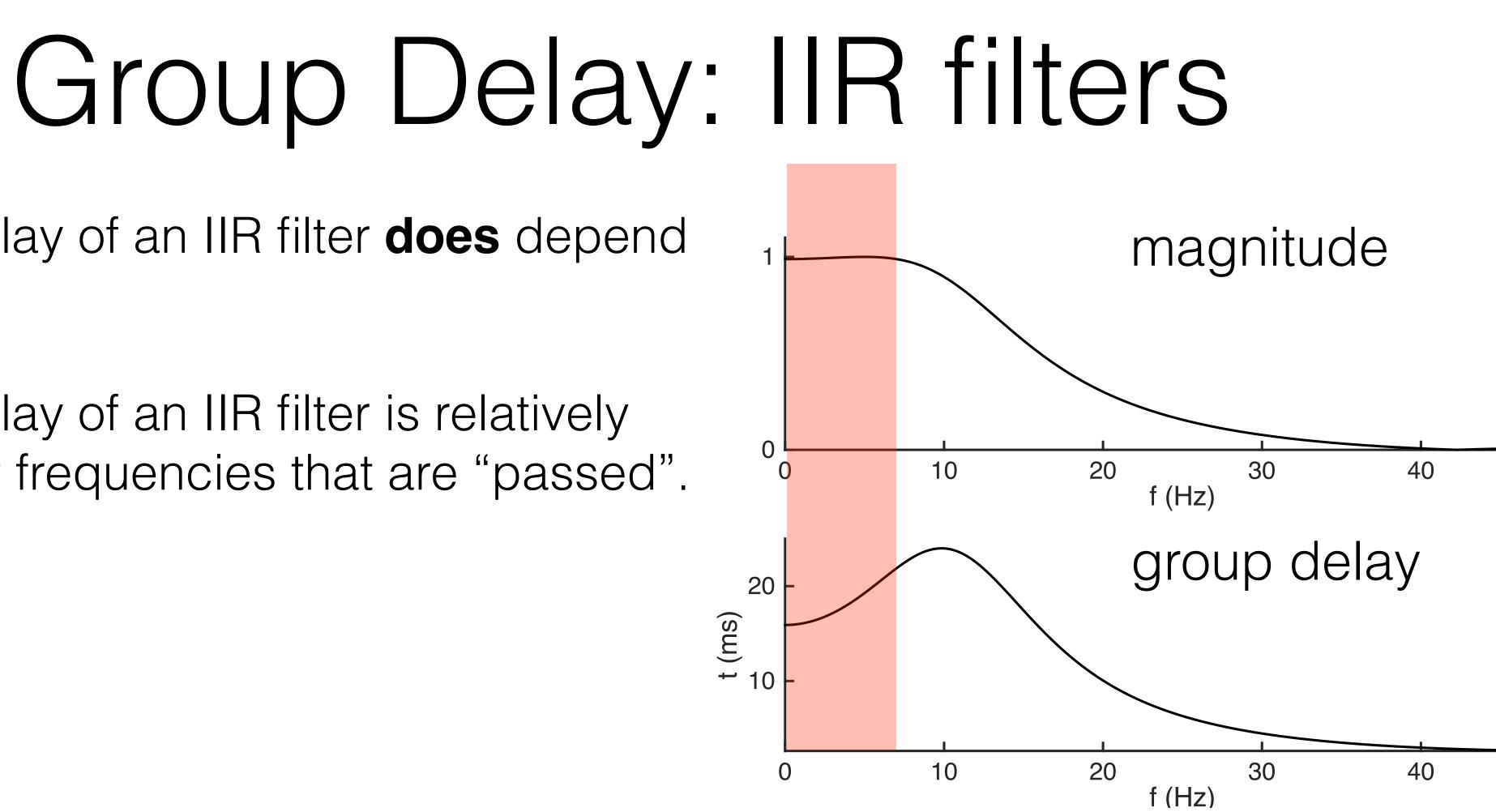
• The group delay of an IIR filter **does** depend on frequency.

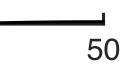






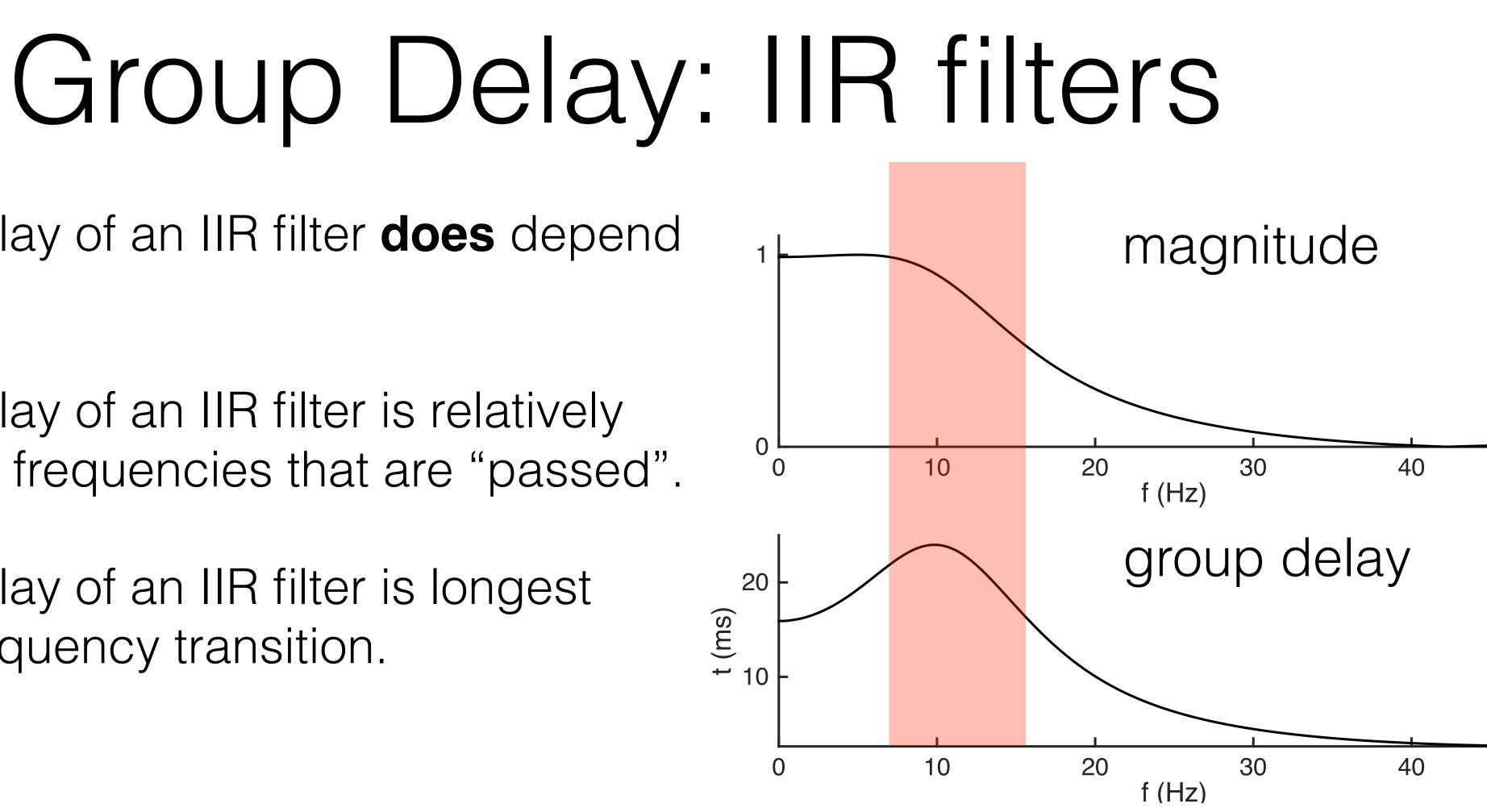
- The group delay of an IIR filter **does** depend on frequency.
- The group delay of an IIR filter is relatively constant over frequencies that are "passed".

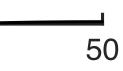






- The group delay of an IIR filter **does** depend on frequency.
- The group delay of an IIR filter is relatively constant over frequencies that are "passed".
- The group delay of an IIR filter is longest during the frequency transition.

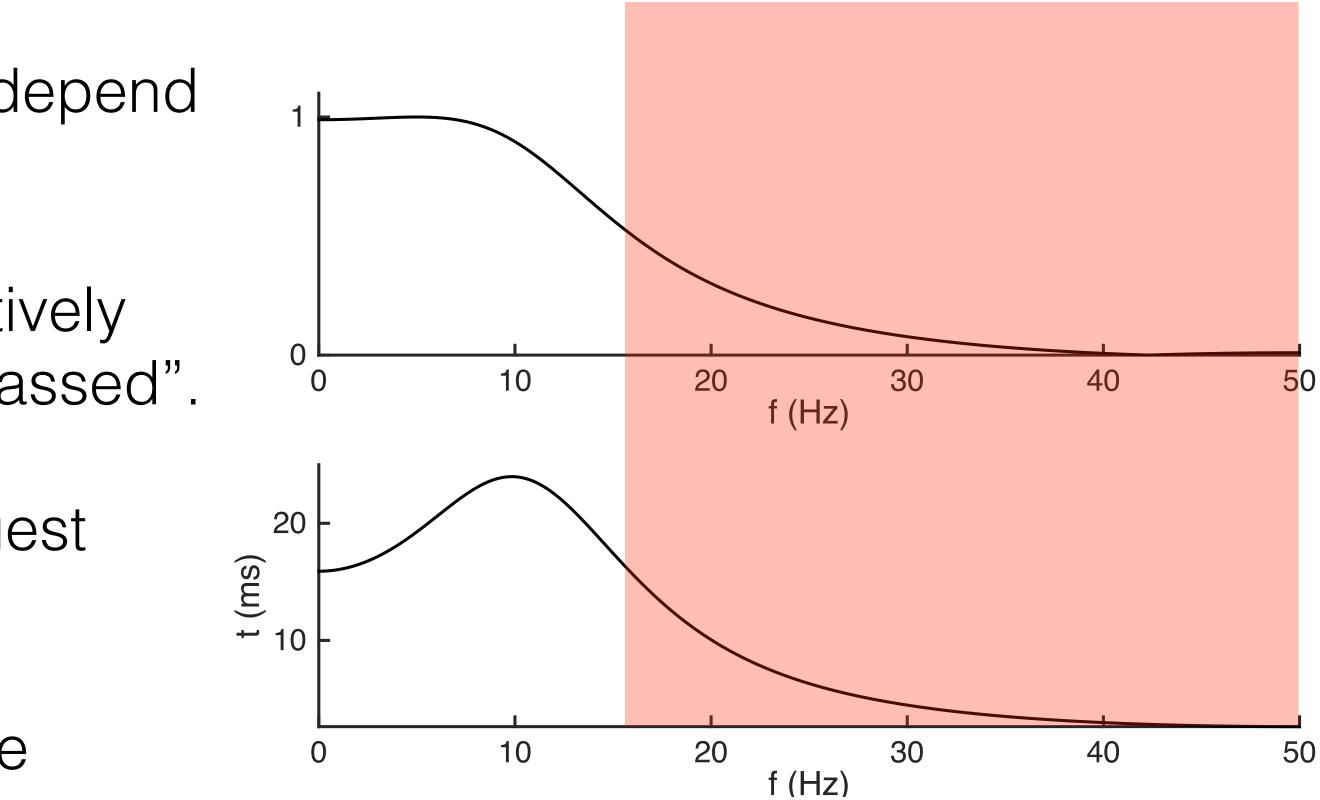






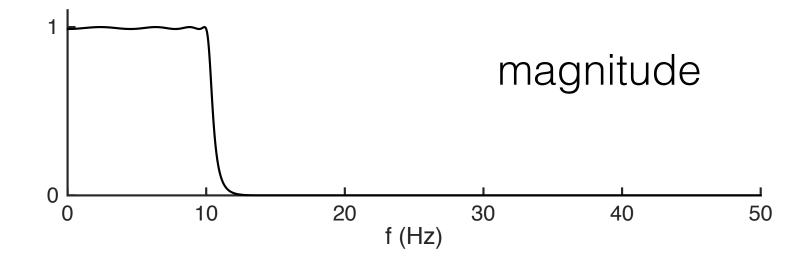
Group Delay: IIR filters

- The group delay of an IIR filter **does** depend on frequency.
- The group delay of an IIR filter is relatively constant over frequencies that are "passed".
- The group delay of an IIR filter is longest during the frequency transition.
- The group delay of an IIR filter may be irrelevant over frequencies that are "stopped".



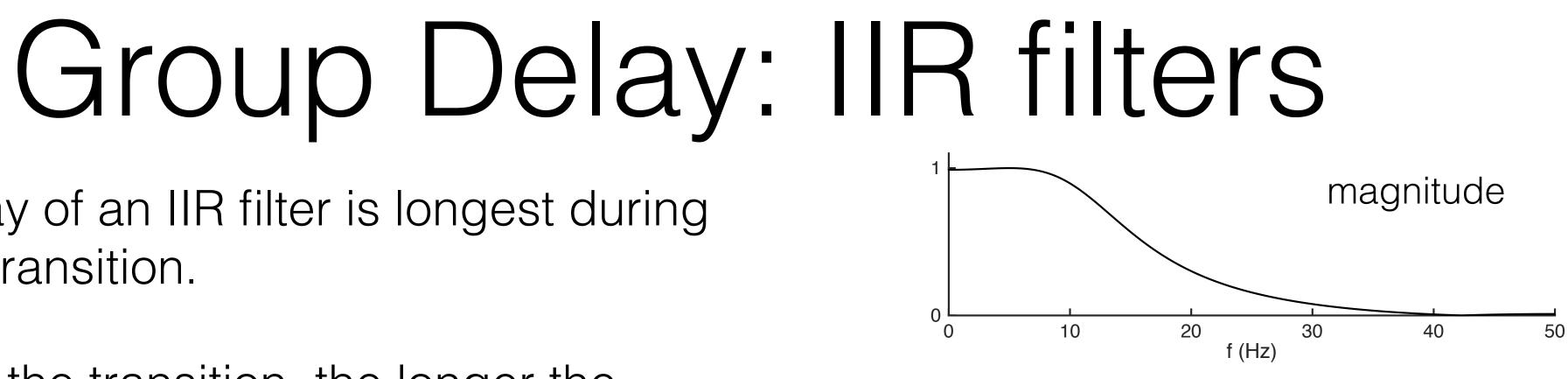
• The group delay of an IIR filter is longest during the frequency transition.

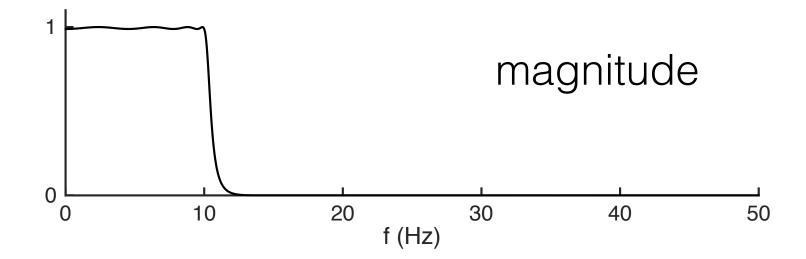




f (Hz)

- The group delay of an IIR filter is longest during the frequency transition.
 - The sharper the transition, the longer the group delay





Group Delay: IIR filters magnitude 0 L 0 20 10 30 50 40 f (Hz) group delay 20 t (ms) 10 F 20 30 50 40 0 10 f (Hz) magnitude 20 30 40 50 10 0 f (Hz) 600 г group delay 400 t (line) 200 t (line) 20 10 30 40 50 0

- The group delay of an IIR filter is longest during the frequency transition.
 - The sharper the transition, the longer the group delay

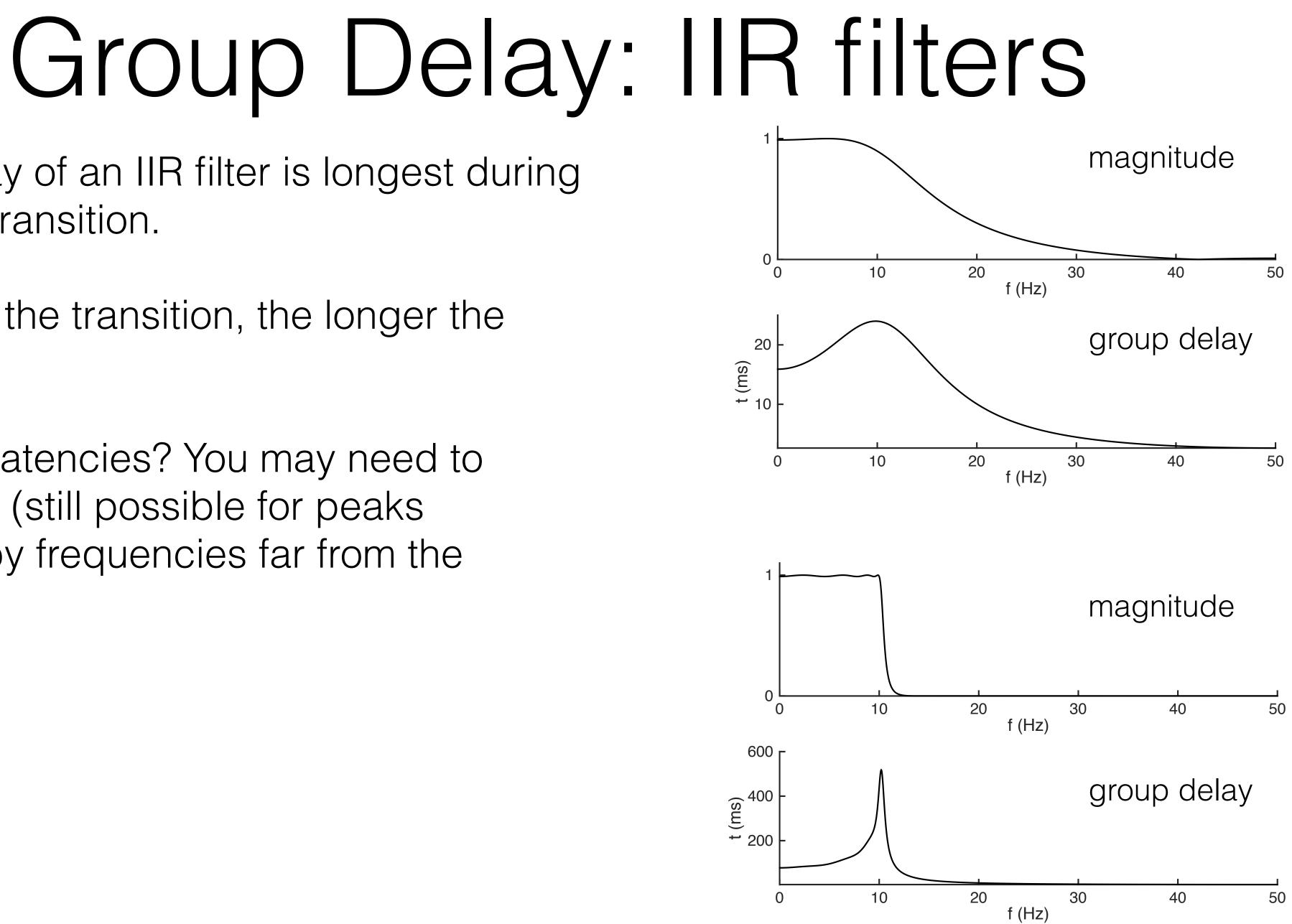
f (Hz)

Group Delay: IIR filters magnitude 0 L 0 20 10 30 50 40 f (Hz) group delay 20 t (ms) 10 F 20 30 50 0 10 40 f (Hz) magnitude 20 30 50 10 40 f (Hz) 600 г group delay 400 (se) 200 20 10 30 40 50 0

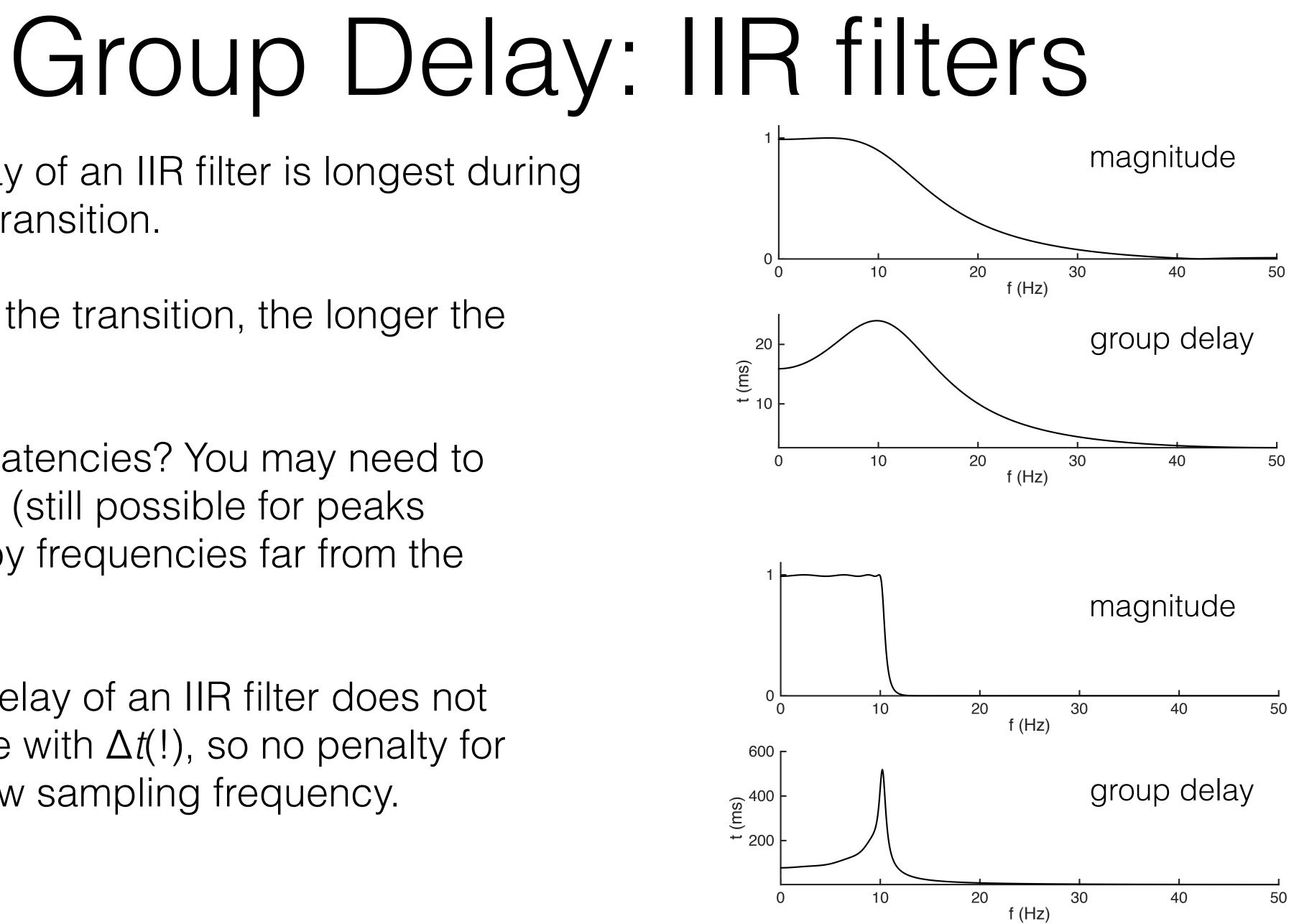
- The group delay of an IIR filter is longest during the frequency transition.
 - The sharper the transition, the longer the group delay

f (Hz)

- The group delay of an IIR filter is longest during the frequency transition.
 - The sharper the transition, the longer the group delay
 - Calculating latencies? You may need to compensate (still possible for peaks dominated by frequencies far from the transition).



- The group delay of an IIR filter is longest during the frequency transition.
 - The sharper the transition, the longer the group delay
 - Calculating latencies? You may need to compensate (still possible for peaks dominated by frequencies far from the transition).
 - The group delay of an IIR filter does not linearly scale with $\Delta t(!)$, so no penalty for filtering at low sampling frequency.



Signal Loss due to Filter Startup

- Output signal value depends on signal values in the past
- When calculating output at the very first moment of time, *there is no past to rely on!*
- Until filter output settles down, in time, the output signal is not well defined.

Signal Loss due to Filter Startup

For FIR filters, this problem goes away entirely after $N_{order} \times \Delta t$.

$$y[t] = \frac{1}{2}x[t] - \frac{1}{2}x[t - \Delta t]: \quad y[0] = \frac{1}{2}x[0] - \frac{1}{2}x[-\Delta t] \qquad y[\Delta t] = \frac{1}{2}x[\Delta t] - \frac{1}{2}x[0]$$

- This works well for small Norder.
- This is another reason to use FIR filters only of low order.

• Recommendation: either keep extra *earlier* data of duration $N_{order} \times \Delta t$, or prepend the same amount of zero signal (Matlab's default). Consider this "warmup" time for the filter. Then toss out this same amount from the output.

This is another reason FIR filters may work best at high sample rates.

Signal Loss due to Filter Startup

For IIR filters, the problem is more subtle

$$y[t] = \frac{1}{10}x[t] - \frac{9}{10}y[t - \Delta t]: \qquad y[0] = \frac{1}{10}x[0] - \frac{9}{10}y[-\Delta t] \qquad y[\Delta t] = \frac{1}{10}x[\Delta t] - \frac{9}{10}y[0]$$

- the past.
- you can afford. Then toss out the same amount from the output.
- data. Using this with reasonable values can really help.

• The output depends not only on the input in the past, but also on the filter output of

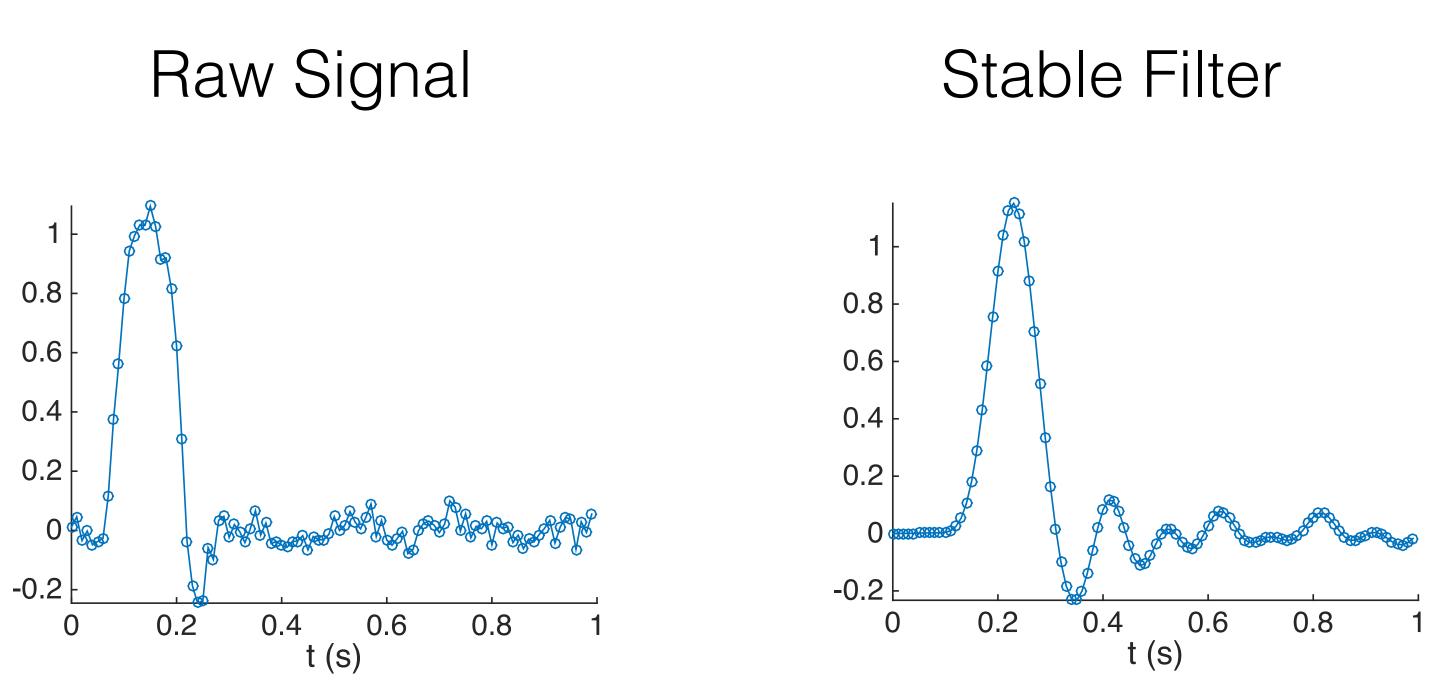
• Recommendation: again keep extra earlier data (warmup time), however much

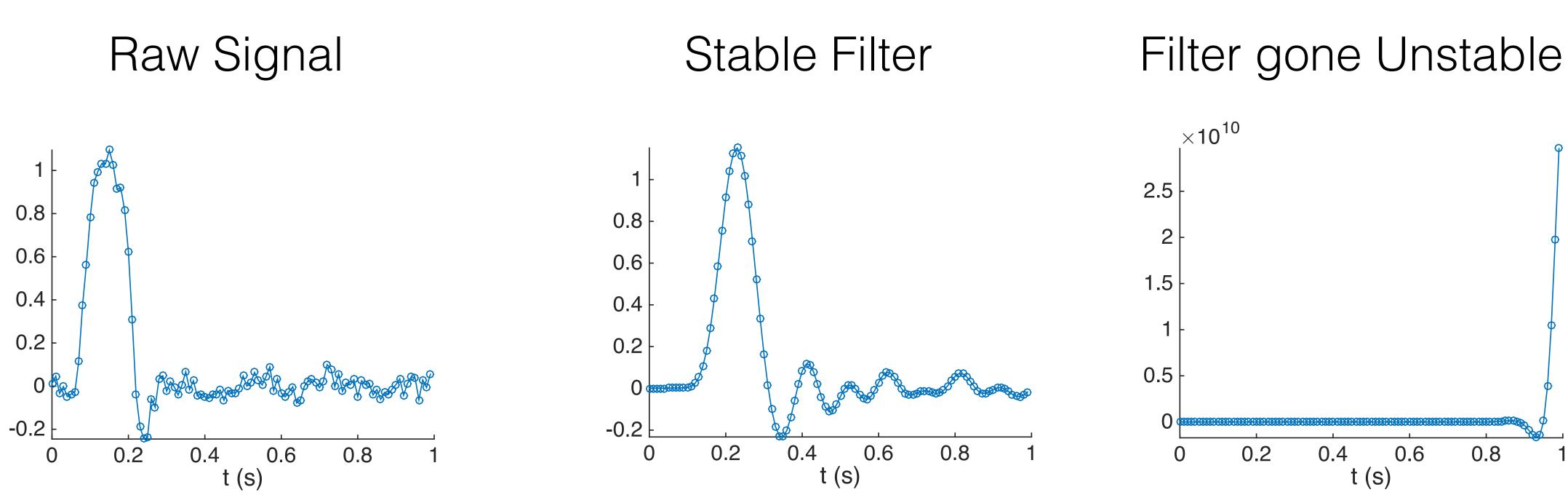
• If keeping enough earlier data not feasible, Matlab permits supplying pre t=0 initial

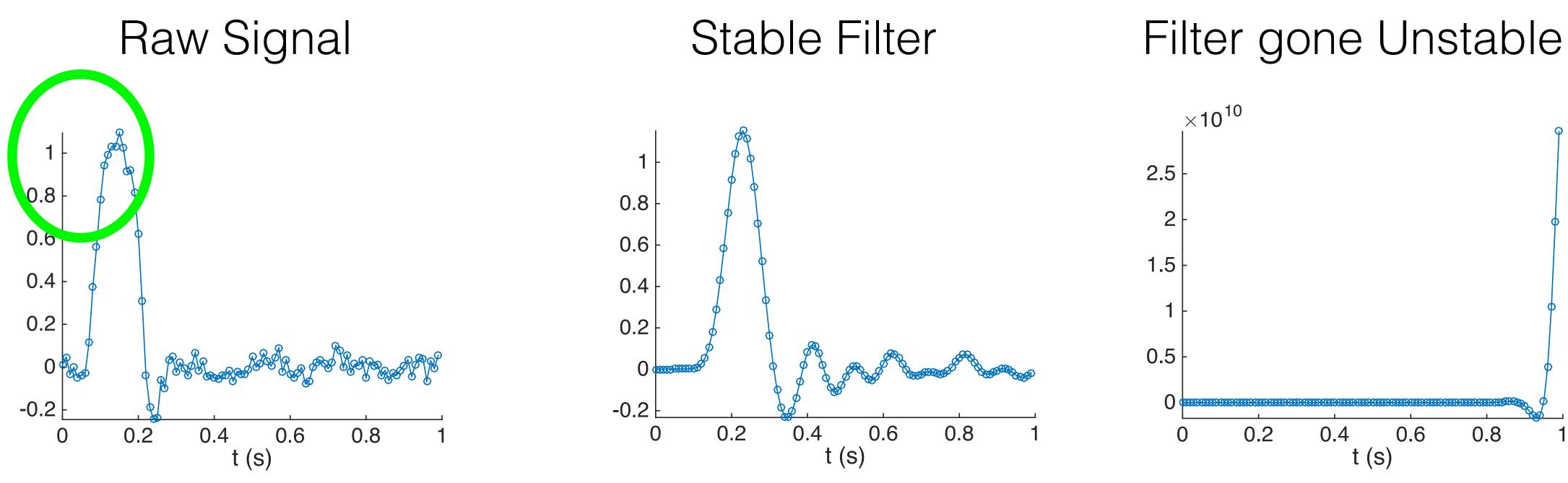
Even data from the end of the signal may help substantially over nothing.

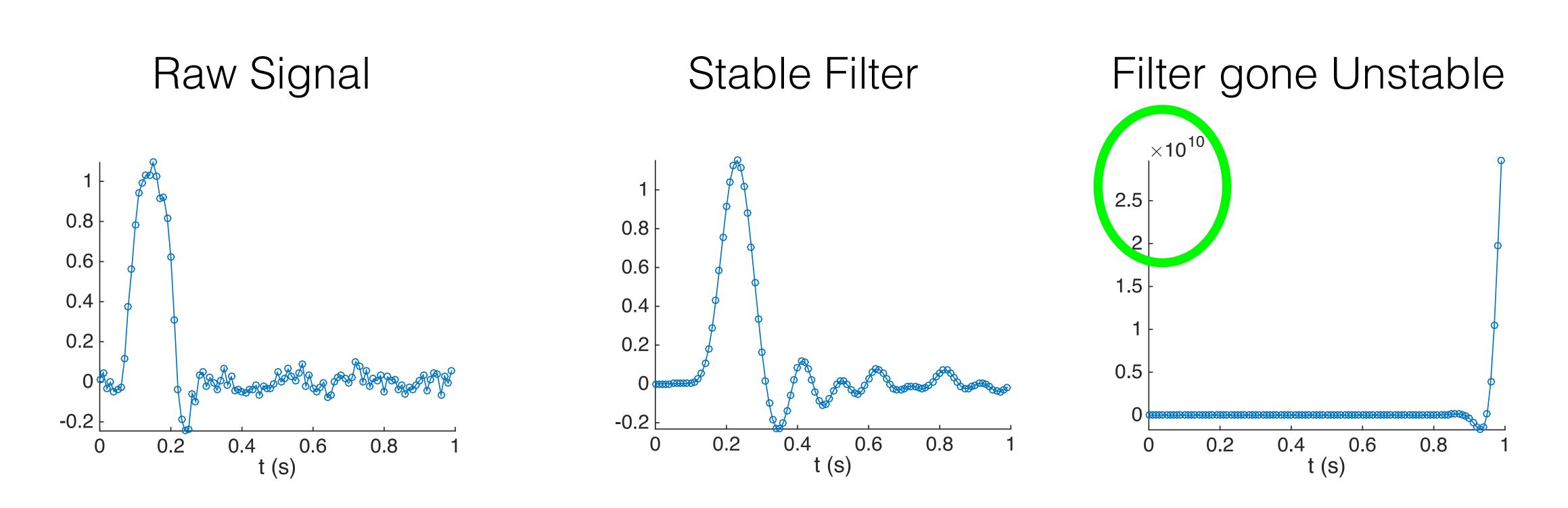
Stability concerns for IIR filters

- IIR filters employ feedback; might be negative (good) or positive (bad)
- Common IIR filters designed to be stable: all feedback negative (good)
- Design can break down due to numerical roundoff error
- Breakdown more likely for higher order filters
- Recommendation: only use low order ($N_{order} < 10$) IIR filters.
 - Lower order IIR filters also have less sharp frequency transitions, so this is rarely a burden.









How Do I Choose a Filter?

For high sampling frequency and plenty of initial data, consider FIR filters

- This is typically the case for raw, un-epoched data.
- Parks-McClellen ("optimal") filters work well. Can choose soft frequency transitions.
 - (Report the filter choice and order, as well as all cutoff frequencies and any other specified parameters, in your *Methods* section.)
- Take care with software "black-box" FIR filters. Maybe good, maybe not.
 - How much quality signal processing does the software author know?

How Do I Choose a Filter?

Otherwise, consider IIR filters

- This is typically the case for epoched data.
- If can't be bothered, Butterworth filters are "fine".
 - If really can't be bothered, use a 4th order Butterworth.
- If you care about your frequency bands, consider using an Elliptic filter.
 - (Report the filter choice and order, as well as all cutoff frequencies and any other specified parameters, in the *Methods* section.)
- Software "Black-box" IIR filters usually not worrisome.

How Do I Choose a Filter?

If you care about your frequency bands, consider using an Elliptic filter

- Needs "slop" factors/tolerances
 - In the pass frequency band, how close to "1" (100% let through) do you really need? If your peak height were off by only 1%, would you even notice?
 - Matlab requires this ("passband ripple") to be in dB: $1\% \approx 0.1$ dB
 - In the stop frequency band, how close to "0" (0% let through) do you really need? If your noise is suppressed only by 100x, would you even notice?
 - Matlab requires this ("stopband attenuation") to be in dB: 100x = 40 dB

- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

Causal & non-Causal Filtering

All filters discussed hear are *causal*.

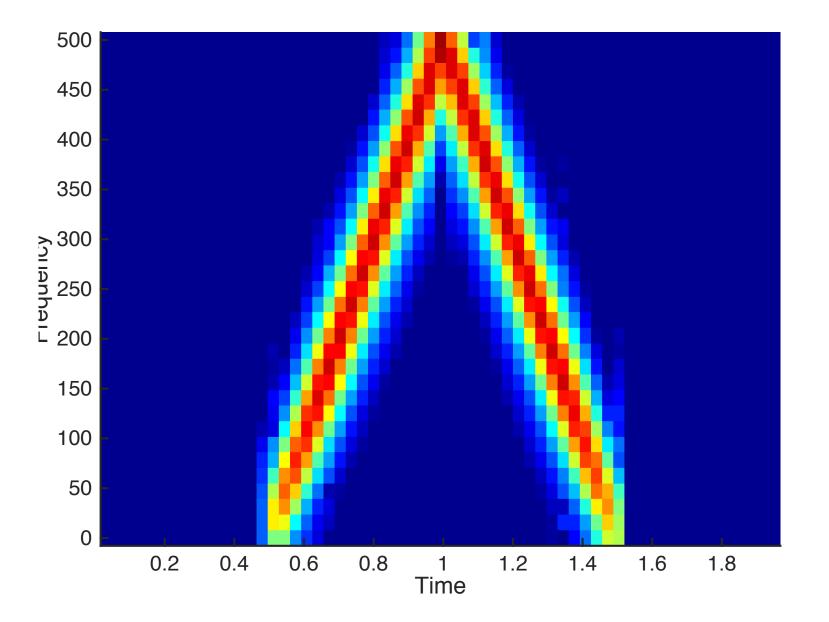
- Changes to the input signal cause changes in the output. The output changes occur at the same moment, or later, but never earlier.
 - Some output changes are desirable: using a low pass filter to slow down fast changes in the input signal.
 - Some output changes are undesirable: ringing due to increase of transition-frequency content in the input signal.

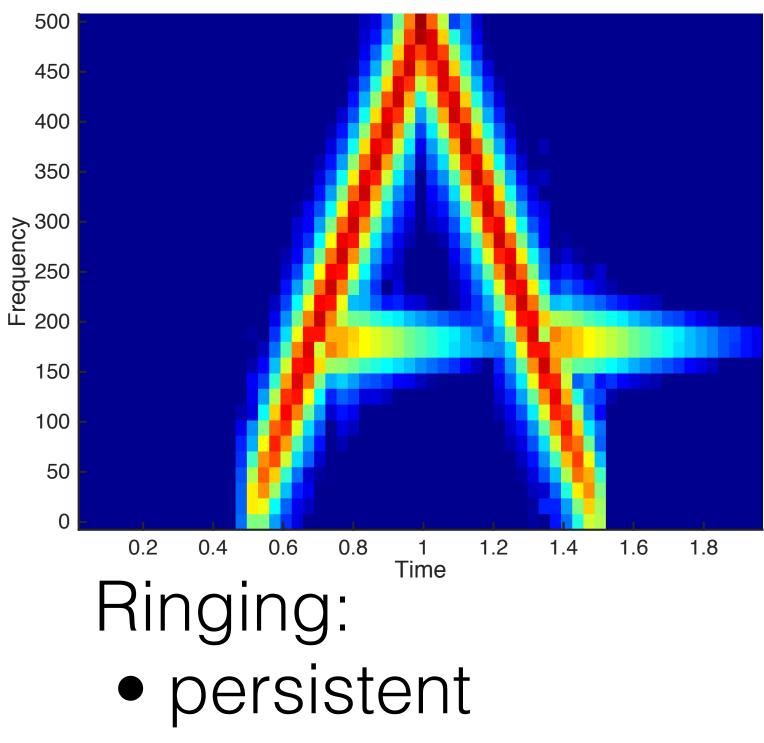
Causal & non-Causal Filtering

- It is mathematically possible (but biologically undesirable!), to temporally "center" all such output changes so they do not seem to be all contribute to delay.
- This (undesirable act) can be achieved with a particular kind of noncausal filtering: *zero-phase* filtering (Matlab "filtfilt").
- Zero-phase sounds wonderful, but it is not (c.f. "ideal" filter).
- Zero-phase filters *do not remove delay-based artifacts*, and in fact they double them.

Zero-Phase Filtering Example

FM Sweep Notched FM Sweep (Spectrogram)

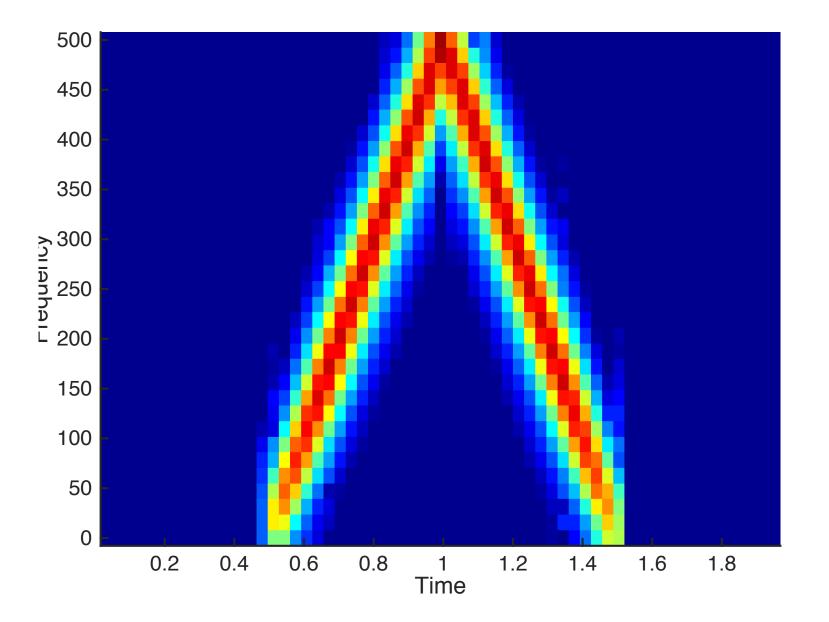


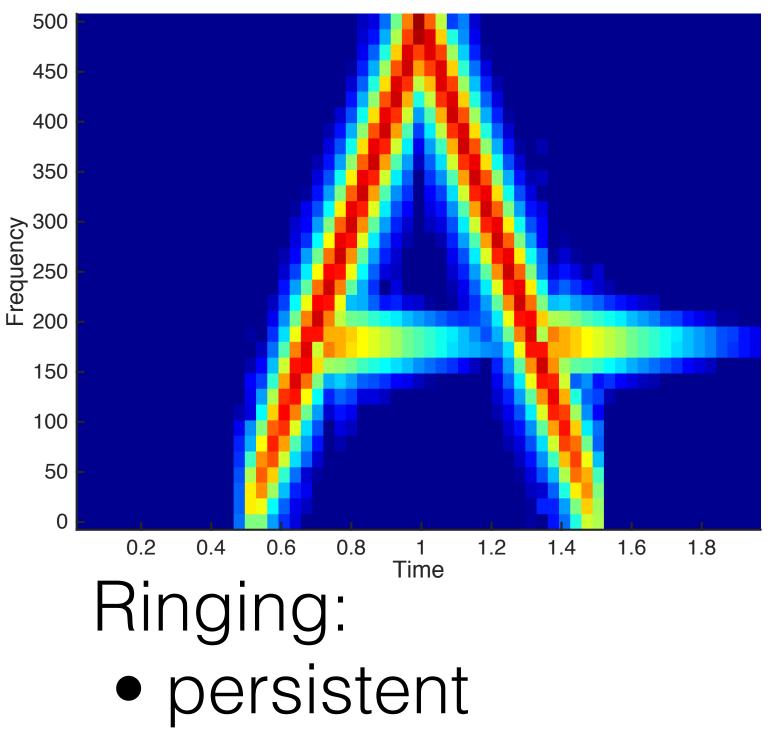


• causal

Zero-Phase Filtering Example

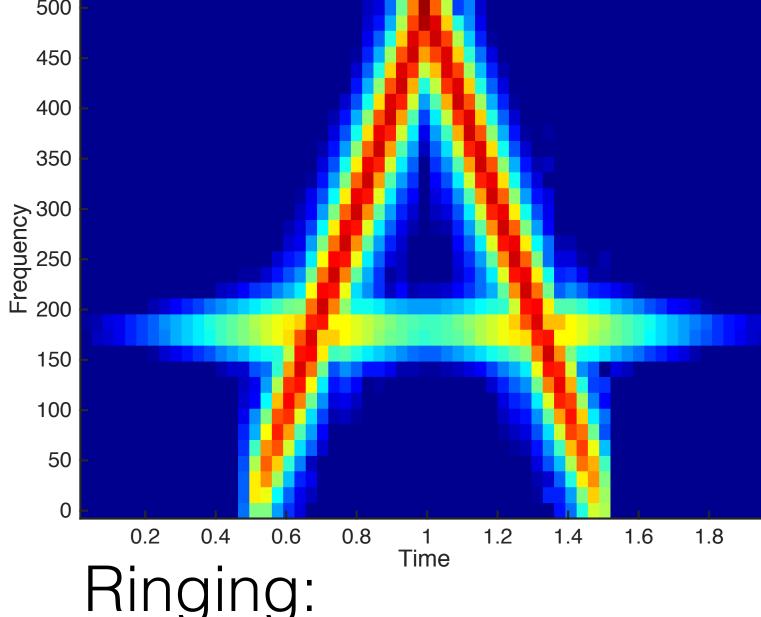
FM Sweep Notched FM Sweep (Spectrogram)





causal

Zero-Phase Notched FM Sweep



 duplicated and flipped no cancellation (except) "on average"?)



Causal & non-Causal Filtering

- out symmetrically.
- similarly spread out by duplication.
 - causal rise.

Zero-phase filters do not remove distortions, but instead spread them

Spreading them out gives zero "on average" but not actually zero.

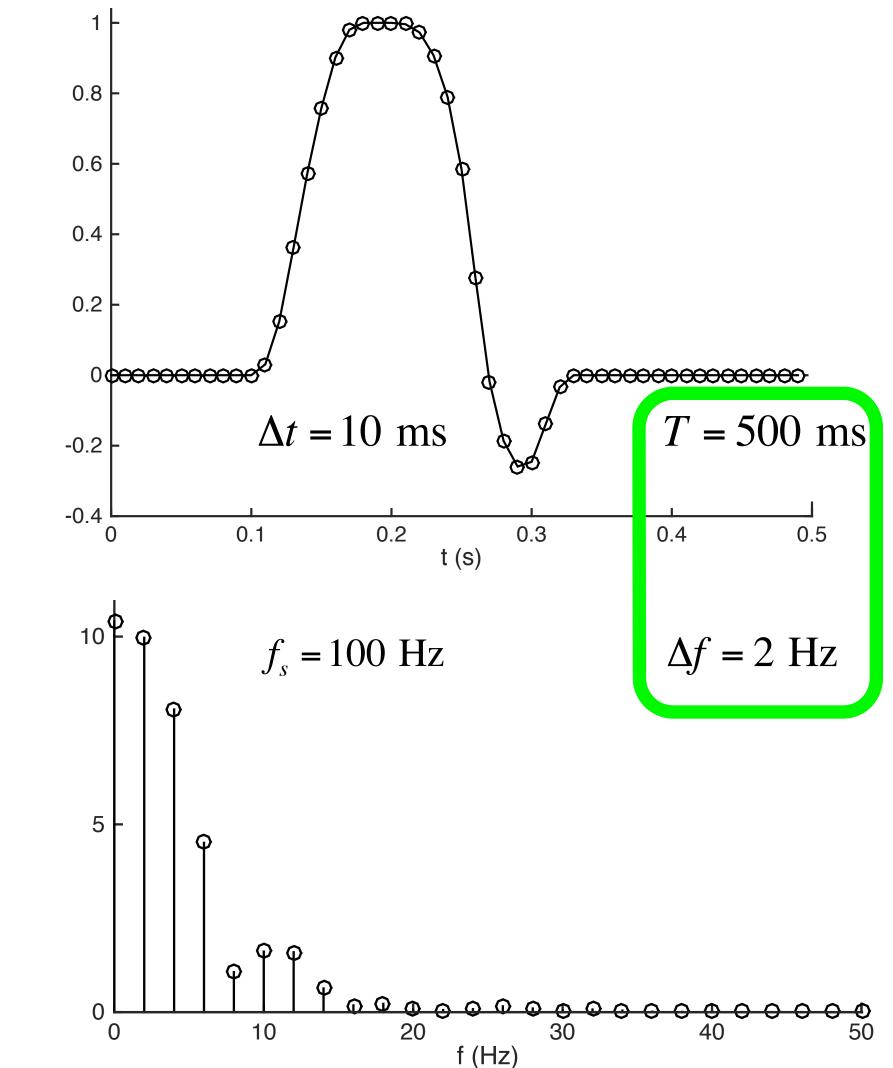
Other temporal features (e.g. smoothing, change detection) get

May remove peak delay, but replaces un-delayed rise with an anti-

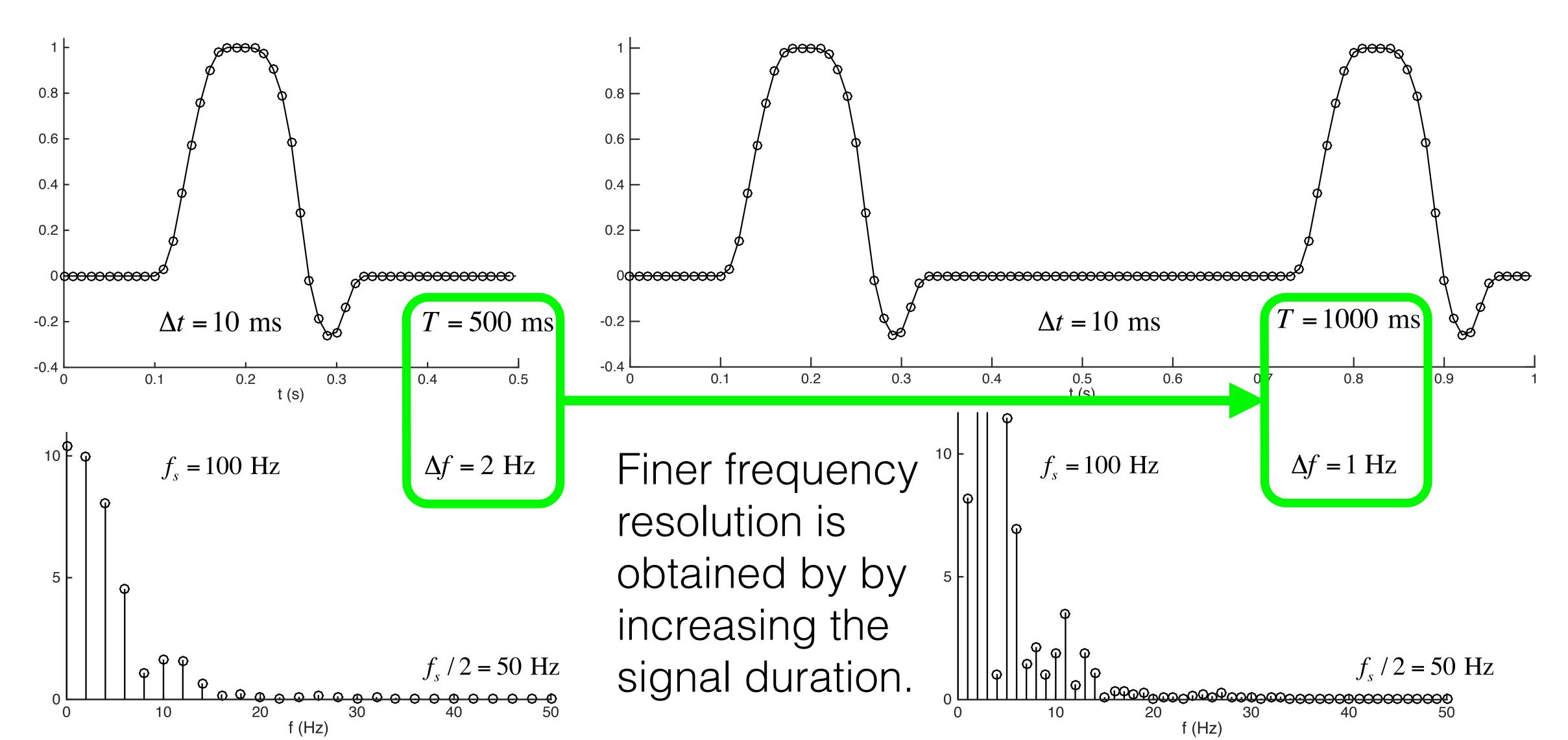
Recommendation: Don't use. Causes far more problems than solutions.

Windowing and Frequency Resolution

- Frequency resolution (Δf), the limiting factor in distinguishing one frequency from another, is determined by the total *duration* of the signal (*T*).
- This relationship is the timefrequency conjugate of the relationship between and *temporal resolution* (Δf) and *sampling frequency* (f_s).



Windowing and Frequency Resolution



Windowing and Frequency Resolution

- It is sometimes desirable to "smear" information temporally (e.g. low-pass filter in order to attenuate noise).
 - The *effective* time resolution is worse, even though Δt remains unchanged.
- Analogously, it is sometimes desirable to "smear" information over frequencies (e.g. to attenuate spectral leakage).
 - The *effective* frequency resolution is worse, even though Δf remains unchanged.
- This frequency smearing is typically accomplished by *windowing* in the time domain.

"Fourier coefficients do not always mean what you think they mean."

-The Princess Bride (paraphrased)

Example 1

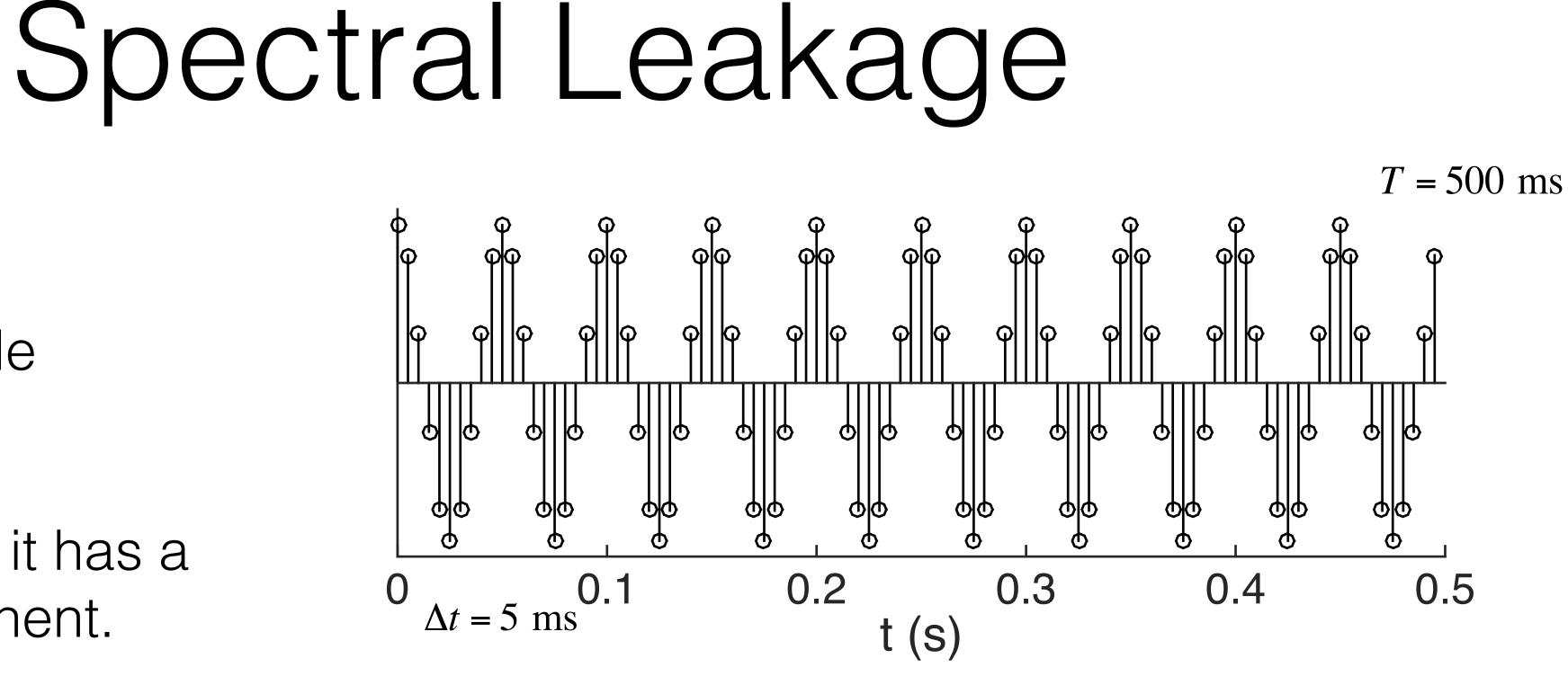
A pure sinusoid (single frequency).

In the Fourier domain it has a single Fourier component.

 $x[t] = \cos(2\pi f_a t)$ = 20Ja



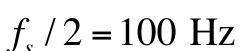
0



Q 20 80 40 60 100 $\Delta f = 2 \text{ Hz}$ f (Hz)







Example 2

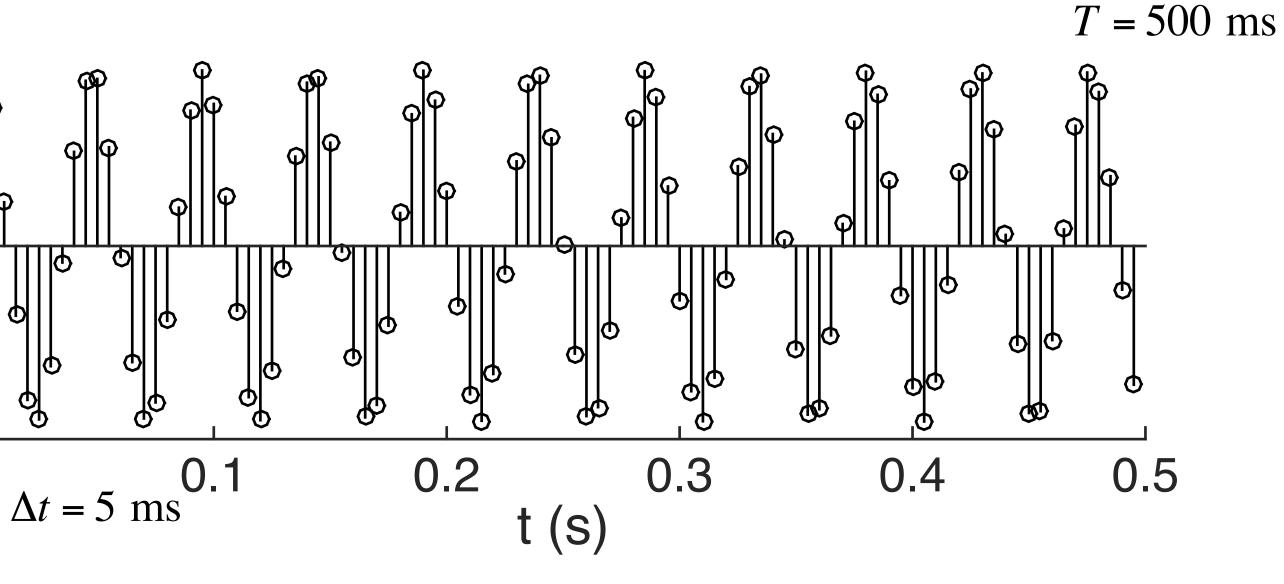
A pure sinusoid (single frequency).

What does it look like in the Fourier Domain?

$$x[t] = \cos(2\pi f_b t)$$
$$f_b = 21 \text{ Hz}$$

0





 $f_{s} = 200 \text{ Hz}$

$$f_{s} / 2 =$$





Example 2

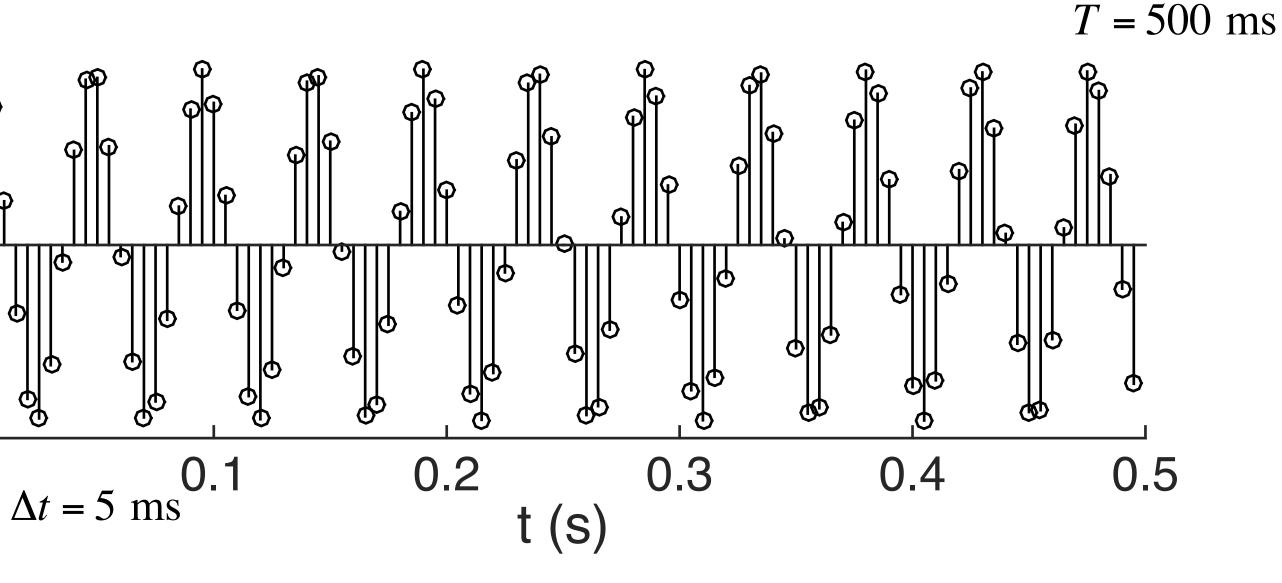
A pure sinusoid (single frequency).

What does it look like in the Fourier Domain?

 $x[t] = \cos(2\pi f_b t)$ =21 hz Jb $\Delta f = 2 \text{ Hz}$

0





 $f_{\rm s} = 200 \, {\rm Hz}$

$$f_{s} / 2 =$$





0

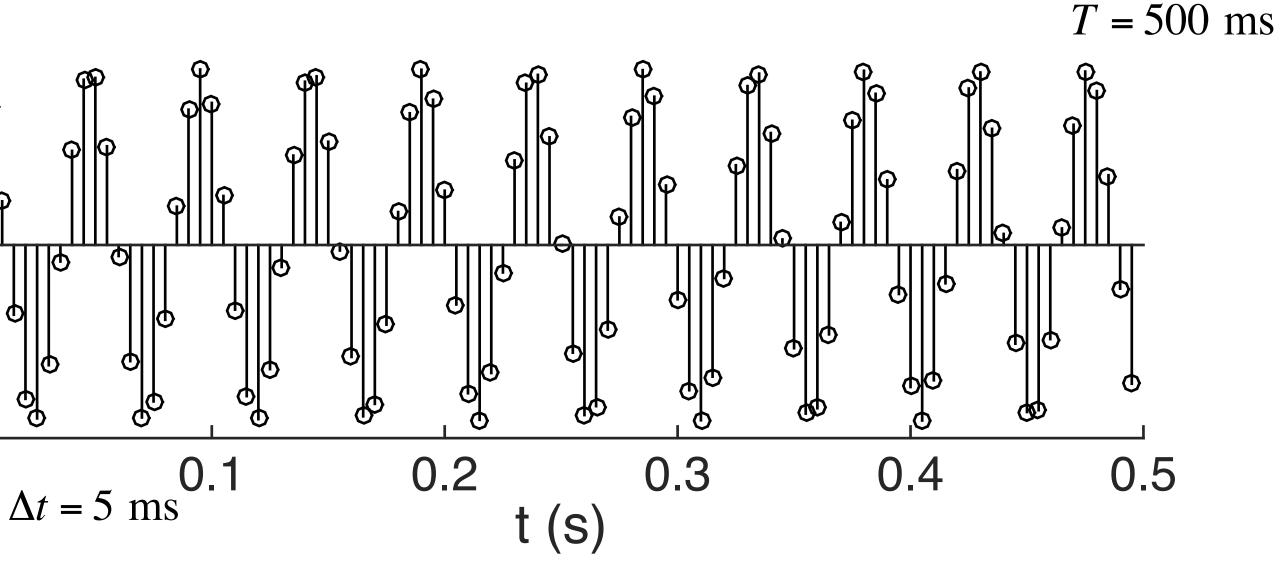
Example 2

A pure sinusoid (single frequency).

What does it look like in the Fourier Domain?

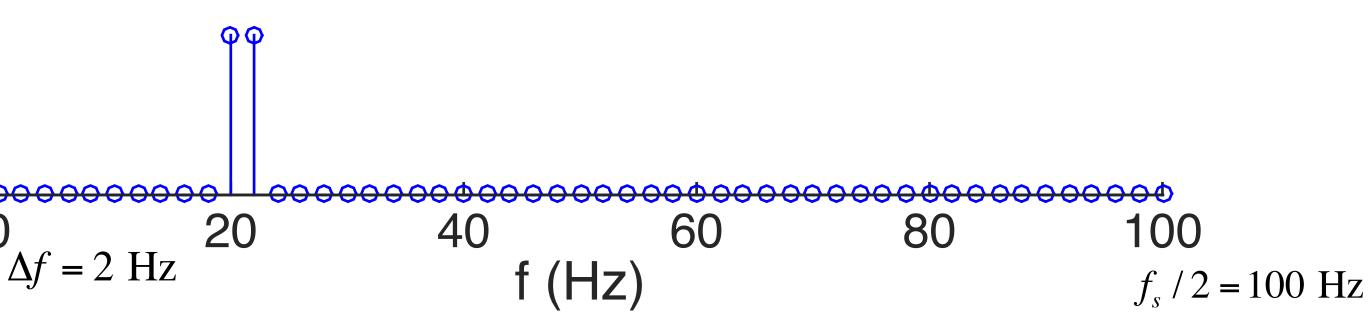
 $x[t] = \cos(2\pi f_b t)$ = 7 Jb f = 2 Hz





 $f_{\rm s} = 200 \, {\rm Hz}$









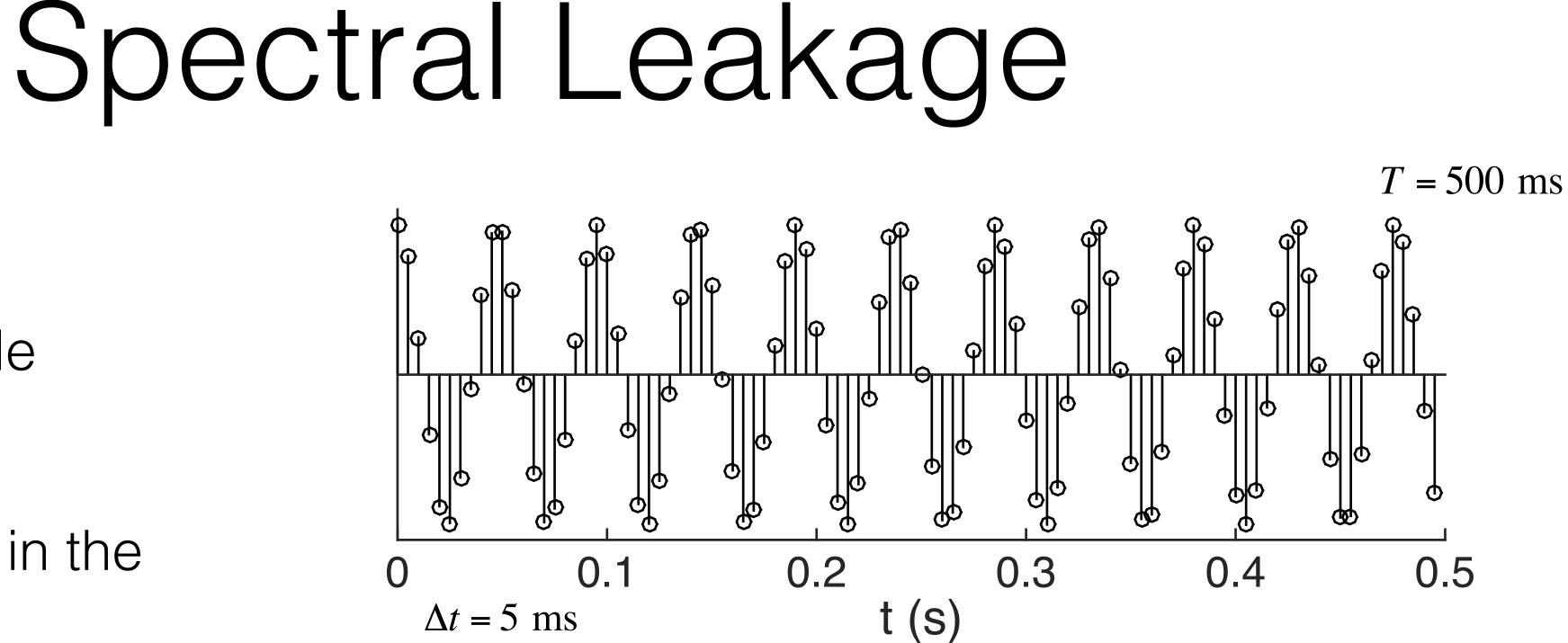
0

Example 2

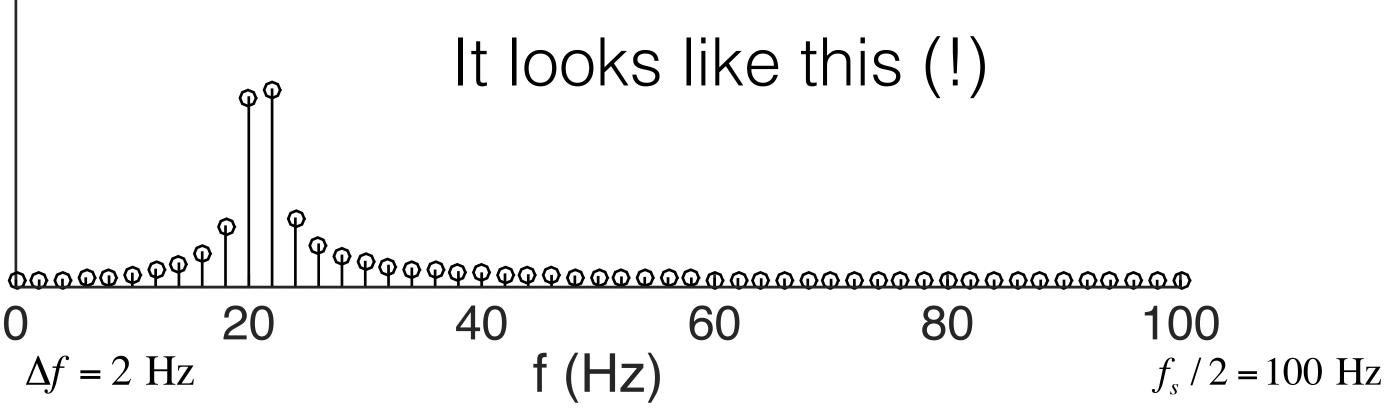
A pure sinusoid (single frequency).

What does it look like in the Fourier Domain?

 $x[t] = \cos(2\pi f_b t)$ =Jb f = 2 Hz



 $f_{\rm s} = 200 \; {\rm Hz}$







0

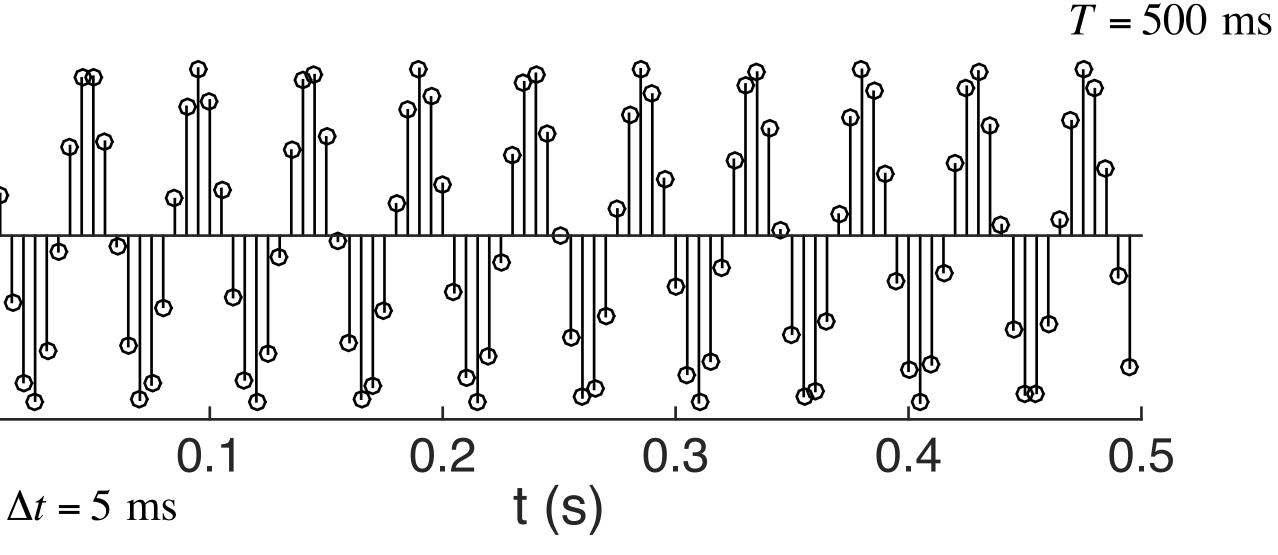
Example 2

A pure sinusoid (single frequency).

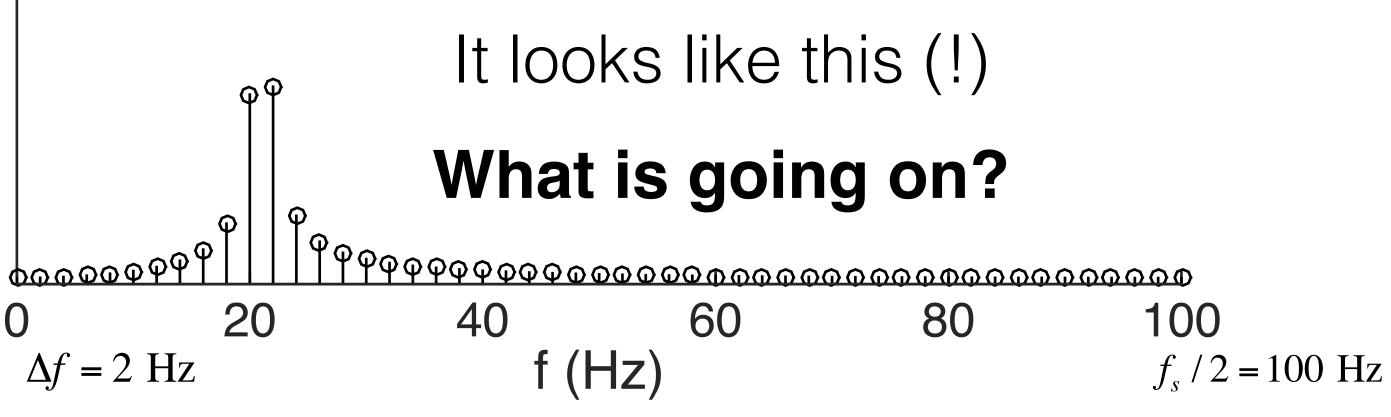
What does it look like in the Fourier Domain?

 $x[t] = \cos(2\pi f_b t)$ =21 hz Jb f = 2 Hz





 $f_{\rm s} = 200 \; {\rm Hz}$







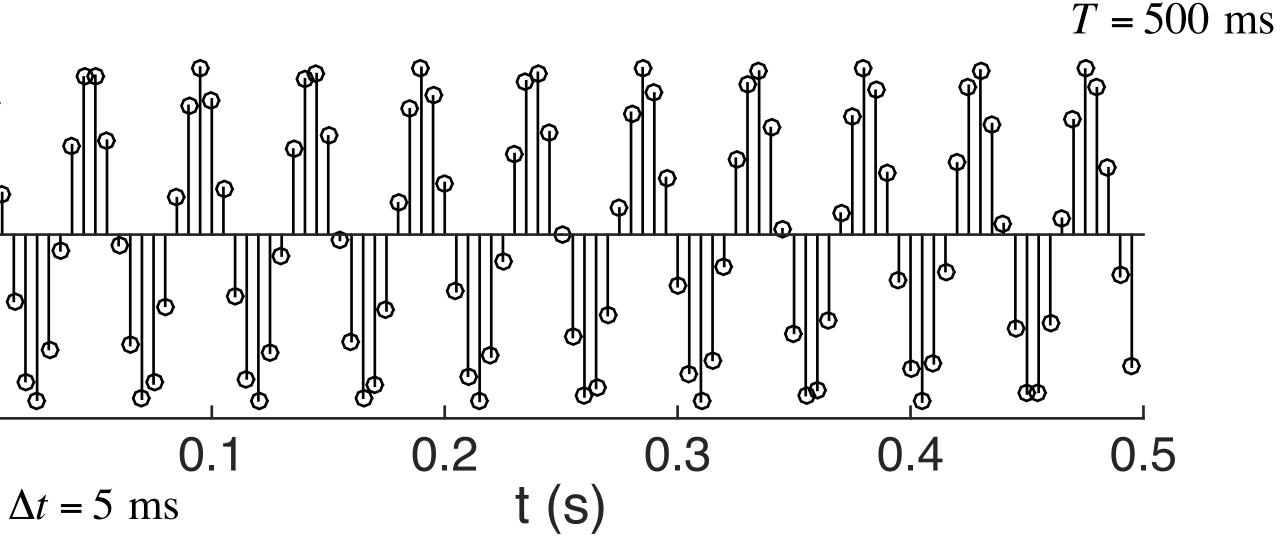
0

A sinusoid whose single frequency is *not* a Fourier frequency exhibits Spectral Leakage.

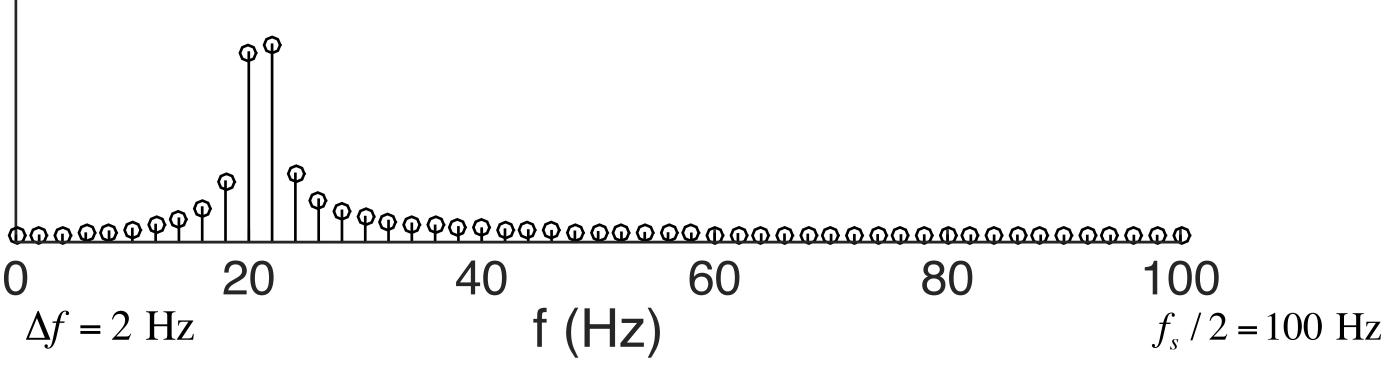
Spectral Leakage of a strong signal component can easily overwhelm weaker nearby signal components.

> $x[t] = \cos(2\pi f_b t)$ Jb f = 2 Hz





 $f_{\rm s} = 200 \; {\rm Hz}$







0

What is the origin of spectral leakage?

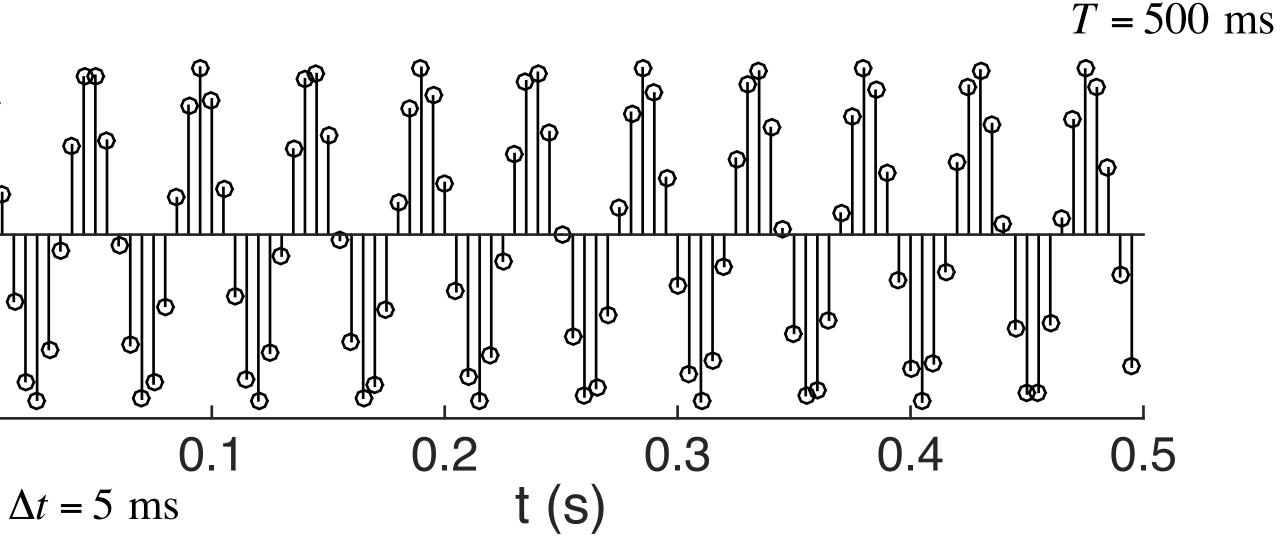
This signal is a cosine, but not periodic with period 2π . The ends do not match.

This can be seen by rotating the signal by T/2, which does affect the Fourier transform in magnitude.

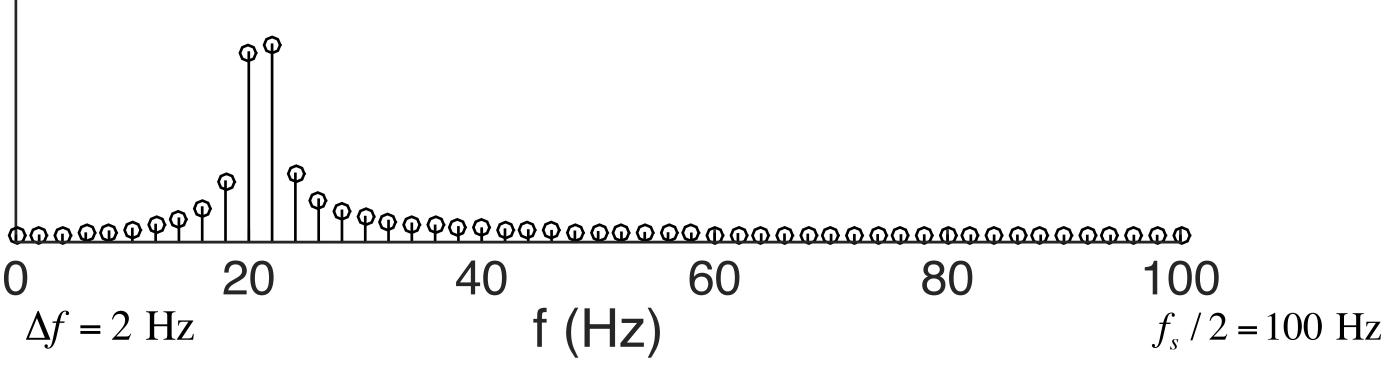
Signal discontinuities are spectrally broadband!

> Jb f = 2 Hz





 $f_{\rm s} = 200 \, {\rm Hz}$







0

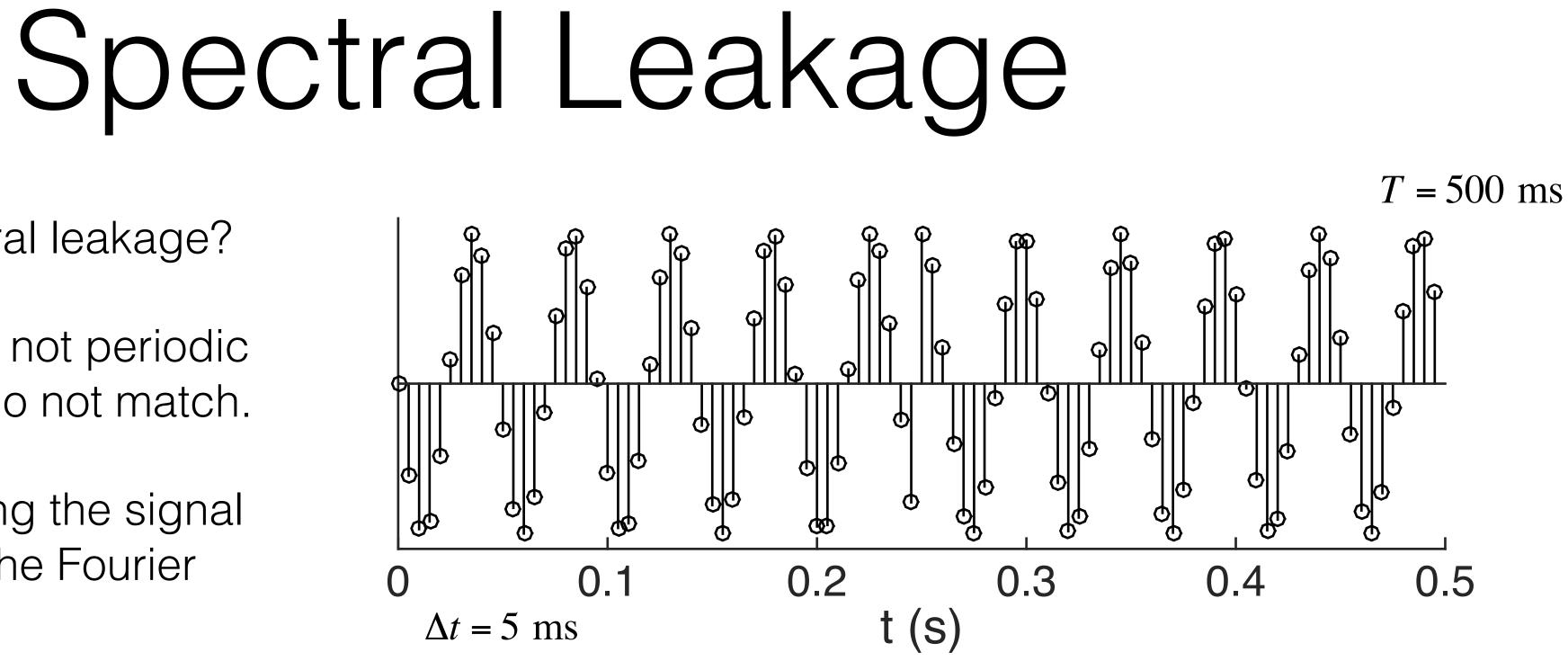
What is the origin of spectral leakage?

This signal is a cosine, but not periodic with period 2π . The ends do not match.

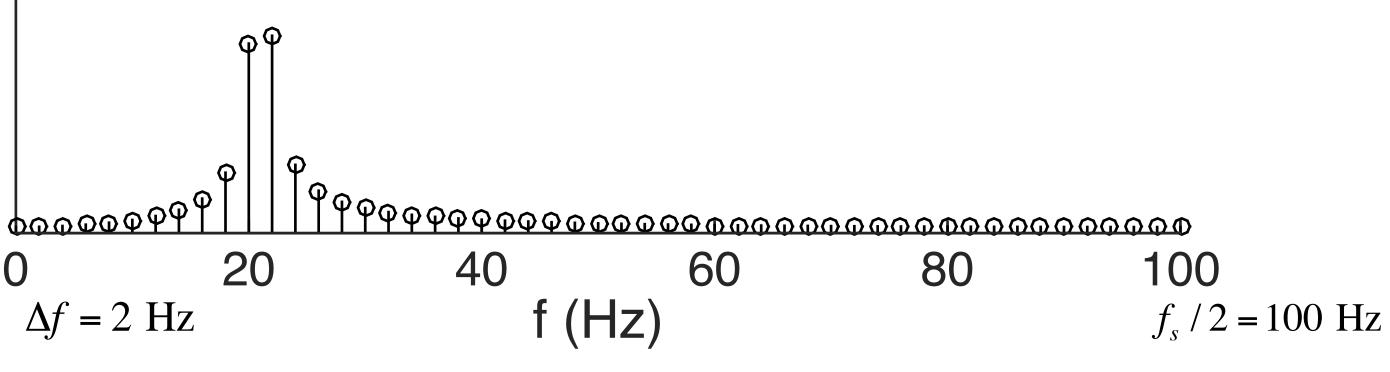
This can be seen by rotating the signal by T/2, which does affect the Fourier transform in magnitude.

Signal discontinuities are spectrally broadband!

> Jb f = 2 Hz



 $f_{\rm s} = 200 \, {\rm Hz}$







0

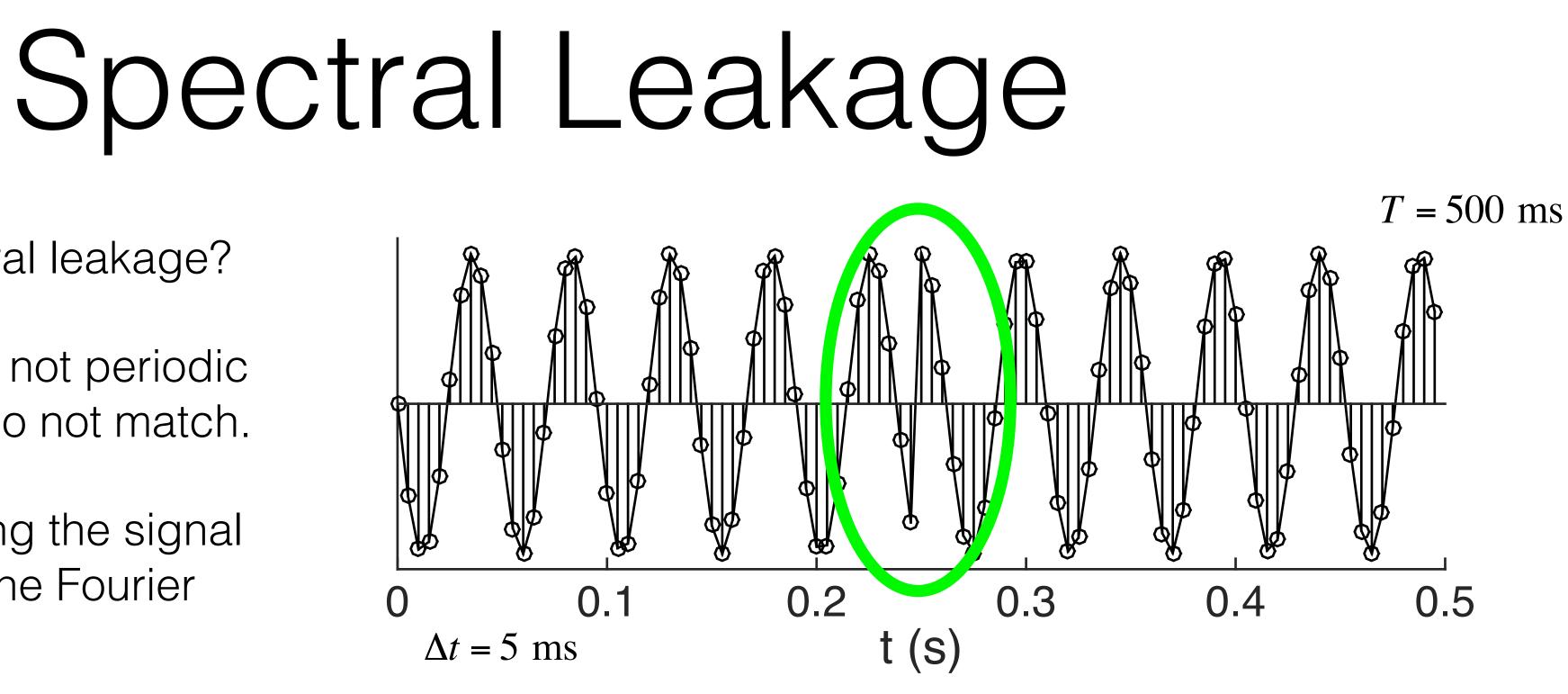
What is the origin of spectral leakage?

This signal is a cosine, but not periodic with period 2π . The ends do not match.

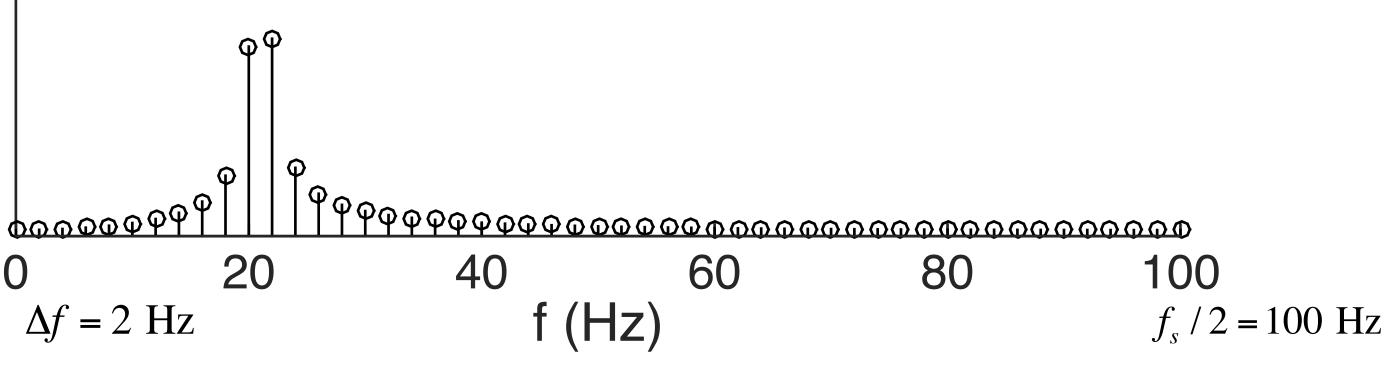
This can be seen by rotating the signal by T/2, which does affect the Fourier transform in magnitude.

Signal discontinuities are spectrally broadband!

> Jb f = 2 Hz



 $f_{\rm s} = 200 \, {\rm Hz}$





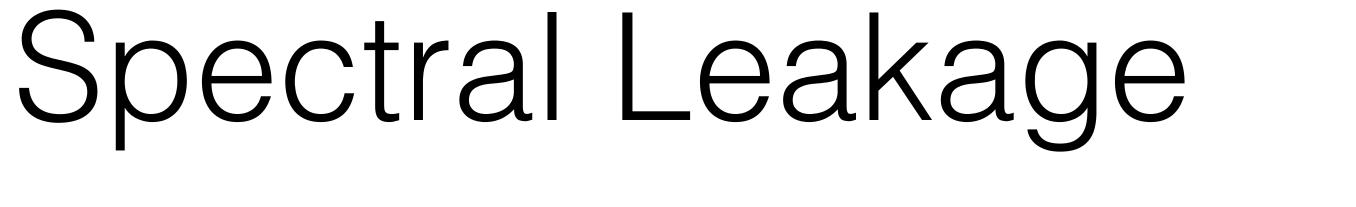


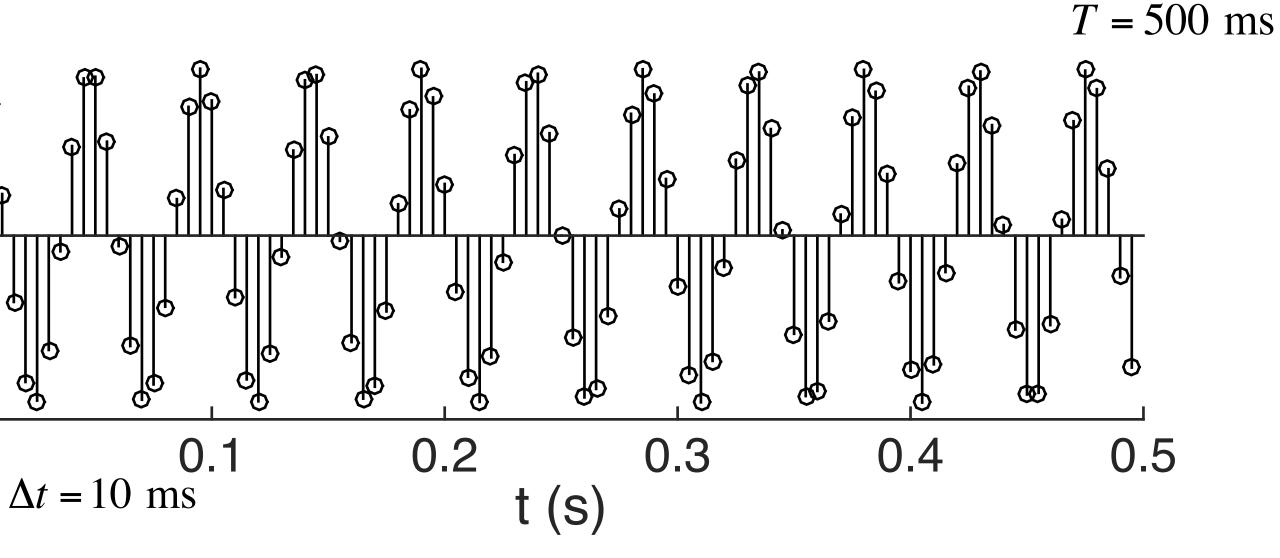
0

How do we ameliorate the edge "discontinuity"?

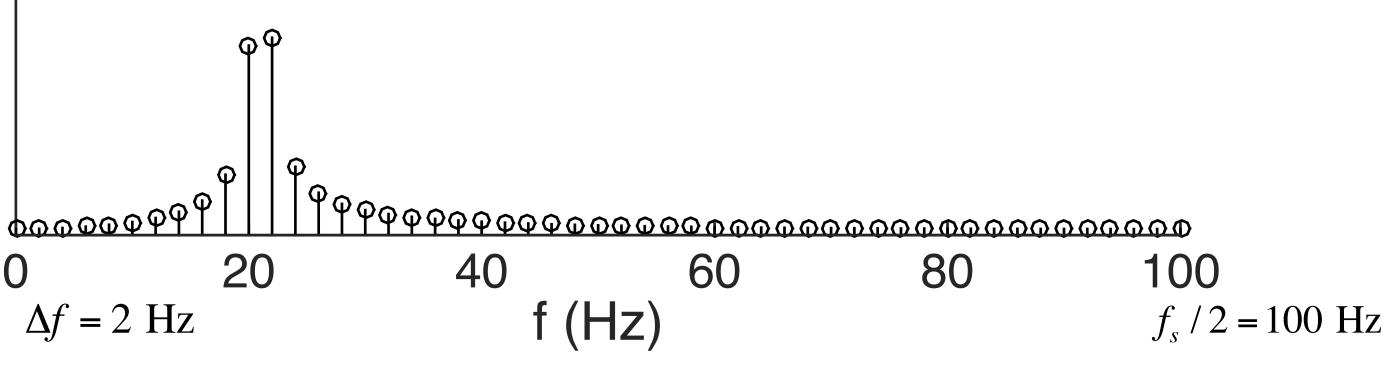
Modulate the signal by a window (i.e., "window" the signal).

> $x[t] = \cos(2\pi f_b t)$ =Jb f = 2 Hz





 $f_{\rm s} = 200 \; {\rm Hz}$







How do we ameliorate the edge "discontinuity"?

Modulate the signal by a window (i.e., "window" the signal).

> $x[t] = \cos(2\pi f_b t)$ = 21 Hz f_b $\Delta f = 2 \text{ Hz}$

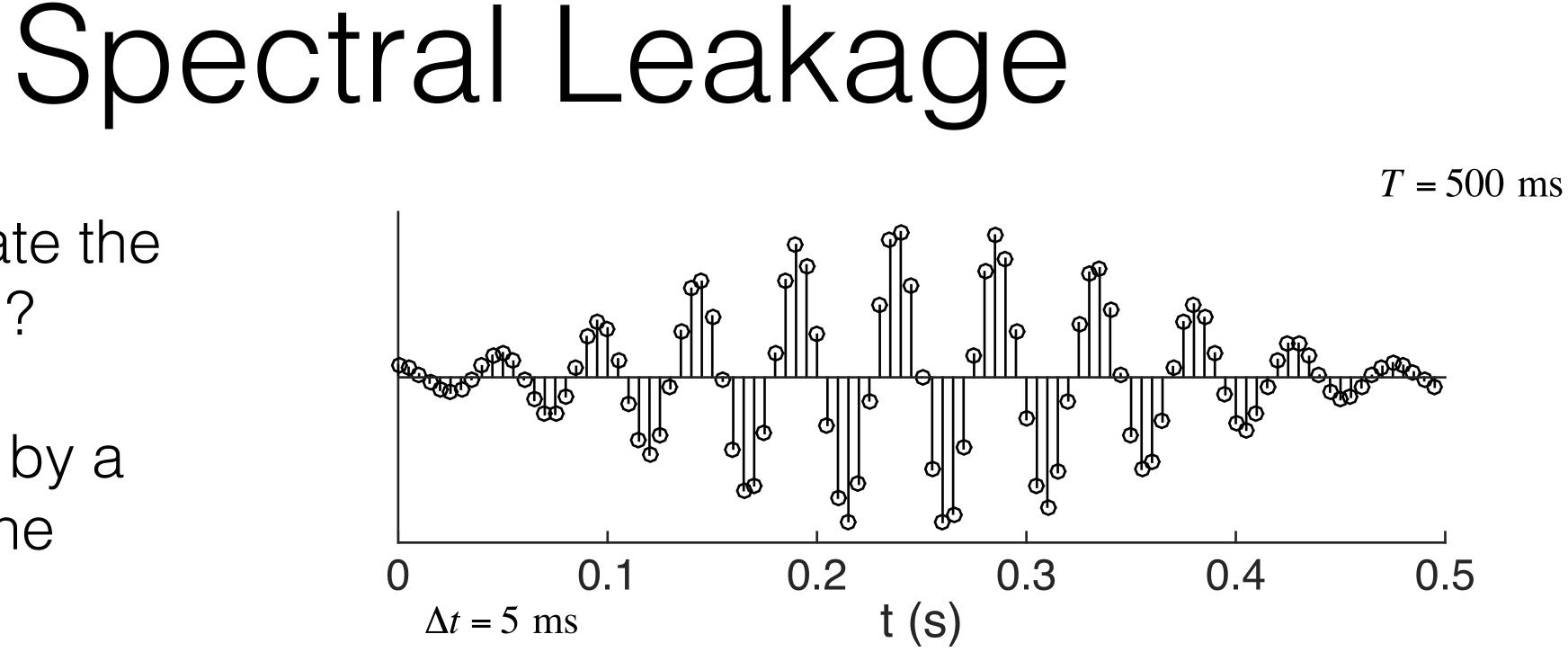




How do we ameliorate the edge "discontinuity"?

Modulate the signal by a window ("window" the signal).

> $x[t] = \cos(2\pi f_b t)$ = 21 HzJb $\Delta f = 2 \text{ Hz}$



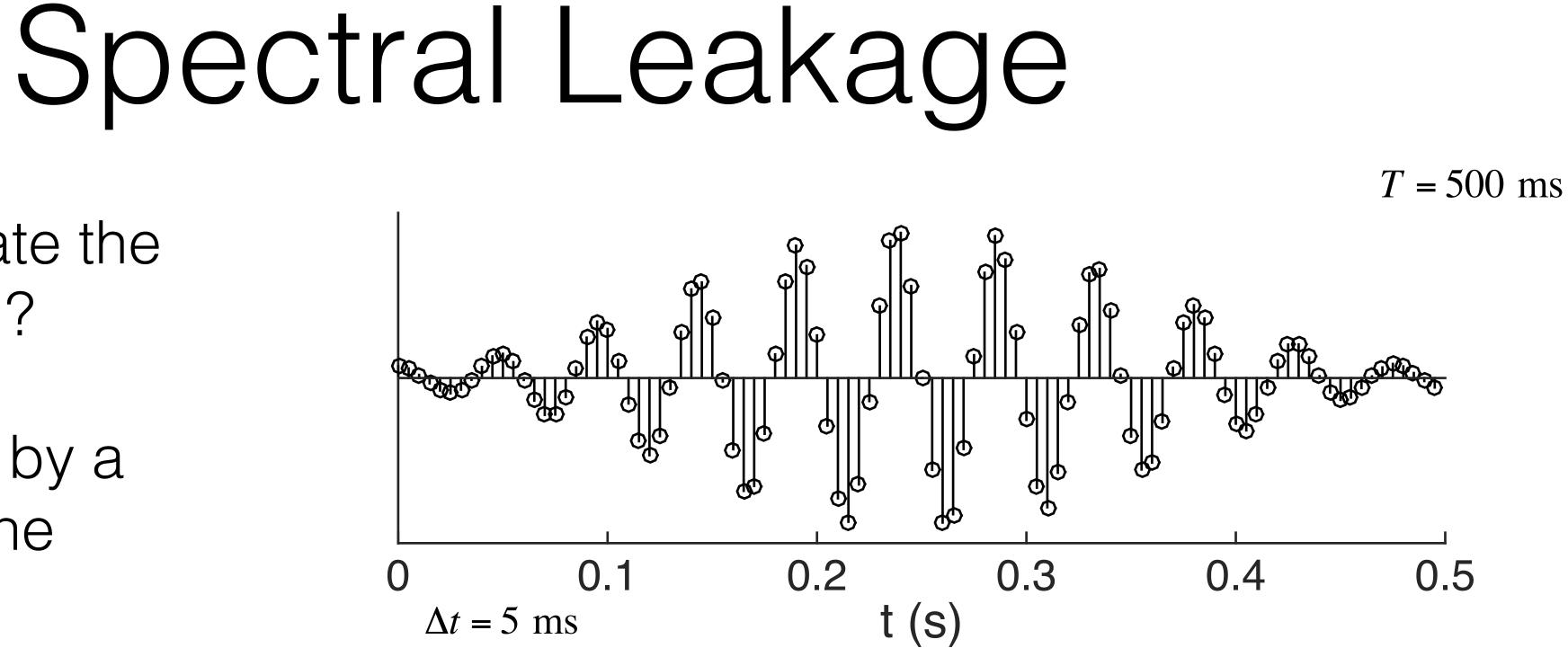


()

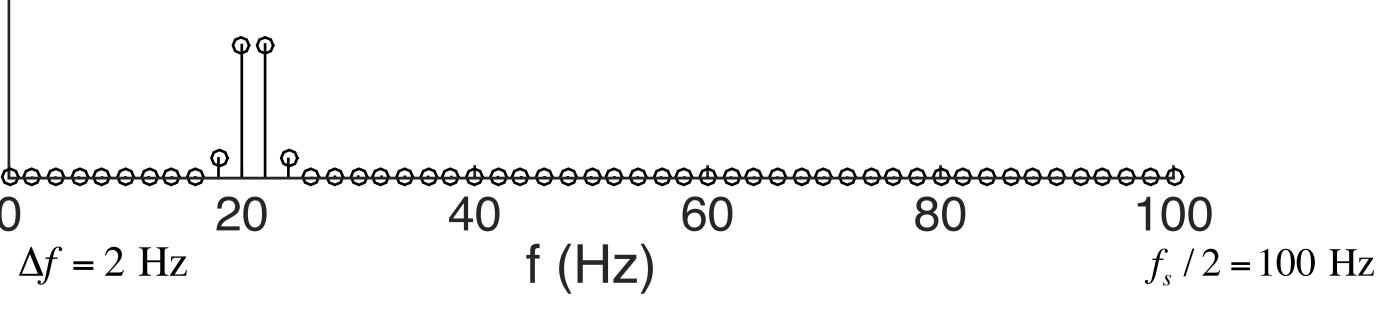
How do we ameliorate the edge "discontinuity"?

Modulate the signal by a window ("window" the signal).

> $x[t] = \cos(2\pi f_b t)$ = 2Jb f = 2 Hz



 $f_{\rm s} = 200 \, {\rm Hz}$





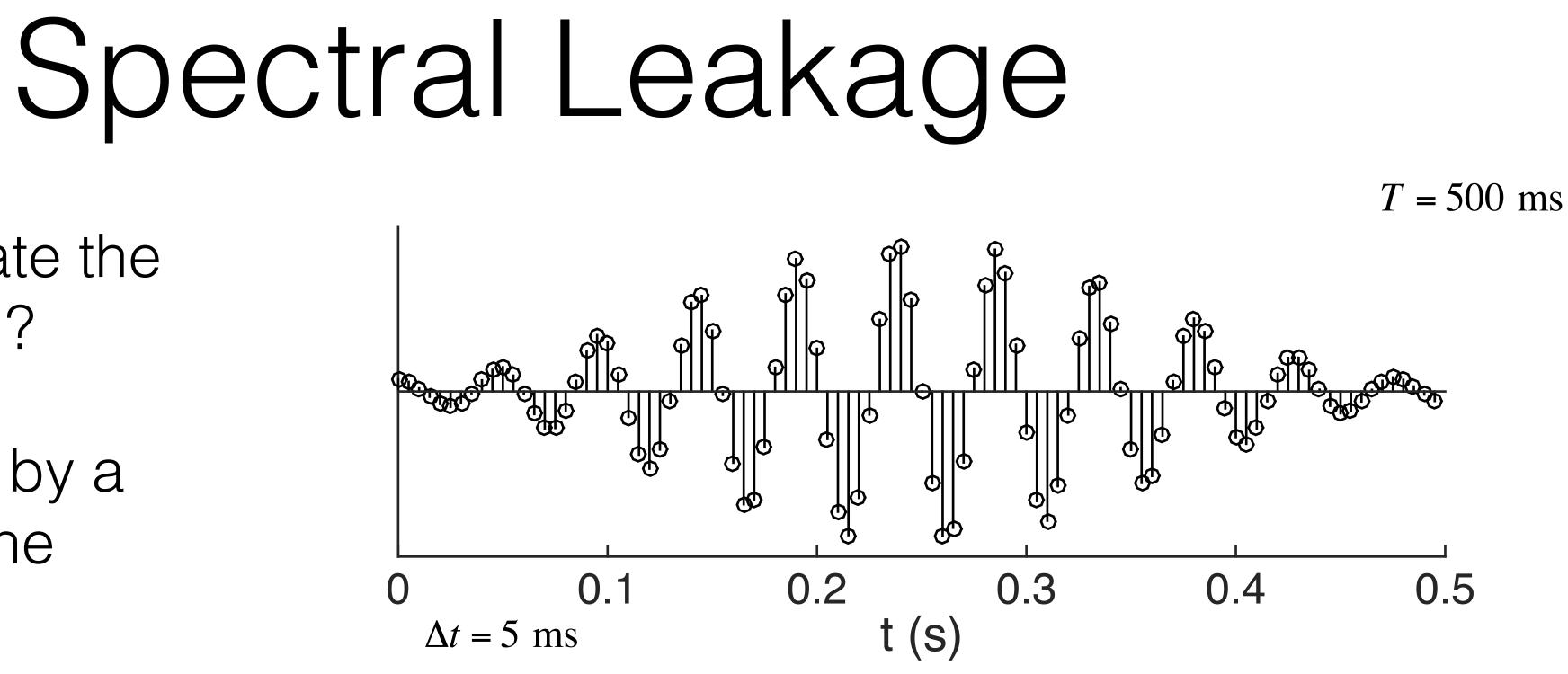


()

How do we ameliorate the edge "discontinuity"?

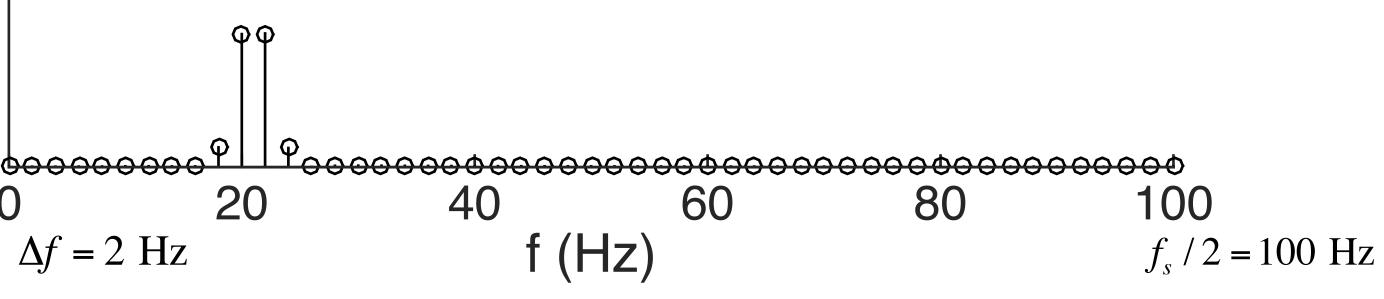
Modulate the signal by a window ("window" the signal).

> $x[t] = \cos(2\pi f_b t)$ = 2 Jb = 2 Hz



 $f_{\rm s} = 200 \; {\rm Hz}$

Spectral Leakage Attenuated!







Windowing & Frequency Resolution

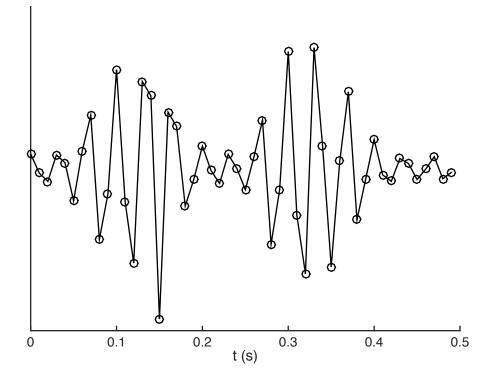
- Windowing to attenuate spectral leakage is critical for frequency estimation (spectral power, spectrogram, etc.).
- Achieves this by blurring the frequency resolution (typically by 2x).
- If you require spectral resolution of Δf , you also require a signal duration of not just $1/\Delta f$, but really $2/\Delta f$.
- For example, 1 Hz resolution, without spectral leakage corruption, requires ~2 s signal duration. 2 Hz resolution, without spectral leakage corruption, requires ~4 s signal duration.

Low Passing of Envelopes

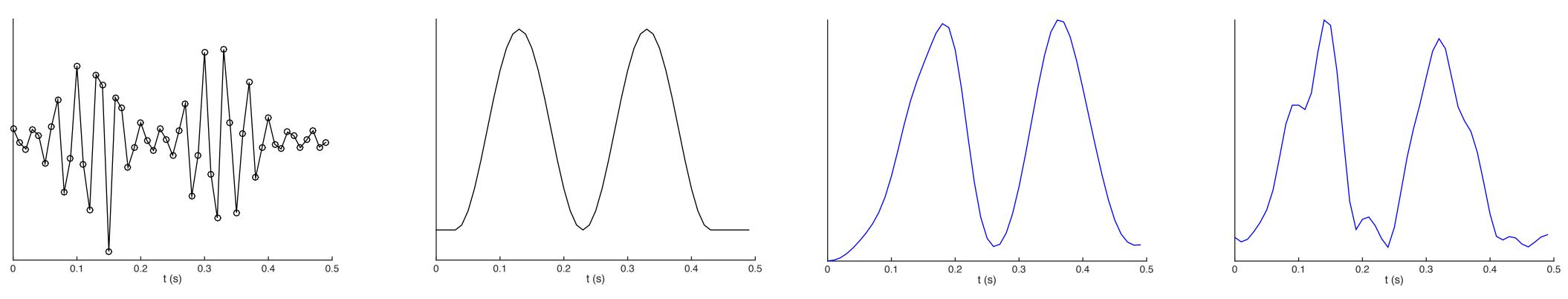
- An envelope is any slow amplitude modulation of a signal
- No single definition of envelope, except that it is slow and positive
- Commonly used definitions
 - Low passed half-wave rectified signal
 - Low passed magnitude of Analytic Signal ("hilbert" in Matlab).
- Note that the low pass filter is not optional
 - An envelope is any *slow* amplitude modulation of a signal.

Low Passing & Envelopes



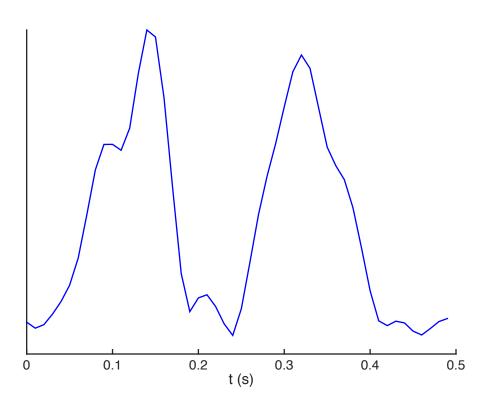


Actual Envelope



Low Passed Analytic Magnitude

Analytic Magnitude



- Filters: What They Do, and How They Do It
- Grab Bag:

Outline

• Fourier Transform: Why It's Useful, and What it Can/Cannot Do For You

• Filters: Why So Many Different Kinds? Which Should I Use and When?

• Use Causal Filters; Windowing is Good; Low-Pass your Envelopes

Conclusions

- Signal Processing is Complicated
- But not Too Complicated
- Mathematical Definitions will always Win/Tie over Intuition
- But Guided Intuition will put on a Strong Show
- Debugging using Guided Intuition faster than using Math \bullet